R4131 SERIES SPECTRUM ANALYZER

INSTRUCTION MANUAL

| Part |

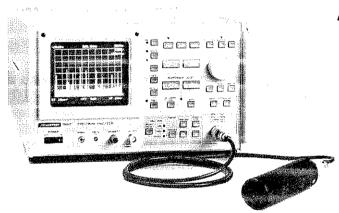
ADVANTEST, ADVANTEST CORPORATION

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R4131 SERIES SPECTRUM ANALYZER

INSTRUCTION MANUAL

MANUAL NUMBER EMOO 9504(A)



Applicable Instruments
R4131C
R4131CN
R4131D
R4131DN

| Column | C

The product is a Strategic Commodity subject to COCOM regulations.

It should not be exported without the proper authorization from Japanese government.

ADVANTEST CORPORATION

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First printing February 17, 1988 Printed in Japan

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1. GENERAL DESCRIPTION

Information and notes necessary to use this instrument for Operating Manual safety are written. Read before this instrument is used.



1.1 How to Use this Operation Manual

1.1 How to Use this Operation Manual

This manual proceeds from basic knowledge to application so that anyone can master the abundant functions of this equipment even when using such an intelligent spectrum analyzer for the first time. Those who are accustomed to using an intelligent spectrum analyzer can start the measurement at once merely by referring to [Chapter 4. OPERATING PROCEDURE]. The functional description of each key is given in [Chapter 3. DESCRIPTION OF PANEL SURFACE LAYOUT AND DISPLAY].

1.2 Outline of Products

The R4131 covers a band-width as wide as 10 kHz to 3500 kHz and is controlled by a microcomputer. This analyzer features easy confirmation of all measuring conditions, since its frequency span is 4 GHz to 50 kHz, resolution is 1 MHz to 1 kHz, level data resolution by a marker is 0.05 dB, tube surface dynamic range is 80 dB, and the setting conditions of the its major functions are shown on its CRT display.

The panel of this equipment enables its three major functions (center frequency, frequency span, and reference level) to be independent of each other, and its layout makes for excellent operability. In addition, the resolution band, sweep time and input attenuator values are set automatically by its AUTO feature.

Table 1-1 lists the other major functions of R4131.

Table 1-1 Major Function of R4131

		_								
Major function*	R4131C	R4131D	R4131CN	R4131DN						
Input impedance		50 Ω	75 Ω							
Accuracy in frequency display	±10 MHz	*100 kHz	±10 MHz	*±100 kHz						
QP value automatic operation		-k								
Antenna factor automatic operation	Standard mounting									
GPIB control										
Сору	Direct plott	ing with a plo	tter							
SAVE/RECALL function	Storing three setting conditions in its non-volatile memory.									
	Storing three display waveforms in its non-volatile memory.									
	Possible to set automatically at power ON.									
Displaying function	WRITE and VIEW Screen display									
	POSI PEAK POST/NEG POSI PEAK POSI, display display display disp									
Occupied band-width	Standard Configuration									

Note: *Where frequency ≤ 2.5 GHz after zero calibration

1.3 Before Starting the Use

1.3.1 Appearance Check and Accessory Check

After R4131 was received, first check flaws or damage in appearance that could have occurred during its transportation.

Next, check the standard accessories for their quantity and standards, referring to Table 1-2 for R4131C/D and to Table 1-3 for R4131CN/DN. If any flaw, damage, shortage in accessories, etc., is found, contact the nearest dealer or the sales and support offices.

Table 1-2 R4131C/D Standard Accessories

No.	Name	Specification	Stock No.	Q'ty	Remarks
1	Fuse	218005	DFT-AA5A	2	
2	Allen wrench	3 mm		1	
3	Input cable	MI-02	DCB-FF0386	1	
4	NC-BNC adapter	JUG-201A-U	JCF-AF001EX03	4	
5	Power cable	A01402	DCB-DD2428X01	1	
		420	JR4131	1	Japanese
6	Instruction manual		ER4131	1 '	English

Table 1-3 R4131CN/DN Standard Accessories

No.	Name	Specification	Stock No.	Q'ty	Remarks
1	Fuse	218005	DFT-AA5A	2	
2	Allen wrench	3 mm	444	1	
3	Input cable	D3S015(Black)	DCB-FF2928X01	1	
4	NC-BNC adapter	BA-A165	JCF-AF001EX04	1	
5	C15 adapter	NCP-NFJ	JCF-AF001EX06	1*	*R4131DN only
6	Power cable	A01402	DCB-DD2428X01	1	
			JR4131	4	Japanese
7	Instruction manual	APRIL 1997	ER4131		English

1.3.2 Environmental Conditions for Use

- (1) Refrain from using this equipment in a place subject to much vibration direct sunlight, and where corrosive gas is generated.

 The unit is designed for indoor use.

 Also, do not use it where the ambient temperature is outside 0°C to 50°C and relative humidity is less than 85%.

 If may occasionally be subjected to temperatures between 0°C and -10°C without degradation of its safety.
- 2 Since this equipment employs a suction type cooling fan to prevent the internal temperature from rising, install this equipment 10 cm or more from the wall, and do not place anything close to its back nor use this equipment in an incorrect position.
- 3 Although the equipment design for noise from the AC power supply line, use it allows where there is low noise as far as possible, and use a noise filter for noisy places.
- 4 The storage temperature range for this equipment is -20°C to +70°C. When this equipment is not used for a long period of time, store it in a dry place away from direct sunlight, covered with vinyl or placed in a cardboard box.

WARNING

1.3.3 Before turning This Analyzer on

Before any other connection is made, make sure this instrument has
been properly grounded through the protective conductor of the AC
power cable to a socket outlet provided with protective earth
contact. Any interruption of the protective (grounding) conductor,
inside or outside the instrument, or disconnection of the protective

2. Before turning this analyzer on, make sure that it is set to the voltage of the power supply (Refer to Table 1-4.).

earth terminal can result in personal injury.

3. If the fuse rating is not as specified, this unit may be broken.

(1) Power Supply Condition

The power supply conditions of R4131 are given in Table 1-4.

Table 1-4 Power Supply Conditions

Power supply	Condition
Input voltage Frequency Power consumption	90 V to 132 V or 198 V to 250 V rmp 48 to 66 Hz Less than 120 VA

CAUTION

When the power supply does not conform the conditions given in Table 1-4, this equipment could break down.

(2) Check for Fuse

The fuse of the power supply AC line is T5 A/250 V for either 90 V to 132 V or 198 V to 250 V in input voltage. Check the fuse set in the power connector of the rear panel as shown in Figure 1-1.

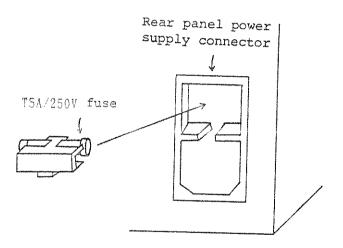


Figure 1-1 Check for Fuse

CAUTION

When used with a fuse not in the specified value, this equipment could break down.

INSTRUCTION MANUAL

1.3 Before Starting the Use

(3) Check for Power Supply Cable

Turn OFF the POWER switch on the front panel of this equipment. Next, connect the attached power supply cable to the AC LINE connector. The plug is a 3-pin type and the round pin in the middle is the earth.

When using the R4131C, R4131CN, R4131D, R4131DN defend the following.

- Connect power plug with the outlet prepared the protective earth terminal.
- . Do not use extension cable without a protective conductor.

When a 2-pin adaptor is used, be sure to connect either the ground wire led from the adaptor or the ground terminal located on the rear panel to the ground.

WARNING -

Any interruption of the protective conductor inside or outside the R4131C, R4131CN, R4131DN or disconnection of the protective earth terminal is likely to make the instrument dangerous. Intentional interruption is prohibited.

(4) Maximum Input

CAUTION .

The maximum level that can be input to the INPUT connector of this equipment is as follows. When a voltage beyond this level is input, the input mixer unit. etc., breaks down, entailing heavy repair expense. When the input signal level might exceed the maximum input level for this equipment, be sure to lower the signal level sufficiently by using an external attenuator, etc., and then input it.

R4131C/D

Maximum input level: +20 dBm (INPUT ATT 20 dB or

more)

AC coupe

±25 VDC max.

R4131CN/DN

Maximum input level:

+127 dBu (INPUT ATT 20 DB or

more)

AC couple : ±25 VDC

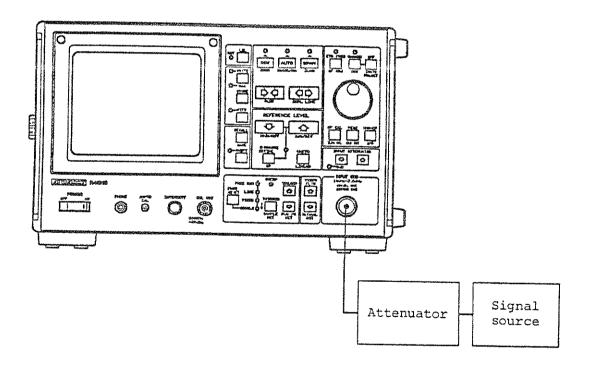


Figure 1-2 Input of Excessive Signal Level

SPECTRUM ANALYZER INSTRUCTION MANUAL

2. Using R4131 for the First Time

2. USING R4131 FOR THE FIRST TIME

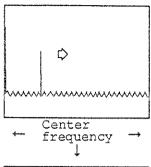
This chapter describes the fundamentals of operating R4131 for those using for the first time.

Note: Before turning ON the power for this equipment, read through Section 1.3, Before Use.

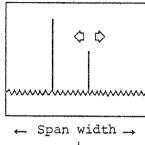


2.1 Screen of Spectrum Analyzer

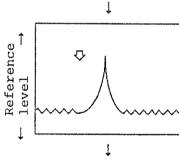
Figure 2-1 shows the screen of R4131, indicating the relationship among the center frequency, span width, and reference level.



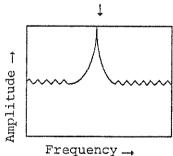
When the center frequency is changed, the position of the screen in the horizontal direction moves right and left.



When the span side is changed, the size of the screen in the horizontal direction increases or decreases.



When the reference level is changed, the position of the screen in the vertical direction moves up and down.



On the screen, the horizontal direction represents the frequency and its vertical direction represents the amplitude (level).

Figure 2-1 Screen of Spectrum Analyzer

2.2 Basic Operating Procedure

While operating actually using the calibration signal of this equipment, learn how to use the most important keys.

(1) Initialization Screen

First, turn ON the power. When the power ON automatic setting function is in operation or a key is pressed after the power ON, press the $\bigcap_{i=1}^{SHIFT}$ and $\bigcap_{i=1}^{SHIFT}$ keys to initialize the screen as shown Figure 2-2.

Note: See Section 4.17, Power ON Automatic Setting.

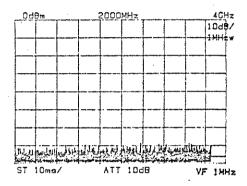


Figure 2-2 Initialization screen

(2) Input of Measurement Signal

Referring to Figure 2-3, input the calibration signal of this equipment to the terminal INPUT.

Calibration signal

R4131C/D Frequency: 200 MHz ±30 kHz

Level : $-30 \text{ dBm } \pm 0.5 \text{ dB}$

R4131CN/DN Frequency: 200 MHz ±30 kHz

Level : 80 dBµ ±0.5 dB

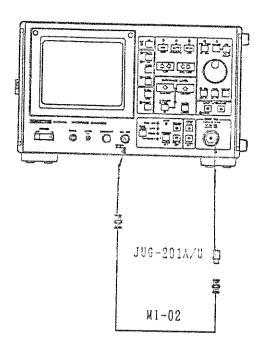
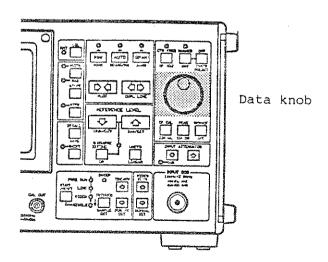


Figure 2-3 Input the Calibration Signal

(3) Setting of Center Frequency

Since the calibration signal is already known to be 200 MHz in frequency and -30 dBM in output, set the center frequency to 200 MHz. Turn the data knob counterclockwise to set the spectrum of the input signal to the center of the CRT.



Turn the data knob, then the waveform moves in the horizontal direction (Figure 2-4).

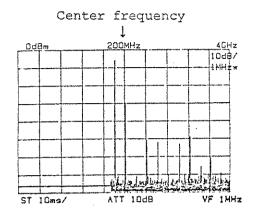
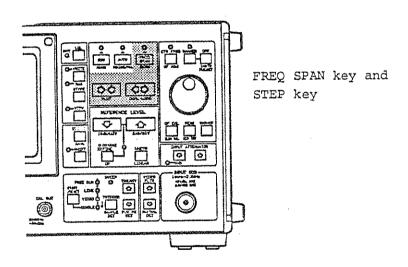


Figure 2-4 Setting Center Frequency to 200 MHz

(4) Setting of Frequency Span

Since the frequency span of this equipment is set very wide to 4 GHz on initialization, change it to 2 MHz.



Press the $\frac{}{\text{PLOT}}$ key, then the frequency span becomes narrower in steps of 1-2-5 (Figure 2-5).

If the spectrum deviates from the center in this case, turn the data knob to change the center frequency and make it narrower while seizing the spectrum in the center.

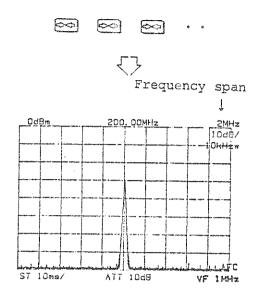


Figure 2-5 Setting the Frequency Span to 2 MHz

Since the rom is selected at initialization in the resolution band width, it is set to the maximum value automatically according to the setting condition of the frequency span.

(5) Setting of Reference Level

The reference level of this equipment -- the horizontal line on the top of the screen grid -- is set to 0 dB at initialization. Change it to -30 dB and set the spectrum of the calibration signal to the reference level.

REFERENCE LEVELL

10d8/DIV 2dB/DIV

COARSE

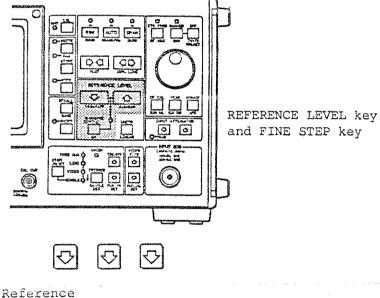
FINE

When the REFERENCE LEVEL key is pressed, the reference level goes up and down in steps of 10 dB. It is set to 10 dB/DIV at initialization.

When the COARSE or FINE key is pressed and FINE is selected, the LED on the upper right of this key lights and the mode is set to FINE.

10dB/DIV 2dB/DIV

The 10DB/DIV or 2DB/DIV key is used to change the set value in 1-dB steps.



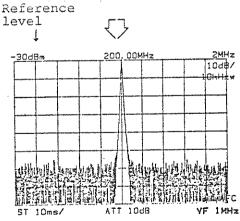


Figure 2-6 Setting the Reference Level to -30 dB

(6) How to Use the MARKER Key

By using the MARKER, you can read the frequency directly at the displayed marker point and level data.

The following is a description of this procedure:

When the \Box key is pressed, the LED on its upper lights and the marker (\Diamond) appears on the center frequency axis.

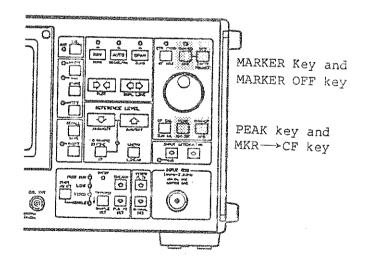
Move the marker with the data knob to set the marker to the measured signal (Figure 2-7). The data of the signal can be read directly according to the marker frequency and its level.

When the marker is cleared, press the key.

- PEAK search

When the $\ \ \$ key is pressed, the marker moves to the maximum level waveform displayed.

- MarKeR → Center Frequency



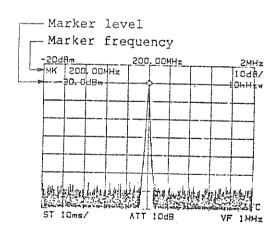


Figure 2-7 Setting the Marker to the Peak of the Measured Signal

SPECTRUM ANALYZER INSTRUCTION MANUAL

2.2 Basic Operating Procedure

(7) How to Improve Frequency Accuracy According to the Correction Routine

When the ____, zero cal key is pressed, the frequency correction routine, ZERO CAL, is executed. (Then, the "ZERO CAL" is displayed on the bottom right of the CRT.) When the ZERO CAL is executed before measurement starts, the center frequency accuracy is improved as shown

R4131C/CN Center frequency accuracy 0 to 3.5 GHz : $\pm 10 \text{MHz}$ R4131D/DN Center frequency accuracy 0 to 2.5 GHz : $\pm 100 \text{kHz}$ 2.5 GHz to 3.5 GHz: $\pm 100 \text{MHz}$

(8) Warm-up Time

below:

To use this equipment at the specified accuracy, take 30 minutes or more for its warm-up.

MEMO O

R4131 SERIES SPECTRUM ANALYZER INSTRUCTION MANUAL

3. Description of Panel Surface and CRT Display

3. DESCRIPTION OF PANEL SURFACE AND CRT DISPLAY

This chapter describes each section on the panel and display screen of this equipment.



3.1 Description of Front Panel

Figure 3-1 shows the front panel.

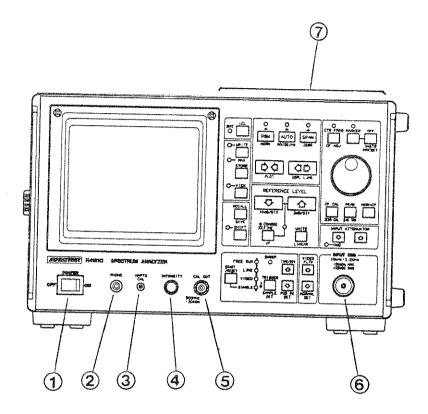


Figure 3-1 Description of Front Panel

1 Power ON/OFF Switch

The waveform is displayed at power ON and after a self-test (self-diagnosis).

2 Earphone Jack

This is a jack used for an 8-ohm earphone, to monitor the received modulated wave with the earphone (TR16191) when this equipment is used as a fixed tuning receiver.

3 Variable Resistor for Correcting Level Display

This is a variable resistor to correct the level display of this equipment.

3.1 Description of Front Panel

- 4 Variable Resistor for Adjusting Brightness
 This is a variable resistor to correct
 the brightness of the CRT display.
- 5 Output Connector of Correction Signal

For R4131C/D

Outputs the signal of 200 MHz and -30 dB.

For R4131CN/DN

Outputs the signal of 200 MHz and 80 dB.

(6) Input Connector

For R4131C/D

The maximum input level is $+20~\mathrm{dBm}$ and $\pm25~\mathrm{VDC}$ max. when the input attenuator is more than 20 dB.

For R4131CN/DN

The maximum input level is $+127~\mathrm{dBu}$ and $\pm25~\mathrm{VDC}$ max. when the input attenuator is more than 20 dB.

CAL OUT



200MHz -30dBm

CAL OUT



200MHz 80dB μ

INPUT 50 Ω 10kHz-3.5GHz



+ 20dBm MAX ± 25VDC MAX

INPUT 75 Ω 10kHz-3.5GHz



127dB μ MAX ± 25VDC MAX

CAUTION

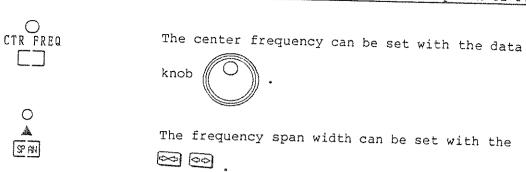
Note that the 75 Ω input connector is vulnerable when using R4131CN/DN. Unless the 75 Ω NC-BNC type is used for the BNC adaptor, the input connector breaks down very easily.

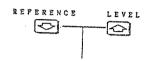
Analyzer Control Key

Three basic keys of the spectrum analyzer -- center frequency, span width, and amplitude level -- and this equipment are separated into three sections to be independent of each other for excellent operability.

SPECTRUM ANALYZER INSTRUCTION MANUAL

3.1 Description of Front Panel



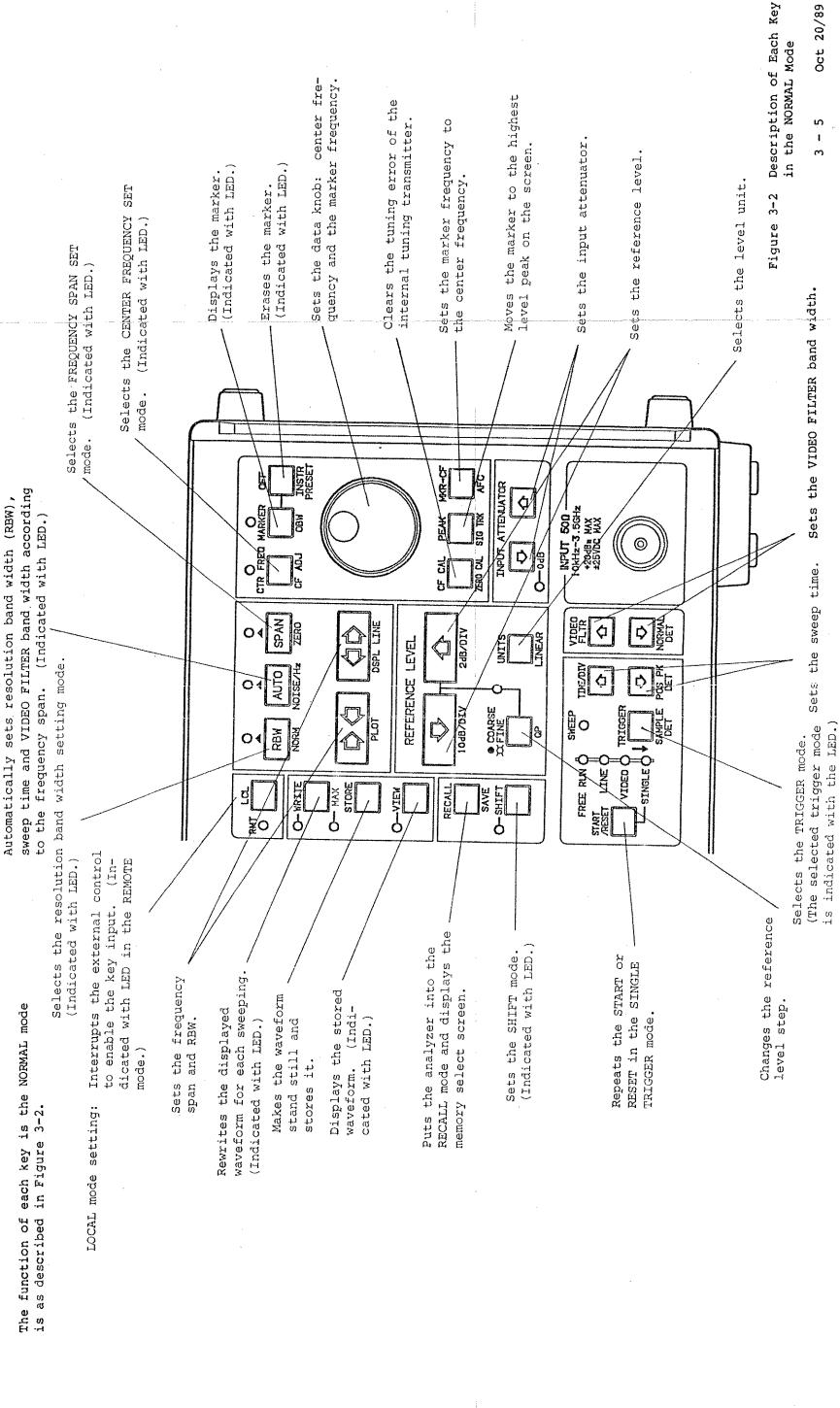


This key can set the reference level.

Also, pressing the SHIFT key sets the SHIFT mode and executes the function whose name is inscribes in blue immediately below the next key you press.

3.2 Description of Each Key (in the NORMAL mode)

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3.3 Description of Each Key (in the SHIFT mode)

3.3 Description of Each Key (in the SHIFT mode)

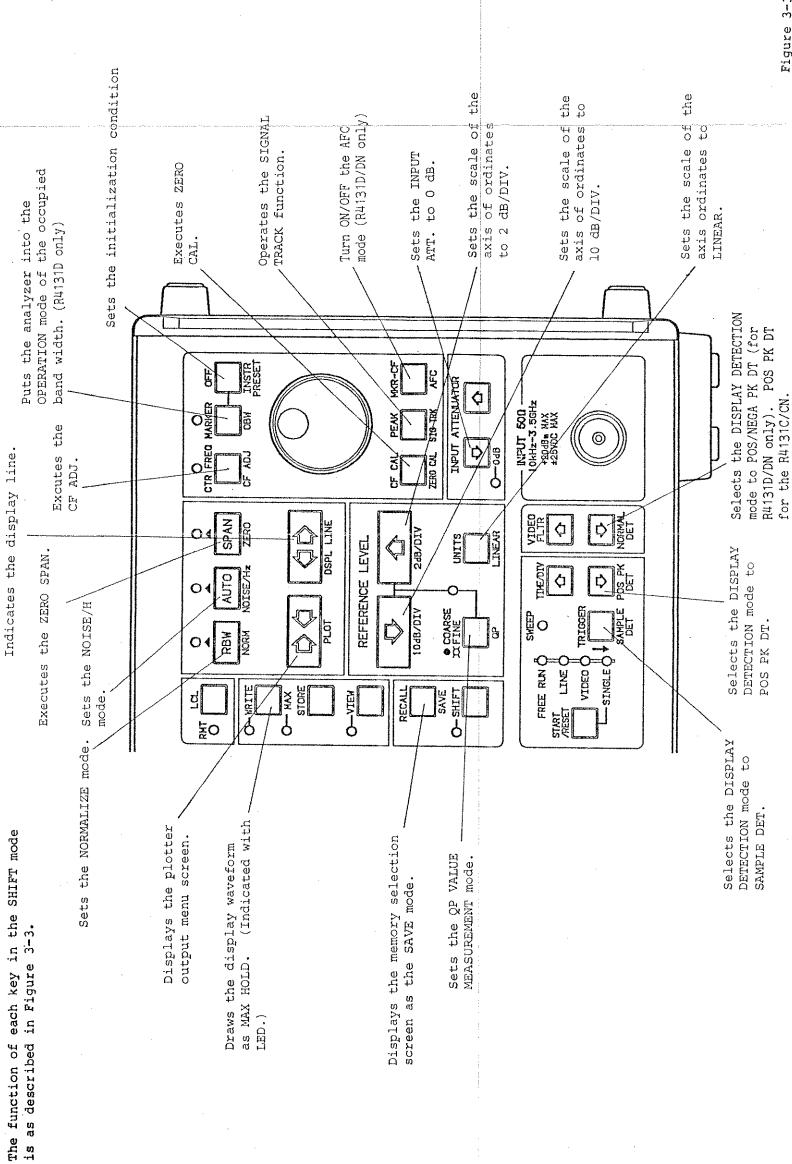


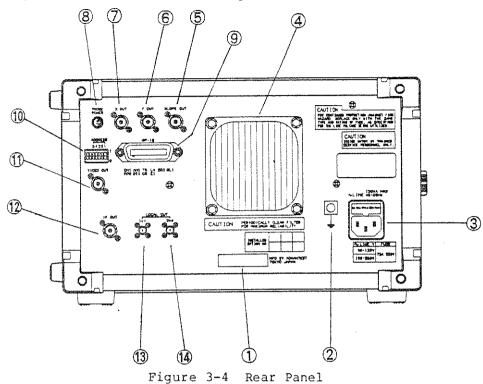
Figure 3-3 Description of Each Key in the SHIFT Mode

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					 		 ,	***************************************		
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3.4 Description of Rear Panel

The rear panel is as described in Figure 3-4.



(1) Serial No.

A serial No. of this equipment is printed.

Japan only -

(2) Ground terminal

Used to connect the unit frame to the ground when neither 3-pin nor 2-pin power cable connector cannot be used.

(3) Connector for AC Power Cable

This connector is a 3-pin type, and the center pin is a terminal for grounding.

When the upper lid is drawn out, the power fuse can be taken out.

(4) Cooling Fan

This is a suction type cooling fan.

Jan 27/94

(5) Connector for Slope Correction

This connector is used to output the slope correcting voltage 2 $\ensuremath{\text{V/GHz}}$ for the tracking generator.

- 6 Output Connector to XY Recorder of WRITE Waveform Y. OUT ... approx. 0 to 4 V, and output impedance approx. 220 Ω
- 7 Output Connector to XY Recorder of WRITE Waveform X. OUT ... approx. -5 V to +5 V, and output impedance approx. 10 $K\Omega$
- (8) Connector for Probe Power

This is the power supply for accessories, e.g., active probe, etc.

3 PROBE 2 1 : NC 2 : GND 3 : -15V 4 1 4 : +15v

9 GPIB Connector

This is a terminal used when this equipment is connected to an external controller or plotter with the GPIB cable.

- (10) Address Switch for GPIB
 The GPIB address is set using 1- to 5-digit switches.
- Output Connector to External CRT Display and VIDEO Plotter, etc. Output impedance ... approx. 75 Ω and 1 V_{p-p} , including the composite signal.
- ① Output Connector for IF Monitor

This terminal is used to supply IF output 3.58 MHz and approx. 50 Ω . The output level can be set according to the input attenuator and reference level.

- 13 LOCAL OUT Connector for Tracking Generator

 1st LOCAL OUT ... more than -5 dBm at 4 GHz to 7.5 GHz
- 14) LOCAL OUT Connector for Tracking Generator
 2nd LOCAL COUT ... more than -5 dBm at 3.77 GHz.

--- CAUTION -

When connector (3) and (4) for the tracking generator is used while opened, accurate measurement can not be occasionally done. Connect with the tracking generator or if you do not use the connector, install attached terminal instrument.

3.5 How to Read CRT Display Indication

Various setting conditions are displayed on the screen. Their indication and the contents of each indication are shown in Figure 3-5.

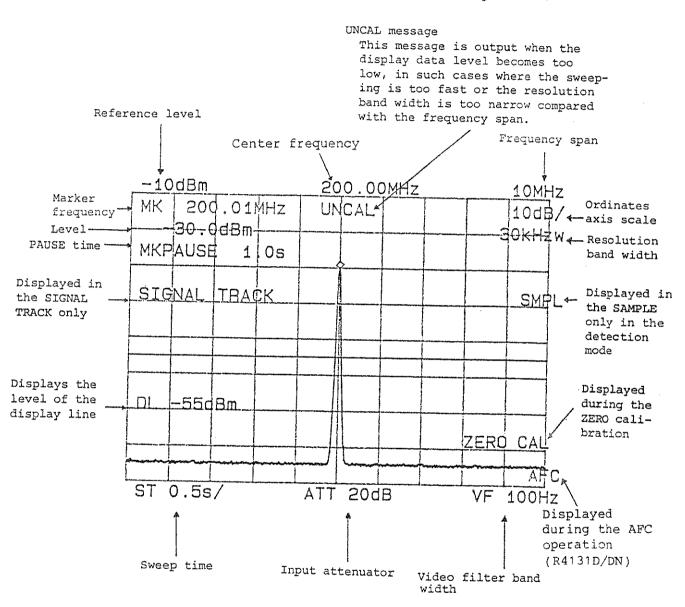


Figure 3-5 Indication of CRT Display

R4131 SERIES SPECTRUM ANALYZER INSTRUCTION MANUAL

4. Operating Method

4. OPERATING METHOD

This chapter describes the basic operating method of this equipment with same measuring examples included.

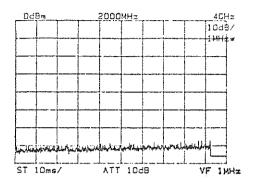


4.1 Initialization

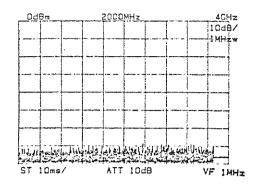
SHIFT \square When the \square , $\Pr_{\text{PRESET}}^{\text{INTER}}$ key is pressed, the equipment is set to the initial values as shown in Table 4-1.

Table 4-1 Initialization Condition

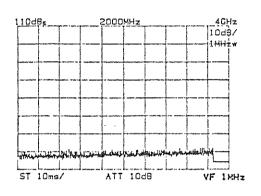
Set item	Initialization condition
Center frequency	2000 MHz
Frequency span	4 GHz
Reference level	0 dBm (:R4131C/D) 110 dBu (:R4131CN/DN)
Resorution band width	1 MHz
VIDEO FLTR band width	1 MHz
SWEEP TIME	10 mS
INPUT ATT.	10 dB
TRIGGER MODE	FREE RUN
Marker	OFF
Ordinates axis scale	10 dB/DIV
DETECTION MODE	POSI-NEGA PEAK (:R4131D/DN) POSI PEAK (:R4131C/CN)
TRACE MODE	WRITE



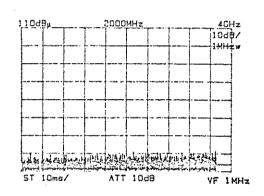
(a) R4131C



(B) R4131D



(c) R4131CN



(d) R4131DN

Figure 4-1 Initial Screen

SPECTRUM ANALYZER INSTRUCTION MANUAL

4.	2	Center	Frequency
- •	-	CGIICGI	TIEUUENCV

4.2 Center Frequency

The equipment is set to the CTR FREQ CHANGE mode at the initialization of the data knob. However, when it is set to the MARKER CHANGE mode, press the ____ key. Then the LED on the key lights and the equipment is set to the CTR FREQ SET mode.

When the data knob is turned, the center frequency changes in a range from 0 MHz to 3620 MHz.

The set resolution is 1/200 of the frequency span.

— Center Frequency Accuracy -

The center frequency accuracy becomes the following range after the execution of the ZERO CAL in the local feed through (zero waveform):

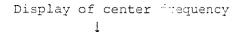
R4131C/CN

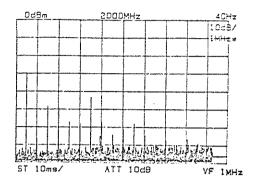
0 Hz to 3.5 GHz : ± 10 MHz or less

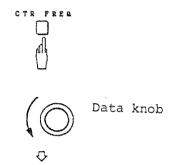
R4131D/DN

0 Hz to 2.5 GHz : ± 100 kHz or less

2.5 GHz to 3.5 GHz: ± 10 MHz or less







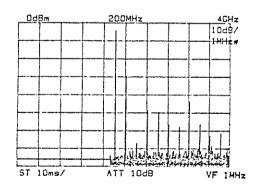


Figure 4-2 Change in Center Frequency

SPECTRUM ANALYZER INSTRUCTION MANUAL

4.3 Function to Improve Center
Frequency Accuracy

4.3 Function to Improve Center Frequency Accuracy

(1) AFC Function (only in R4131D/DN)

Since the AFC circuit is mounted in R4131D/DN, the AFC turns ON from when the frequency span becomes lower than 200 MHz and displayed as AFC on the bottom right of the screen. Consequently, the center frequency accuracy becomes ± 100 kHz or less after the execution of the ZERO CAL, described later. (It is confined to the case where the center frequency is 0 Hz to 2.5 MHz, however.)

To use this equipment with this AFC function kept OFF, key in the and \bigcap_{AFC} keys. (When the AFC is turned OFF, the tracking time is shortened and the total sweep time becomes shorter.)

To use the equipment with the AFC kept ON again, press the \bigcap_{AFC} and keys, then the AFC circuit starts operating.

(2) ZERO CALibration

Press the and zero cal keys, then ZERO CAL is executed. ("ZERO CAL" is then indicated on the bottom right of the screen.)

After correcting the center frequency 0 MHz in the local feed through (zero waveform), the equipment returns to the setting before the execution of ZERO CAL, thus improving the center frequency accuracy.

Incidentally, although the ZERO CAL data is stored in the non-volatile memory, execute the ZERO CAL over again to read its correct value.

(3) CF CALibration

Press the key, the CF CAL and degausing are executed. Since this equipment uses an oscillator capable of sweeping a wide band width as its local oscillator, an error occurs in the oscillation frequency for the setting when the center frequency is changed sharply (more than 1 GHz) where the frequency span is narrower (less than 200 MHz). This error can be removed by executing the CF CAL. To change the center frequency of the R4131 by 1 GHz or more, the frequency span is widened (as 2 GHz or 4 GHz span) in general. (Since the center frequency set resolution is 1/200 of the span, the center frequency does not move in big steps unless the span is widened.) Consequently, the sweeping is made under the status where the span is wide, and the degausing is executed naturally. No CF CAL need be executed in this case.

Usually, it is not necessary to use this CF CAL. Use it only to move sharply the frequency where the span is narrow in the GPIB control, etc. CF CAL is not executed when the AFC function is turned ON in the R4131D/DN.

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(4) CF ADJustment

Press the , cF ADJ is executed. By using this function, the center frequency accuracy can be improved further using the known input signal. The following is a description of the case where 2.2 GHz (11 times of CAL OUT 200 MHz) is used as the known frequency signal to read the value of an unknown frequency in the vicinity of 2.2 GHz.

- (1) Set the center frequency to 2.2 GHz. (See Figure 4-3 (a).)
- 2) Make the frequency span narrow in a range from which the spectrum does not protrude from the tube surface. (See Figure 4-3 (b).)
- 3 When the , cr AD, keys are pressed, the frequency display remains unchanged, but the spectrum moves to the center and the center frequency accuracy becomes 11 times the CAL OUT signal accuracy. (See Figure 4-3 (c).)
- 4 Input an unknown frequency and read the frequency. (See Figure 4-3 (d).)

Although the value of the unknown frequency is obtained as 2199.5 MHz in this example, care should be taken, because value indicates the error of the CAL OUT signal and also the marker error.

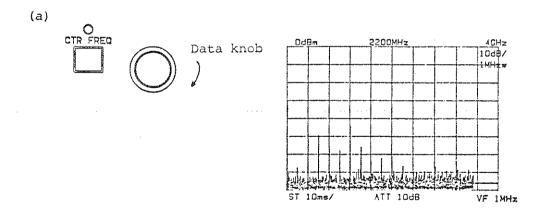
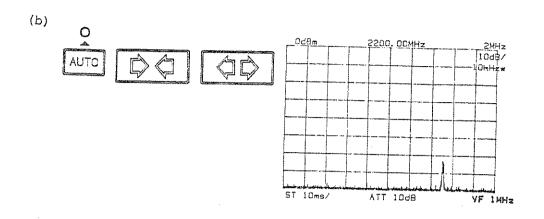
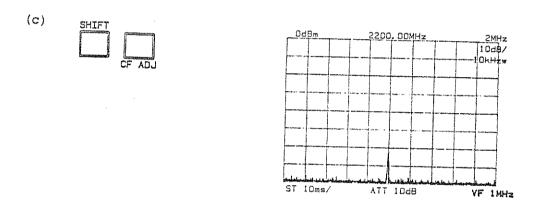


Figure 4-3 CF ADJ

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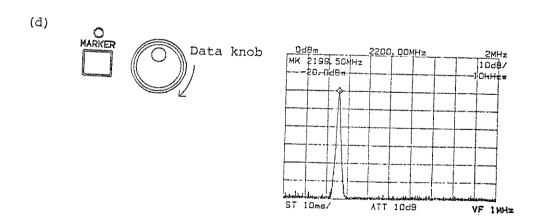


Figure 4-3 CF ADJ (cont'd)

4.4 Frequency Span

When the FREQUENCY SPAN SET mode is selected, the frequency range from

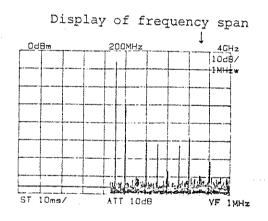
4 GHz to 50 kHz can be set with 1-2-5 steps by pressing the \Leftrightarrow or \Leftrightarrow key. The 1/10 of the set frequency span becomes the frequency span of one scale of the quadrature axis.

If the spectrum deviates from the center of the screen when the frequency span is narrowed, return the spectrum to the center of the screen by turning the data knob.

(1) What Is Zero Span (Displayed in the Time Axis)?

Pressing \square , and $\underset{\text{ZERO}}{\square}$ keys sets the ZERO SPAN mode, in which this equipment functions as a fixed tuning receiver and becomes a tube surface quadrature axis display. To clear this ZERO SPAN mode, press the \bowtie or \bowtie key.

When either key is pressed, the frequency span returns to the span before the setting of the ZERO SPAN mode.



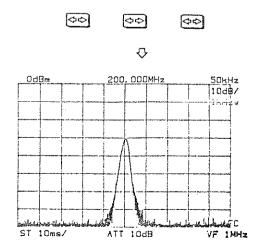


Figure 4-4 Making the Frequency Span Narrow and Spectrum Expand

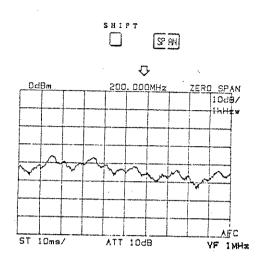


Figure 4-5 ZERO SPAN Mode

4.5 Interlocking Function (AUTO)

When the ADD key is pressed and the LED on the key is lit, the frequency span ,resolution band width (RBW) and sweep time are all interlocked to be

set to the optimum condition when the or or key is pressed. Incidentally, when the video filter band width (see Section 4.11) is changed, the video filter band width and sweep time are interlocked to be set to the optimum condition automatically.

4.6 Resolution Bank Width (RBW)

When the LED on the $\fbox{RU 10}$ key is lit, the resolution band width is interlocked with the frequency span to be set automatically.

When Rew key is pressed and then or or, the resolution band width can be set manually. When the or key is pressed, the spectrum narrows and the resolution rises. It is therefore possible to separate the equipment from the nearby noise of the measured spectrum, or to separate spectrums themselves.

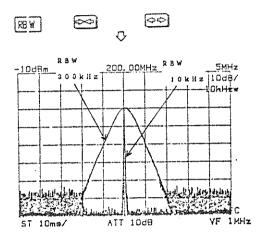


Figure 4-6 Change in Resolution Band Width

4.7 Reference Level and Ordinate Axis Scale

(1) Reference Level

The reference level is the top of the quadrature axis on the screen. REFERENCE LEVEL

By pressing the \bigcirc key, it is possible to set a range of -69.75 dBm to +40 dBm for R4131C/D and 40.25 dB μ to 150 dB μ for R4131CN/DN with a resolution of 0.25 dB maximum.

REFERENCE LEVEL

Each time the key is pressed, the reference level goes up or down by one step.

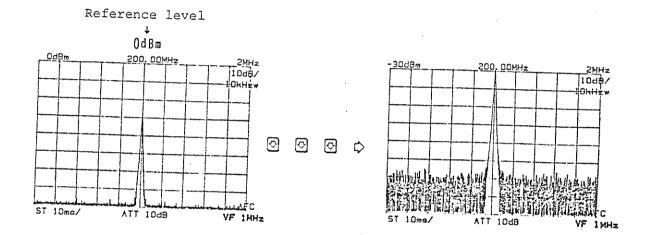


Figure 4-7 Change in Reference Level

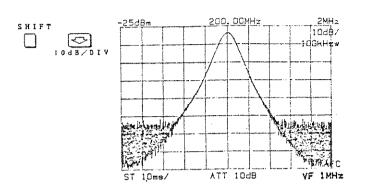
(2) Quadrature Axis Scale (dB/DIV)

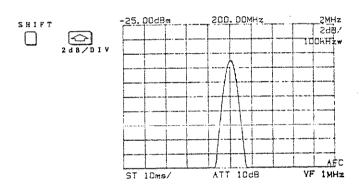
When the and 10dB/DIV keys are pressed, the ordinates axis scale is set to 10 dB/DIV.

When the and and address are pressed, the ordinates axis scale is set to 2 dB/DIV.

When the and weys are pressed, the ordinates axis scale is set to LINEAR.

In LINEAR, the lower end of the screen grid becomes 0 V.





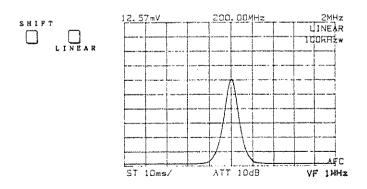


Figure 4-8 Ordinates Axis Scale

(3)	Reference Level Step Width (COARSE/FINE)
	A FINE 9
	When COARSE/FINE is selected in (the LED lights when
	selected to FINE), the step width becomes as shown in the table below
	at 10 dB/DIV and 2 DB/DIV:

Ordinates axis scal	e (dB/DIV)	Step width			
		COARSE	FINE		
10 dB/DIV 2 dB/DIV		10 dB 1 dB	1 dB 0.25 dB		

(4) Unit (UNITS)

When the \square key is pressed, four types of units, dBm, dB μ /m (A through D) and dBmV can be selected in the reference level. The dB μ /m is described in Section 4.18 Measurement of Electric Field Intensity.

(5) Calibration of Ordinates Axis Level

The ordinates axis level can be calibrated by setting the signal level to -30 dBm using the variable resistor for calibrating the level display on the front panel with the calibration signal 200 MHz CAL. The ordinates axis level may change later in some cases if the calibration is executed before the equipment has warmed up for 30 minutes.

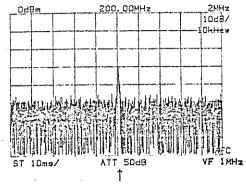
Care should also be taken because the ordinates axis level can change when the working temperature changes sharply.

4.8 RF Input Attenuator

INPUT ATTENUATOR

Pressing the ② , ② key sets the value of RF ATT between the INPUT connector and first mixer from 10 DB to 50 dB in steps of 10 dB. It is usually interlocked with the reference level to be set automatically.

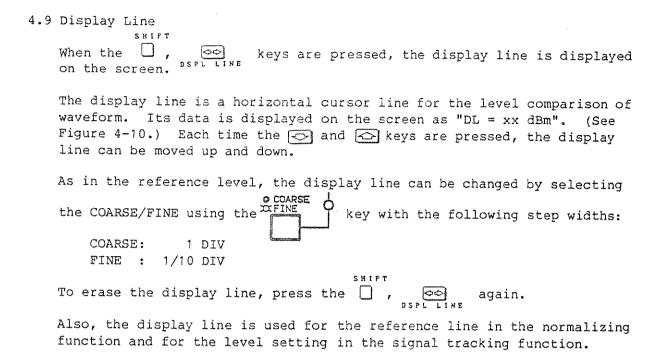
Also, when the equipment is initialized the 10 dB attenuator is always set for the protection of the first mixer.



Input attenuator value

Figure 4-9 Input Attenuator Displaying Position

CAUTION —
The attenuator can be set to 0 dB by pressing the



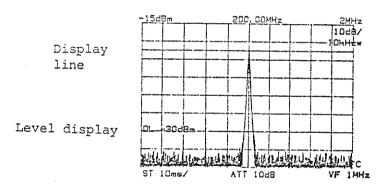


Figure 4-10 Display Line

4.10 Marker Function

(1) Display of Marker

MARKER

When the \bigcup key is pressed, the \Diamond shaped marker appears in the center of the frequency axis or a previously set position. In addition, the frequency and level of the marker are displayed on the upper left of the screen. The marker can be moved freely on the trace using the data knob.

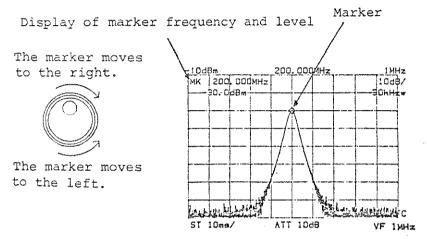


Figure 4-11 Operation of Marker

CTR FREQ

(2) Erasing of Marker

OFF

When the \(\sum \) key is pressed, the display of the marker and marker data is erased.

MARKER

When the key is pressed once more, the marker appears again on the frequency axis where it had disappeared.

(3) PEAK Search

PEAN

When the \(\begin{align*} \) key is pressed, the marker moves to the peak of the waveform with the highest level on the trace (Figure 4-12). This is a convenient function for setting the marker to the measuring signal.

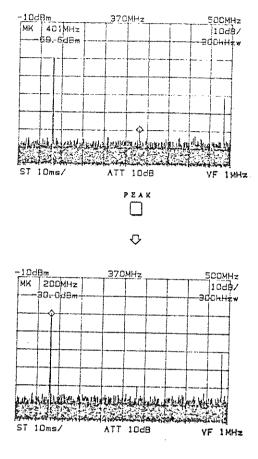


Figure 4-12 PEAK Search

(4) MarKeR → Center Frequency

When the \Box key is pressed, the marker and spectrum on which the marker is present move to the center of the screen to coincide with the center frequency. (Figure 4-13)

The spectrum can also be moved to the center of the screen by setting the center frequency using the known data. When this key is used, the spectrum can be moved to the center very quickly.

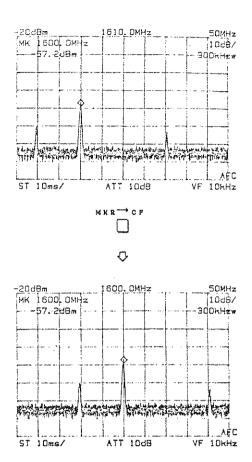


Figure 4-13 MarKeR → Center Frequency

(5) SIGnal TRack SHIFT When the and keys are pressed, the signal tracking function operates.

When this function is used, the frequency with the highest peak on the screen is automatically set as the center frequency each time the sweeping is done and when adopted makes it possible to always seize the signal in the center of screen, even if the signal drifts.

The signal tracking function of this equipment merely performs the PEAK searching on the screen and repeats it for each sweeping as MKR CF. However, the PEAK searching level can be selected by the display line. It is therefore possible to track only the signal which is higher in level than the display line.

By this, the center frequency never flies off due to noise without tracking the signal, even if the input signal is missed temporarily.
Press the \bigcap and $\bigoplus_{\text{DSPL LINE}}$ keys, and the display line is displayed on the screen.
Then, move the display line using the 10 dVB/DIV and 2 dB/DIV keys to determine the level for PEAK searching.
When the \bigcup and \sum_{TRK} keys are pressed, the signal above the value determined by the display line is tracked. (See Figure 4-14.)
Even if the display line is erased, the signal tracking is still carried out with the value determined earlier. To clear the signal tracking function, press the and keys over again, or
press the MARKER OFF SIG TRK MARKER OFF SIG TRK
SHIFT DSPLLING DSPLLING SIGNAL TRACK SHIFT SIG TRK DL 650Bm SIGNAL TRACK Figure 4-14 SIGnal TRACK
(6) MARKER PAUSE
After making the marker display on the screen, press the and warker and keys, or and keys in succession, then the MARKER PAUSE function operates.
This function stops the sweeping temporarily at the position of the marker. Although the stop time is 1 sec at first under the MARKER PAUSE status, it can be changed from 1 sec, in steps of 0.5 sec. It can be set in steps of 0.5 sec between 0 and 10.0 sec. (See Figure 4-15).
To clear this MARKER PAUSE function, set the stop time to 0 sec by HARKER TIME/DIV repeatedly pressing the and keys, or press the

key.

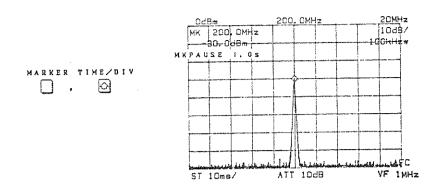


Figure 4-15 MARKER PAUSE

(7) Measurement of NOISE/Hz

After displaying the marker on the screen, press the and AUTO NOTSE/Hz keys in succession, then the NOISE/Hz function operates.

This function can measure the rms of the noise level which is normalized by the noise voltage band width of 1 Hz at the marker position.

The display detection mode at this time is automatically set to the SAMPLE DET. (See Figure 4-16.)

To clear the NOISE/Hz function, press the and wolfer are keys again.

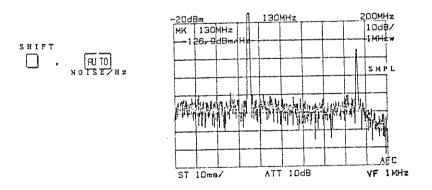


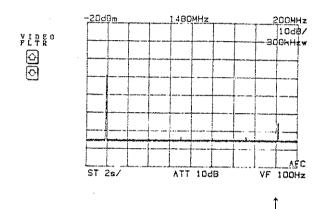
Figure 4-16 Setting of NOISE/Hz

4.11 Video Filter Band Width (VIDEO FiLTER)

Each time the \bigcirc key is pressed, the video filter band width can be changed with seven steps of 1 MHz \longrightarrow 300 kHz \longrightarrow 100 kHz \longrightarrow 10 kHz \longrightarrow 1 kHz \longrightarrow 100 Hz \longrightarrow 10 Hz.

Also, the video filter band width is interlocked with the sweep time to be set automatically to the optimum sweep time.

When the video filter band width is made smaller step by step, the signal which is buried in noise can be searched for, but it takes a long sweep time.



Video filter band width

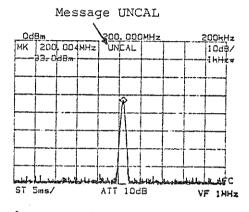
Figure 4-17 VIDEO FiLTeR

4.12 Setting of Sweep Time

Since the equipment is set to AUTO at initialization, the sweep time is automatically set to a range which does not cause a level error for the frequency span, resolution band width, and VIDEO FiLTER, etc.

TIME/DIV

When the key is pressed, the automatic setting is cleared and the sweep time can be set to a range from 5 ms/DIV to 100 s/DIV in steps of 1-2-5. The message "UNCAL" is displayed in the center of the screen when it is set in a manner to cause an error in the level display because of too rapid sweeping. Change the measuring condition, by making the sweep time longer for instance. (See Figure 4-18.)



\$\text{Sweep time}\$

Figure 4-18 Sweep Time

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4.13 Selection of Sweep Mode/Trigger Mode

4.1	3	Selection	of	Sweep	Mode/Trigger	Mode
				~ ~ ~ ~	TIOMO TITUMET	LIUCIE

Each time select in correspond	TRICGER the SINGLE O
FREE RUN:	This is a continuous sweep mode to automatically repeat the sweeping internally.
Line :	This is a mode to repeat the sweeping in synchronism with the AC power supply frequency.
VIDEO :	
SINGLE :	This is a single sweep mode and controlled by the \(\bar{\bar}\) key.
	START /RESET
	When the key is pressed in the single sweep mode, the
	sweeping is executed once.
	When the 'RESET' key is pressed in the middle of sweeping, the
	sweeping is reset for the next single sweeping. This function
	is used to retry sweeping during screen rewriting when the
	sweep time is long.

4.14 Display Detection Mode

This is the mode to specify which amplitude value should be converted from analog to digital when the amplitude data within a certain time during the sweeping is converted from analog to digital.

swee	ping is converted from analog to digital.
	display detection mode affects the display of noise or that of lsive signals.
	SAMPLE DETection
Ţ	When the and sample keys are pressed, the SAMPLE DET mode is
	selected and "SMPL" is displayed in the middle right of the screen. (See Figure 4-19.)
	This mode displays the result of sweeping at moments set at each point of the frequency axis.
	The SAMPLE DET mode is set automatically for measurement of the ${ m NOISE/Hz}$.
(2)	POSi PeaK DETection
-	- R4131D/DN
	When the and pos pk keys are pressed, the system goes into the
	POS PK DET mode.
	This mode displays the maximum value during the period set at each point of the frequency axis.
	Since this POS PK DET mode soundly seizes the spectrum peak, it is effective for the level measurement of a fine spectrum. (See Figure 4-20.)
-	- R4131C/CN
	R4131C/CN is set to the POS PK DET mode when it is initialized.

- (3) NORMAL DETection (POSI/NEGA DET)
 - R4131D/DN

When the $\bigcap_{p \in T}$ and \bigcap_{NORMAL} keys are pressed, the system enters the POSI/NEGA PK DET mode.

This mode displays the maximum value or minimum value of the periods set at each point of the frequency axis. (See Figure 4-21.)

R4131D/DN is set to the NORMAL (POSI/NEGA) DET mode when it is initialized.

- R4131C/CN

When the SHIPT and NORMAL keys are pressed, the system enters the POSI PK DET mode.

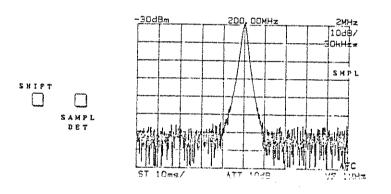


Figure 4-19 SAMPLE DET (R4131)

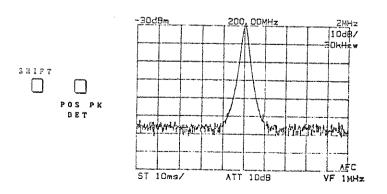


Figure 4-20 POSI PK DET (R4131)

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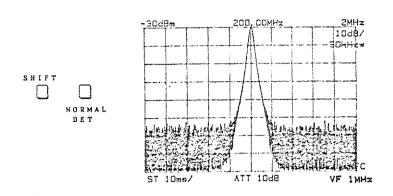
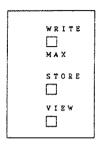


Figure 4-21 NORMAL DET (R4131B/BN/D/DN)

4.15 Selection of Trace Mode

The trace memory of this equipment provides two memories. One is the WRITE memory which rewrites the data for each sweeping the other is the VIEW memory which stores the waveform for any screen of the WRITE memory. The waveform of the WRITE memory or VIEW memory can be called, or both can be displayed on the screen to make a two-screen display.



(1) WRITE

When the key is pressed, the memory contents are rewritten at each sweeping.

The waveform of the WRITE mode is rewritten for each sweeping. The trace mode at initialization is set to this WRITE mode.

(2) STORE

STORE

When the \(\begin{aligned} \text{key} is pressed, the waveform data written in the WRITE mode at that time is held in the memory. The screen displays the waveform data held in the memory and then holds still. In other words, the system enters the VIEW mode and the leftward LED of the \(\begin{aligned} \text{VIEW} \\ \end{aligned} \)

key lights.

(3) VIEW

IEW

The key is used to call the waveform stored in the WRITE memory in the WRITE mode. Since the stored waveform data keeps its contents until new waveform data is stored again in the WRITE mode, this function is convenient for the comparative survey between the WRITE waveform after a change in setting conditions and the stored waveform data (the VIEW data).

(4) WRITE and VIEW (2-screen display)

When the display data, which is rewritten each sweeping by means of the key, is stored and the key is pressed again, both write leftward LEDs of the and keys light and the stored waveform data and the sweep data in the WRITE mode are displayed in two screens. To return the two screens to a single screen, erase the unnecessary screen using the write or view key.

The following describes how to use this function taking the comparative measurement of the secondary harmonic level as an example.

Operating procedure

- 1) Input the signal of CALibration OUTput, 200 MHz and -30 dBm, of this equipment.
- 2 Set as follows:
 Center frequency 200 MHz
 Reference level -30 dBm
 Frequency span 10 MHz
 In addition, set the POS PK DET to make it easier to compare two screens.
- 3 Set the spectrum of the measured signal to the center of the screen (Figure 4-22).

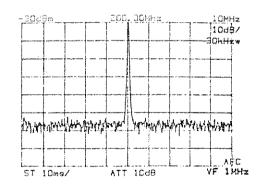


Figure 4-22 Setting the Measured Signal to the Center Frequency

- Press the key.
 Then, the trace mode becomes VIEW. The sweeping stops, the last sweep waveform is displayed, and the screen stands still. This data is stored in the internal memory.
- Then, a new WRITE waveform data is displayed together with the waveform of the memory (Figure 4-23).

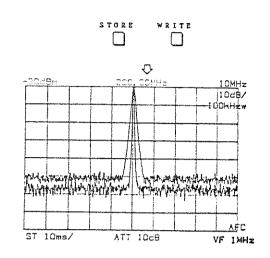


Figure 4-23 Two-screen Display with a New WRITE Waveform

6 Set the center frequency to 400 MHz and make the secondary harmonic wave move to the center of the screen.

Then, the measured value can be read from the difference in display between the two screens. (Figure 4-24)

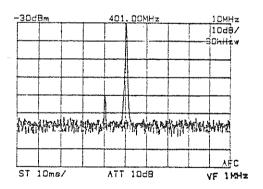


Figure 4-24 Two-Screen Display of Secondary Harmonic wave and STORE Waveform

4.15 Selection of Trace Mode

To erase the VIEW waveform while it is kept held in the memory to observe the secondary harmonic wave (WRITE waveform) further, press the key, Then, the screen becomes a single screen display of the WRITE waveform.

To display the memory waveform only, press the key. That is, press the key on the erased side.

(5) MAX HOLD

When the \Box and \Box keys are pressed, the stored data is rewritten and displayed on the screen, at each sweeping, any data that exceeds the former level at each point on the frequency axis is updated.

Consequently, the screen displays the maximum value up to then, for each point. (Figure 4-25)

In Figure 4-25, it can be seen that the signal is drifting in a range of approx. 4 MHz by putting it on MAX HOLD.

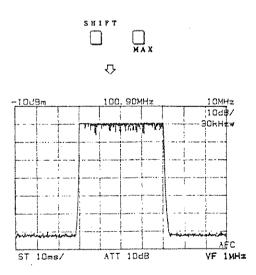
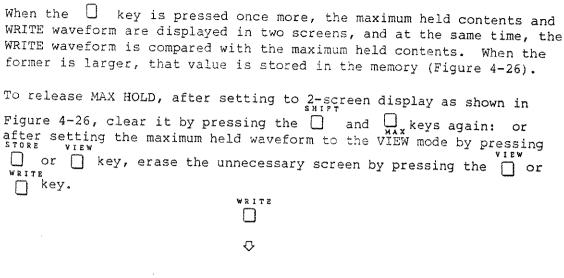


Figure 4-25 MAX HOLD

WRITE



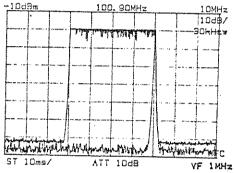


Figure 4-26 Two-screen Display of The Maximum Hold Contents and WRITE Waveform

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4.16 Setting Conditions and SAVE/RECALL of Displayed Waveform

4.16 Setting Conditions and SAVE/RECALL of Displayed Waveform

This equipment can save three states of the displayed waveform and it is setting conditions as shown table 4-2 in the non-volatile memory.

This function is convenient, because the setting conditions and displayed waveform can be recalled when the system is set up again since they are saved in the memory even if the power is turned OFF. It is also possible by using this function to compare waveforms and to block them out all together since the displayed waveforms can be recalled.

Table 4-2 SAVE/RECALL Enabled Panel Setting

Center frequency
Frequency span
Interlocking function (AUTO)
Resolution band width
Reference level
Reference level step width (COARSE/FINE)
INPUT attenuator
Video filter band width
Sweep time

When the setting conditions and displayed waveforms saved in the memory are recalled, the setting conditions are set in the WRITE screen at first and then the saved waveforms are recalled on the VIEW screen.

It is possible by pressing the \square key to see the waveforms which were saved in the memory (Figure 4-27).

(Non-volatile memory)

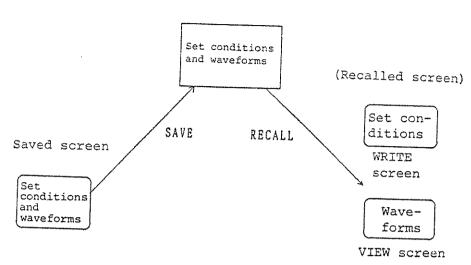


Figure 4-27 SAVE/RECALL Waveform Memory

Table 4-3 shows the relations with the screen stored in the memory in each trace mode.

Table 4-3 Screen Stored in Each Trace Mode

Trace mode	Screen stored
WRITE DISPLAY only VIEW display only WRITE/VIEW display MAX HOLD only WRITE/MAX HOLD display	Stores the WRITE screen. Stores the WRITE screen. Stores the WRITE screen. Stores the MAX HOLD screen. Stores the WRITE screen.

(1) SAVE

When the \Box and \Box keys are pressed, the system is enters the SAVE mode and the screen becomes as shown in Figure 4-28.

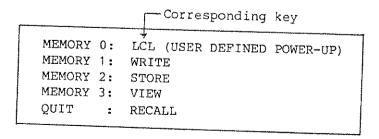


Figure 4-28 SAVE Screen

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4.16 Setting Conditions and SAVE/RECALL of Displayed Waveform

	Select the MEMORY 1, MEMORY 2, or MEMORY 3 using [], [], or [] key and select the memory to store.
	To quit the SAVE mode halfway, press the key.
	The MEMORY 0 is described in Section 4.17 Automatic Setting at Power ON.
(2)	RECALL
	RECALL
	When the key is pressed, the system enters the RECALL mode and
	the screen becomes as shown Figure 4-29.
	Corresponding key
	MEMORY 0: LCL (USER DEFINED POWER-UP)
	MEMORY 1: WRITE
	MEMORY 2: STORE
	MEMORY 3: VIEW
	QUIT : RECALL
	Figure 4-29 RECALL Screen
	WRITE STORE
	Select the MEMORY 1, MEMORY 2, or MEMORY 3 by using the \square , \square , or
	key to select the memory to call.
	To quit the RECALL mode halfway, press the \(\sum_{\text{key}}\) key.
	To quit the RECALL mode halfway, press the \(\subseteq \text{ key.} \)

SPECTRUM ANALYZER INSTRUCTION MANUAL

4.17 Automatic Setting at Power ON	V
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4.1	7	Automatic	Setting	at	Power	ON
-----	---	-----------	---------	----	-------	----

This is a function to call the setting stored in the non-volatile memory each time the power is turned ON. The setting of the equipment selected by yourself can always be called at power ON.

To store the setting to appear at power ON, press the \Box and \Box keys to put the system into the SAVE mode.

Then, the screen becomes as shown in Figure 4-28.

Press the MEMORY 0 and \square keys, then the set conditions are stored in the memory.

R4131 SERIES SPECTRUM ANALYZER INSTRUCTION MANUAL

4.18 Electric Field Intensity
Measurement (dBu/m)

4.18 Electric Field Intensity Measurement (dBμ/m)

The spectrum analyzer which can observe a wide frequency band at a time can also be used as a field intensity measuring instrument. When an antenna manufactured by ADVANTEST is used, this analyzer displays the level data by correcting the antenna factor, making it possible to read directly the field intensity through this analyzer. However, this correction value is effective only when the attached 5D2W cable, 10 m. is used. When using any other cable an error results.

Operating procedure

- (1) Connect the antenna to the input terminal (50 Ω) of this equipment. When the impedance of the antenna is not 50 Ω , be sure to match the impedance using a matching circuit.
- 2 Set the center frequency and frequency span, etc., to facilitate the observation.

3 Press the key and select the level unit to match the antenna as follows:

For TR1722 half-wave dipole antenna: dB μ /m (A) For TR1711 log helical antenna : dB μ /m (B) For TR17203 active antenna : dB μ /m (C) For TR17204 log helical antenna : dB μ /m (D)

PEAK

4 Press the key and set it to the peak of the spectrum to measure the marker.

The relationship between the marker point display level, that is, the input end voltage ex $(dB\mu V)$ of this equipment, and the actual field intensity Ex $(dB\mu V/m)$ is as shown below:

Ex = ex + K Where, K: antenna factor (dB)

When the above antenna is used, this antenna factor K is automatically corrected and the marker display indicates the field intensity.

When any antenna other than those mentioned above is used, correct the value referring to the following "Correction Coefficient of Antenna":

SPECTRUM ANALYZER INSTRUCTION MANUAL

4.18 Electric Field Intensity
Measurement (dBu/m)

- Correction Factor of Antenna

$$Ex = ex + K = (ex + 6) + La - He + Ba$$

Where,

EX : Field intensity (dB μ V/m)

ex : Input terminal voltage (dBµV)
K : Antenna correction factor (dB)
He (dB): Effective length of antenna

La (dB): Cable loss Ba (dB): Balun loss

The factor K of the half-wave dipole antenna is obtained according to the following equation:

$$K = 20 \text{ Log } \frac{\pi}{300} \text{ F} + 6 + \text{La} + \text{Ba} \qquad \text{F (MHz): Receiving frequency}$$
$$= -33.6 + 20 \text{ Log F} + \text{La} + \text{Ba}$$

For the broad band width logarithm frequency type antenna, deduct the antenna gain (half-wave dipole antenna ratio) from the obtained value.

Figure 4--30 shows the relationship between the frequency and calibration factor of TR1722 half-wave dipole antenna (including the cable loss).

4.19 QP Value Measurement (Quasi-peak Value Measurement)

4.19 OP Value Measurement (Quasi-peak Value Measurement)

The QP value measurement is for measuring the pulse characteristic noise. Various constants in this measurement are defined values in the CISPR Standards as shown in Table 4-4.

Table 4-4 CISPR Standards for QP Value Measurement Basic Characteristic

Measuring band				6 dB band width		Charging time constant		Discharging time constant	Mechanical time constant		
A	10 k	Ηz	to	150	kHz	20	Hz	45 r	ns	500 ms	160 ms
В	150 k	Hz	to	30	MHz	9	kHz	1 г	ns	160 ms	160 ms
С	30 M	Ηz	to	300	MHz	120	kHz	1 r	ns	550 ms	100 ms
D	300 M	Hz	to	1	GHz	110	kHz	1 r	ns	550 ms	100 ms

Note: This equipment has no A-range (10 kHz to 150 kHz, and 200 Hz band width).

Operating procedure

- ① Set the center frequency and frequency span to be measured. Since the QP band width is automatically set as the center frequency is set, select the frequency span in the band to be measured. For B-band for instance, the center frequency and span are selected as 25 MHz and 5 MHz, respectively.
- While observing the waveform, press the and keys and increase or decrease the input attenuator with in steps of 10 dB to check that the waveform level does not change. If changed, it indicates that the input stage of this equipment is saturated, so increase the attenuator value or add B.P.F (Band Pass Filter) to its input.
- When the level can be checked not to change, change the reference level so that the output peak level meets the reference level.
- 4 Press the and Exercise Reys.

The system enters the QP measurement mode under this status and the screen becomes 5 dB/DIV and eight scales.

4.19 QP Value Measurement (Quasi-peak Value Measurement)

5	Since a large time constant is entered when the QP value is measured as shown in Table 4-4, make the sweep time long enough. As a yardstick in this setting, set 1 sec per 10 kHz in the measuring band B (150 kHz to 30 MHz) and 1 sec per 10 kHz in the measuring bands C and D (30 MHz to 1 GHz).
6	Press the key to output the marker.
	Then, the QP value of the input terminal is displayed in terms of the marker frequency.
7	When an antenna manufactured by ADVANTEST is used, press the key and set the level unit to the antenna and select the unit as follows:
	TR1722 half-wave dipole antenna: $dB\mu/m$ (A) TR1711 log helical antenna : $dB\mu/m$ (B) TR17203 active antenna : $dB\mu/m$ (C) TR17204 log helical antenna : $dB\mu/m$ (D)
	Then, the antenna factor is automatically corrected, the level unit at the marker point becomes $dB\mu/m$, and the QP value is displayed directly on the screen.
	This correction is made only when the attached 5D2W antenna, 10 m, is used. When any other antenna is used, obtain the correction factor referring to the electric field intensity measurement in Section 4.18

and calculate the QP value.

8 Press either one of the

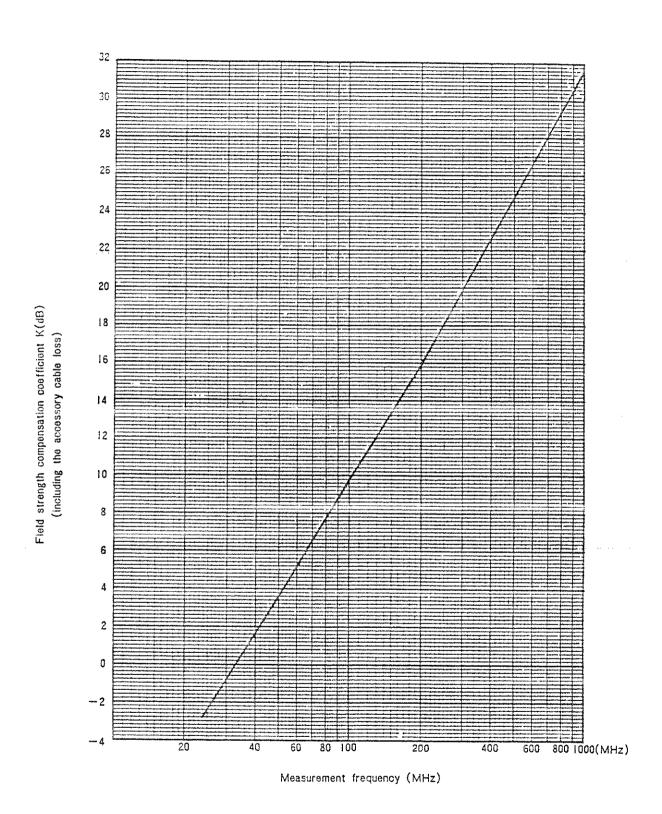


Figure 4-30 Relationship between Frequency and Calibration Factor in the half-wave dipole antenna

4.20 Normalize

The normalizing function is used to correct the frequency characteristic of this equipment itself and measuring systems including this equipment and to perform a relative comparison of displayed waveforms on the tube surface.

The following is the operating procedure for the measurement of the insertion loss of high frequency cables using the TR4153A/B tracking generator as an example.

Operating procedure

① Connect this equipment to TR4153A/B through the measuring system excluding the cable to be measured (Figure 4-31).

(The frequency characteristic in this measuring system includes the insertion loss of the connected cable and the frequency characteristic of this equipment. The cable insertion loss of the measured device is measured on the basis of this characteristic.)

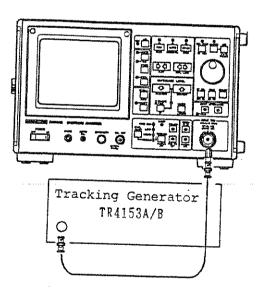


Figure 4-31 Direct Connection between Tracking Generator and System

2 TRACE : Set to WRITE (Initialization)

dB/DIV.: Set to 2 dB/DIV Span : Set to 2 GHz

3 To change the reference level and to widen the dynamic range on the lower side of the tube surface for measurement of the cable loss, move the through waveform to the upper side of the tube surface as shown in Figure 4-32.

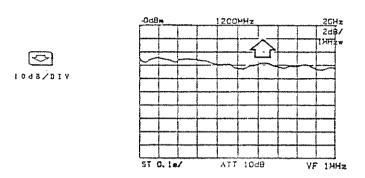


Figure 4-32 Moving the through Waveform

4 Then, the Display Line is Displayed on the Screen.

Move the display line close to the through waveform to make it the reference line of the normalizing (Figure 4-33).

The display line can be moved using the key.

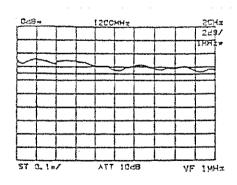


Figure 4-33 Moving the Display Line

(5) Normalize

When the and $\mathbb{R}^{\mathbb{R}}$ keys are pressed, the frequency characteristic of the measuring system is corrected and "NORM" is displayed on the tube surface and the through waveform coincides with the display line (Figure 4-34).

When the \bigcap and $\bigcap_{N \in \mathbb{N}} \mathbb{R}^{\mathbb{N}}$ keys are pressed directly without making the display line display, the level in the center of the tube surface is normalized as the reference line.

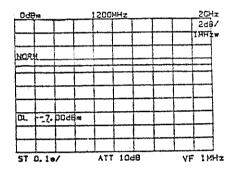


Figure 4-34 Normalize

6 Observation of insertion loss of a cable to be measured

Connect the measured cable to this equipment (Figure 4-35).

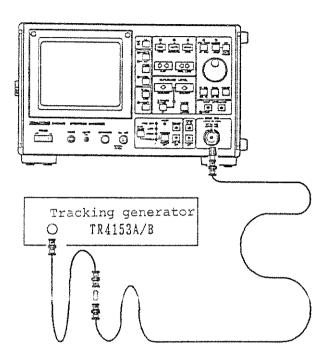


Figure 4-35 Connection of Measured Cable

7 Then, the measured waveform is displayed apart from the display line according to the cable loss (Figure 4-36).

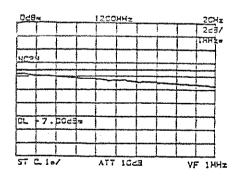


Figure 4-36 Cable Loss Characteristic

8 When the marker is displayed, the relative value between the marker point on the measured waveform and display line can be read directly in the marker level (Figure 4-37). To clear the NORMALIZE mode, press the and keys again.

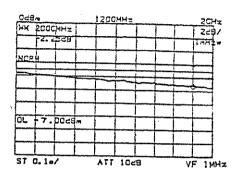


Figure 4-37 Reading the Characteristic of Waveform from the Marker Display

4.21 Occupied Frequency Band Width (OBW) Measurement (only for R4131D)

This function is mounted on R4131D only. This function obtains the occupied frequency band width from the data on the screen measured with this equipment. This operation is made as follows:

There are 701 points of data for the frequency axis on the screen of this equipment. Where one of the voltage is taken as Vn, the total power P on the screen can be obtained according to the following equation:

701
$$Vn^2$$

 $P = \sum_{n=1}^{\infty}$ (R: Input impedance of this equipment)

If X is taken as the point at which the sum of the power levels being displayed in sequence from the left end of the screen becomes 0.5% of P, the following equation can be established:

$$0.005 P = \sum_{n=1}^{X} \frac{\text{Vn}^2}{\text{R}}$$

If X is taken as the point at which the sum of the power levels being displayed in sequence from the left end of th screen becomes 99.5% of P, the following equation can be established:

$$0.995 P = \sum_{n=1}^{Y} \frac{V_n^2}{R}$$

Obtain X and Y from the above three equations and obtain the occupied frequency band width (OBW) from the frequency span SPAN according to the following equation:

$$OBW = \frac{fSPAN (Y-X)}{701}$$

The following is the operating procedure of the OBW display.

Operating procedure

- 1) Make the spectrum to be measured display in the center of the screen and set the screen ordinates axis scale to 10 dB/DIV.
- ② When the $\bigcap_{0 \text{ BW}}^{\text{SHIFT}}$ and $\bigcap_{0 \text{ BW}}$ keys are pressed, the function menu will then be displayed.

#OBW

3dB DOWN

3dB DOWN LOOP

NEXT PEAK

QUIT: OFF

Select a function after moving the # mark using the REFERENCE LEVEL

 \bigcirc and \bigcirc keys.

3 Press the key to execute the function.

Then, the operation of the OBW starts; when the operation ends, two markers appear as Y-point and X-point as mentioned above, and the OBW is displayed on the upper left of the screen (Figure 4-38).

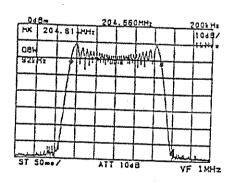


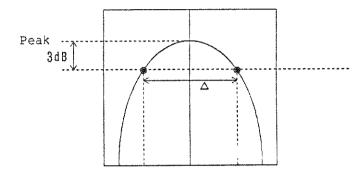
Figure 4-38 Example of OBW Measurement

4 When the MKR OFF switch is pressed, the display for the occupied frequency band width is erased and R4131D returns to the normal measuring mode.

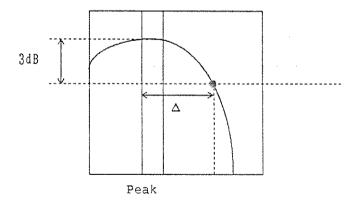
When the IF band width is set narrower when measuring the OBW, the measurement can be done with less error. When the MAX mode is used in combination with this, it is also possible to measure the maximum value of the OBW.

- 4.22 3dB DOWN, 3dB DOWN LOOP, NEXT PEAK Function (Only for R4131D)
 - (1) 3 dB DOWN
 - (1) If the marker is off

The spread in frequency between points at which the level has decreased by 3 dB from the peak will be calculated. If the decrease of 3 dB occurs at both a point on the displayed waveform that is lower than that of the peak level in frequency and at a point higher than that of the peak level in frequency, then the differences in frequency as well as in level between those two points will be displayed.



If the decrease of 3 dB occurs only at one point, the differences in frequency as well as in level between that point and the peak point will be displayed.

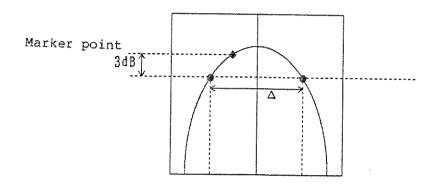


54434 CDDTDC

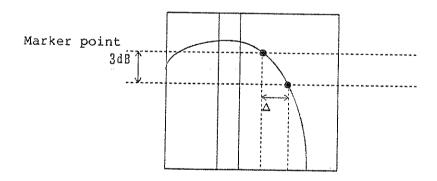
2 If the marker is already on

The spread in frequency between points at which the level has decreased by 3 dB from the level corresponding to the marker point will be calculated.

If the decrease of 3 dB occurs at both a point on the displayed waveform that is lower than the marker point in frequency and at a pont higher than the marker point in frequency, then the differences in frequency as well as in level between those two points will be displayed.



If the decrease of 3 dB occurs only at one of the two points mentioned above, the differences in frequency as well as in level between that point and the marker point will be displayed.



4.22 3dB DOWN, 3dB DOWN LOOP, NEXT PEAK Function

(2) 3 dB DOWN LOOP

The 3 dB DOWN LOOP function is valid only while the TRACE mode remaines set for WRITE. This function cannot be used for MAX HOLD.

If this function is selected, the peak level of the waveform will be detected at the end of sweep. Following this, the point(s) on the waveform where the power level decreases by 3 dB from the peak will be detected. As with the 3 dB DOWN function described above, if the decrease of 3 dB occurs at two points (or at one point only), then the differences in frequency as well as in level between those two points (or between that point and the marker point) will be displayed. In this case, operation will be the same, irrespective of the on or off status of the marker.

(3) NEXT PEAK

(1) If the marker is off

The marker will be placed at a position that corresponds to the signal having the second largest level.

(2) If the marker is on

The marker will move to a position that corresponds to the signal having the next larger level to that of the current marker point.

(3) If the display line is on

A search operation will be performed only on the signal having a level larger than the display line.

(4) Operating procedure

When the $\bigcap_{0 \text{ BW}}^{\text{SHIFT}}$ and $\bigcap_{0 \text{ BW}}$ keys are pressed, the function menu will then be displayed.

#OBW
3dB DOWN
3dB DOWN LOOP
NEXT PEAK
QUIT: OFF

Select a function after moving the # mark using the REFERENCE LEVEL

and keys.

- 2) Press the \bigcap^{MARXER} key to execute the function.
- 3 Press the key to return to the usual measurement mode.

4.23 Plotter Output

The tube surface data can be plotted using the ADVANTEST manufactured plotter and HP Corp. manufactured 7440 or its equivalent.

Operating procedure

- ① Connect this equipment to the plotter through the GPIB connector.
- 2 Then, the screen to be plotted can be stored and kept standing still. It is also possible to sweep it with the single trigger to make it stand still.
- 3) When the and PLOT keys are pressed, the system is made into the PLOTTER mode and the PLOT function selecting screen is displayed on the tube surface (Figure 4-39).

For instance, the # mark moves to either side of ALL or WAVE ONLY each time the $\boxed{\mathbb{R}^{k}}$ key is pressed.

- 4 Move the # mark using the associated keys and select any function. The plot type is selected with the min key and the size is selected with the min key.
- \bigcirc To quit from the PLOTTER mode at this point, press the \bigcirc key.
- 6 When the DSPL LINE (EXECUTE) key is pressed, the plotting is started.
- When the (CANCEL) key is pressed, the plotting can be stopped even halfway.

The PLOT TYPE of each plotter is selected as shown in Table 4-5.

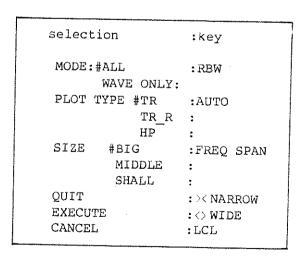


Figure 4-39 PLOT Function Selection Screen

Table 4-5 PLOT TYPE of Each Plotter

PLOT TYPE	Plotter name
HP	R9833, and HP Corp. manufactured 7470 or equivalent

Note: The plot type for R9833 is set to "HP" when they are delivered from the factory, since the HP-GL-1 (HP-GL) was then assumed to be used. When the FP-GL-2 (GP-GL) is used, set the plot type to "TR". The TR R is for the case where continuous roll paper is used.

When the connection to the plotter is no good or the power is not turned ON, "PLOTTER ERROR" is displayed in the center of the screen. Recheck the connection and setting and then reset with any key and then set the PLOT mode over again.

MEMO Ø

5. Applied Measuring Method

5. APPLIED MEASURING METHOD

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This chapter describes the overall operating method of this equipment through the measuring examples of AM wave and FM wave.



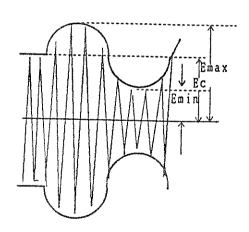
5.1 Measurement of Modulation Frequency and Index of AM Signal

The AM signal wave when expressed in the time axis becomes as shown in Figure 5-1 (a) and the modulation index m (%) can be obtained from the maximum value and minimum value of its waveform.

When expressed in the frequency axis, the AM signal wave becomes as shown in Figure 5-1 (b) and the modulation index m (%) can be obtained by measuring the frequency level of the carrier and that of the sideband.

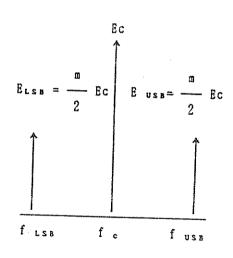
When the modulation frequency is low and its spectrum cannot be separated completely, the signal wave is observed in the ZERO SPAN mode. When the modulation frequency is high, the modulation index is generally obtained from the difference between the frequency of the upper sideband and that of the carrier in the FREQUENCY SPAN mode. When the modulation is small and the signal wave is difficult to see even though the modulation frequency is low, observe it in the FREQUENCY SPAN mode. The measurement precision rises when the signal wave is observed in the LINEAR mode when the modulation index is more than 10%, or in the LOG mode when the modulation index is less than 10%.

The following describes the measuring procedure for when the modulation frequency is low and when it is high.



$$m (%) = \frac{Emax - Ec}{Ec} \times 100$$
$$= \frac{Emax - Emin}{Emax + Emin} \times 100$$

(a) Time Axis Display of AM Signal Wave



$$m (%) = \frac{2E_{SB}}{EC} \times 100$$

(b) Frequency Axis Display of AM Signal Wave

Figure 5-1 AM Signal Wave

5.1 Measurement of Modulation Frequency and Index of AM Signal

5.1.1 Measurement of AM Wave When the Modulation Frequency Is Low and Modulation Index Is Large

Operating procedure

① Connect the AM transmitter output to the INPUT connector of this equipment by making it pass through the external attenuator when necessary (Figure 5-2).

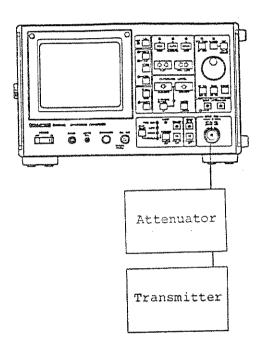
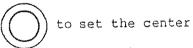


Figure 5-2 Set-up in Measurement of Modulation Wave

Set the center frequency to the frequency of the signal to be measured.
Data knob

Press the key and turn the frequency to 903 MHz (Figure 5-3).



Od 4K	904			903				٦,	1047 1049/
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	-		-			<u> </u>			
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4	1	opinion.		n jan	district the same of the same	PARTY	(Brown)	婚姻神	eprocessori

Figure 5-3 Setting the Center Frequency to the Frequency of the Measured Signal

. Oct 20/89

SPECTRUM ANALYZER INSTRUCTION MANUAL

5.1 Measurement of Modulation Frequency and Index of AM Signal

3	" the Call wild will will keys and sot the dans."
(4)	MARKER CHARLES the modulation frequency.
	Press the and set the marker to the peak of the Data knob
	measured signal with the ().
	(When the key is pressed, the marker is automatically set to the peak of the measured signal.)
	REFERENCE LEVEL
(5)	Press the lodB/DIV key and set the marker (the peak of the measured signal) to the reference level.
	PINE
6	Press the, key and set the ordinates axis scale
	to LINEAR (Figure 5-4).
	10.57mV 903.956MHz 2CCHHz
	1 21/4.577

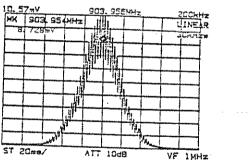


Figure 5-4 Setting the Ordinates Axis Scale to LINEAR

7	Press mode.	the	CHIFT	, [ERO	key	and	enter	the	system	into	the	ZERO SPAN	
			SHIFT	, S.A		key							SAMPLE mo	

5.1 Measurement of Modulation Frequency and Index of AM Signal

								Data knop				
9	Press	the	CTR FREG	key	and	turn	the		to	adjust	the	signal

level to make it the maximum.

- TRIGGER

 TRIGGER

 Ney and set the trigger mode to VIDEO.

 SAMPLE
 DET
- Press the POS PK ber observed easily. key and set the sweep time to a value that can be

Data knob

Press the and turn the to set the marker to the peak of the modulation signal.

Keep recording the time indication of the marker at this time (Figure 5-5).

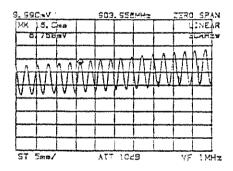


Figure 5-5 Reading the Time Display of Marker

Move the marker to the next peak and obtain the difference T(s) between the time indication of that marker and the time indication in step \bigcirc . In this example, it can be obtained as 18.6 - 16.0 = 2.6 (ms) (Figure 5-6).

Frequency fm of the modulation signal becomes as follows in this example:

$$fm = \frac{1}{T(s)}$$

 $fm = \frac{1}{2.6 \text{ (ms)}} = 384 \text{ (Hz)}$

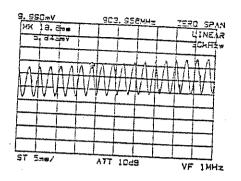


Figure 5-6 Reading the Difference from the Time Indication of the Adjacent Peak

(14) Read the marker level Emax (Figure 5-7).

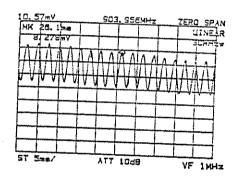


Figure 5-7 Reading the Emax

(15) Set the marker to the minimum value of the waveform and read the level Emin (Figure 5-8).

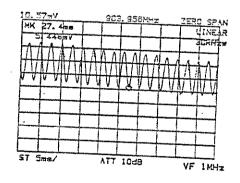


Figure 5-8 Reading the Emin

5.1 Measurement of Modulation Frequency and Index of AM Signal

(16) The modulation index m (%) becomes as follows in this example:

$$m = \frac{Emax - Emin}{Emax + Emin} \times 100 (%)$$

$$m = \frac{8.278 - 5.448}{8.278 + 5.448} \times 100 = \frac{2.830}{13.726} \times 100 = 20.6 (%)$$

5.1.2 Measurement of AM Wave When Modulation Frequency is High and Modulation Index is Small

Operating procedure

- 1 Connect the AM transmitter output to the INPUT connector of this equipment by making it pass through the external attenuator when necessary as shown in Figure 5-2.
- (2) Set the center frequency to the frequency of the carrier.



3 Set the frequency span to less than 10 times the modulation frequency.

(25 HM)		or $\Diamond \Diamond$
	PLOT	DSPL LIN

4) Set the marker to the peak of the carrier and keep recording that frequency (Figure 5-9).



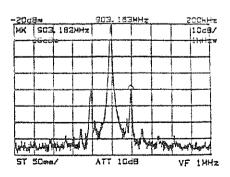


Figure 5-9 Measurement of AD Wave When Modulation Frequency is High and Modulation Index is Small

5.1 Measurement of Modulation Frequency and Index of AM Signal

5 Move the marker to the peak of the modulation signal spectrum.

Data knob

6 Compare the marker frequency and level at that time with the frequency kept recorded in step 4; then the modulation frequency and modulation index can be obtained from the difference between the frequency and level according to the following equation:

fm = Difference from the marker frequency indicated value

$$m = Log^{-1} \frac{(E_{SB} - E_C + 6)}{20} \times 100$$
 (%)

In the example of Figure 5-9, fm = 20 kHz and m = 2%.

Figure 5-10 shows the relationship between the value of (Sideband level ESB - carrier level EC) and modulation index m (%).

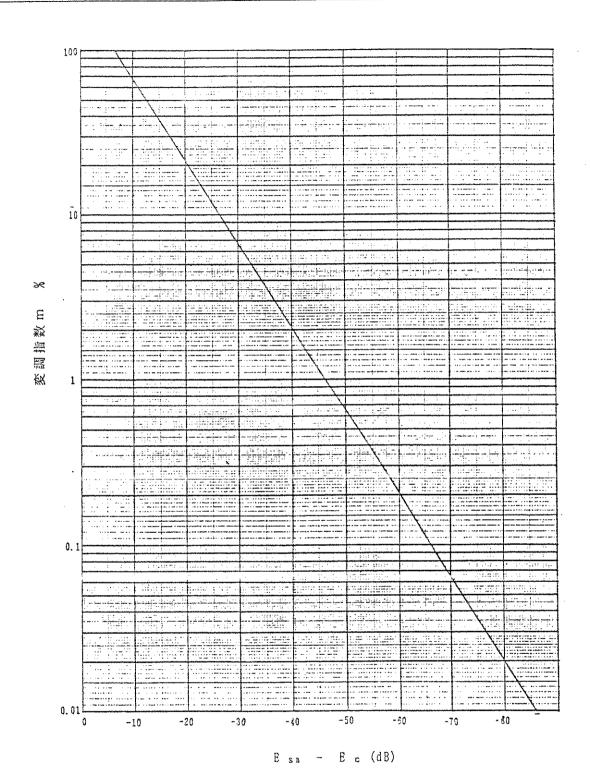


Figure 5-10 Relationship Between the Value of (Sideband Level $E_{\rm SB}$ - Carrier Level $E_{\rm C}$) and Modulation Index m (%)

5.2 Measurement of FM Wave

When observing the FM wave, it is possible to obtain modulation frequency fm, modulation index m, and peak deviation Δf peak. When the modulation frequency is low, set the ordinates axis to the ZERO SPAN, make it operate as a fixed tuning receiver, demodulate the frequency using the slope of the IF filter, and measure it on the time axis.

When the modulation frequency is high, measure it on the frequency axis and obtain the modulation frequency from the frequency of the sideband. When the modulation index m is small (when it is less than approx. 0.8), obtain it from the relationship between the carrier level and the first sideband level.

The following describes this measurement example in either case.

5.2.1 Measurement of FM Wave When Modulation Frequency Is Low

Operating procedure

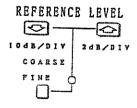
- 1 Connect the FM transmitter output to the INPUT connector of this equipment by making it pass through the external attenuator when necessary, as shown in Figure 5-2.
- ② Set the carrier of the signal so that it becomes the center frequency, and make it the span suitable for analyzing the spectrum.



3 Set the marker to the peak of the signal.



(4) Set the marker level to the reference level.



(5) Lower the reference level (Figure 5-11).

REFERENCE LEVEL

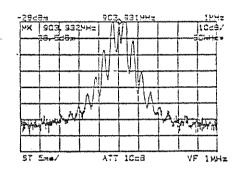
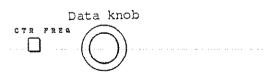


Figure 5-11 Measurement of FM Wave When Modulation Frequency Is Low

(6) Make the system into the ZERO SPAN mode.



(7) Change the center frequency so that the demodulation wave becomes the center of the screen.



Make the resolution band width to more than three times the modulation frequency so that the demodulation wave can be seen easily.



(9) Set the trigger mode to VIDEO.

TRIGGER

10 Select a sweep time for easily seeing the demodulation wave.

TIME/DIV
Or O

POS PK
DET

11) Put the marker on the peak of the demodulation wave and keep recording its time indication (Figure 5-13).



ΗK	22. 2	යා ය ක්වාන				WH-		Ì	SP/ 10ds
1	4 1	11	1	١٩٠			10.0		
		1111		111	11 11		11 11	1	
 	##	/	111	VV	444	Щ	<u>;; </u>	<u> </u>	Ш
	├-							-	-
	-								
	-								_
~ ~	Sme/				100				F 1M

Figure 5-12 Putting the Marker on the Peak of Demodulation Wave and Reading Its Time Indication

(2) Move the marker to the adjacent peak and read its time indication (Figure 5-13).



From the time interval T(s) of the peak of the demodulation wave, the modulation frequency (fm) can be obtained as follows:

$$fm = \frac{1}{T(s)}$$

Since T(s) = 2.1 (ms) in this example, the modulation frequency (fm) can be obtained as follows:

$$fm = \frac{1}{2.1 \text{ (ms)}} = 476 \text{ (Hz)}$$

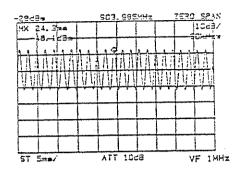


Figure 5-13 Obtaining the Time Interval T(s) of Demodulation Wave

5.2.2 Measurement of FM Wave for High Modulation Frequency

Operating procedure

- ① Connect the FM transmitter output to the INPUT connector of this equipment by making it pass through the external attenuator when necessary, as shown in Figure 5-2.
- (2) Set the carrier frequency to the center frequency.



Set the frequency span to a value lower than 10 times of the modulation frequency.

4 Put the marker on the peak of the carrier and keep recording the marker frequency at this time (Figure 5-14).



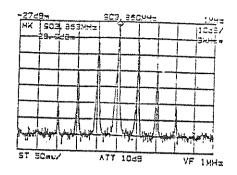


Figure 5-14 Measurement of FM Wave When Modulation Frequency Is High

(5) Move the marker to the adjacent peak and read the indication of the marker frequency (Figure 5-15).



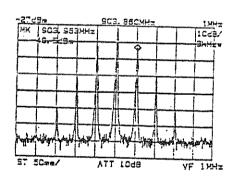


Figure 5-15 Reading the Modulation Frequency from the Marker Display

6 The difference from the frequency indication of the marker becomes the modulation frequency (fm).

For this example, the modulation frequency can be obtained as follows:

fm = 903.963 - 903.863 = 100 (kHz)

5.2.3 Measurement of Peak Deviation (Δf peak) of FM Wave

Operating procedure

- 1 Connect the FM transmitter output to the INPUT connector of this equipment by making it pass through the external attenuator when necessary, as shown in Figure 5-2.
- (2) Set the center frequency to the carrier frequency.



3 Set the frequency span to a value enabling easy measurement according to the peak deviation.

4 Set the resolution band width to a value including the principal sideband (more than five times the modulation frequency).

(5) Figure 5-16 shows a case where Δf_{peak} is small and Figure 5-17 shows a case where it is large. Measure the Δf_{peak} from the waveform.

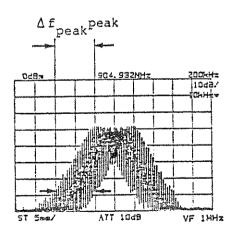


Figure 5-16 Waveform When Δf_{peak} Is Small

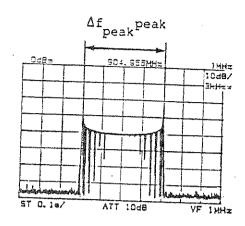


Figure 5-17 Waveform When Δf_{peak} Is Large

 $\Delta f_{\mbox{\scriptsize peak}}$ and modulation index m can be obtained from the following equation:

 $\Delta f_{peak} = \frac{1}{2} \times \Delta f_{peak}$ peak

$$m = \frac{\Delta f_{peak}}{fm}$$

For the two figures, the measurement is carried out as follows, respectively:

• Figure 5-16: When Δf_{peak} is small

fm = 2 kHz, and Δf_{peak} is read as approx. 40 kHz:

$$\Delta f_{peak} = \frac{1}{2} \times 40 \text{ (kHz)}$$

$$m = \frac{20 \text{ (kHz)}}{2 \text{ (kHz)}} = 10$$

• Figure 5-17: When Δf_{peak} is large

fm = 400 Hz, and Δf_{peak} is read as approx. 400 kHz:

$$\Delta f_{peak} = \frac{1}{2} \times 400 \text{ (kHz)}$$

$$m = \frac{200 \text{ (kHz)}}{400 \text{ (Hz)}} = 500$$

5.2.4 How to Obtain Modulation Index m when FM Modulation Index m Is Small

When the modulation index m of the FM wave is less than approx. 0.8, the following equation can be formed:

$$m = \frac{2E_{SB}}{E_{C}}$$

Where,

EsB: 1st sideband level

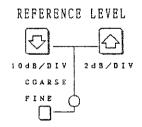
E_C: Carrier level

Operating procedure

- 1) Connect the FM transmitter output to the INPUT connector of this equipment by making it pass through the external attenuator when necessary, as shown in Figure 5-2.
- ② Set the center frequency and frequency span so that the carrier can be observed easily.



(3) Set the carrier level to the reference level as shown in Figure 5-18.



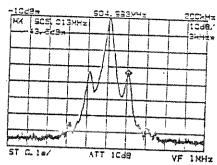


Figure 5-18 How to Obtain Modulation Index m When FM Modulation Index m is Small

4 Read and keep recording the carrier frequency $f_{\mathbb{C}}$ from the display of the center frequency and also the carrier level $E_{\mathbb{C}}$ from the display of the reference level.

In the case of this example, they become as follows:

$$f_C = 904.993$$
 MHz, $E_C = -10$ dBm

 $\stackrel{\textstyle (5)}{}$ Set the marker to the first sideband and read its frequency f_{SB} and level E_{SB} from the display of the marker.



For this example, they become as follows:

$$f_{SB} = 905.103 \text{ MHz}, E_{SB} = -43.6 \text{ dBm}$$

6 The FM modulation index m can be obtained from the following equation:

$$m = 2 \times \frac{E_{SB}}{E_{C}} = Log^{-1} \frac{E_{SB} - E_{C} + 6}{20}$$

For this example, it becomes as follows:

$$m = Log^{-1} \frac{-43.6 - (-10) + 6}{20} = Log^{-1} (-1.38) = 0.04$$

5.2 Measurement of FM Wave

7) The modulation frequency fm can be obtained from $fm = |f_{SB} - f_C|$.

For this example, it becomes as follows:

fm = 20 kHz

(8) The frequency deviation Δf_{peak} can be obtained from $\Delta f_{peak} = m \times fm$.

For this example, it becomes as follows:

 $\Delta f_{peak} = 0.04 \times 20 \text{ (kHz)} = 800 \text{ Hz}$

MEMO Ø

6. GPIB Connection and Programming

6. GPIB CONNECTION AND PROGRAMMING

This equipment features the measurement bus GPIB (General Purpose Interface Bus), which conforms the IEEE Standards 488-1978, as standard equipment to enable full remote control by an external controller.

6.1 Outline of GPIB

The GPIB is an interface system which can connect a measuring instrument to a controller and its peripheral equipment, etc. with a simple cable (bus line). Compared with conventional interfacing methods, it has excellent expandability, is easy to use, and is compatible with products of other companies electrically, mechanically, and functionally. This allows versatile configuration from a simple system to a high-level automatic measuring system with one bus cable.

In the GPIB system, it is first necessary to preset an "address" of separate component equipment connected to its bus line. These equipment can perform one or two of three roles -- controller, talker (speaking party), and listener (listening party).

During the system operation, only one talker can send data to the bus line and a multiple listeners can receive the data. The controller specifies the address of a talker and listener to transfer data from the talker to listener, or the controller itself (a talker in this case) sets measuring conditions, etc., of the listener.

For data transfer between equipment, the GPIB system uses eight data lines of bit parallel and byte serial types and also transmits data in both directions asynchronously. Being an asynchronous system, high speed devices and low speed ones can be connected to each other.

The data (messages) exchanged between devices consists of measuring data, measuring conditions (programs), and various commands. The system uses the ASCII code.

In addition to the above eight data lines, the GPIB provides three handshaking lines to control sending and receiving asynchronous data, and five control lines to control the flow of data on bus lines.

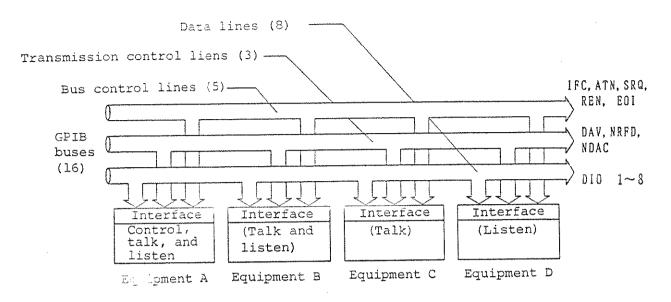


Figure 6-1 Outline of GPIB

• The following signals are used for handshaking lines:

DVA (Data Valid) : This is a signal to indicate that the data

is valid.

NRFD (Not Ready For Data): This is a signal to indicate that the data

is ready for receiving.

NDAC (Not Data Accepted) : This is a signal to indicate that the data

reception is completed.

The following signals are used for control lines:

ATN (Attention) : This is a signal used to distinguish that the signal on the data line is either

address or command, or some other data.

IFC (Interface Clear) : This is a signal to clear the interface.

EQI (End or Identify) : This is a signal used when the data transfer ends.

CIRCO.

SRQ (Service Request) : This is a signal used to request a service

from any equipment to the controller.

REN (Remote Enable) : This is a signal used when remote programmable equipment is controlled

remotely.

6.2 Standards

6.2.1 GPIB Specifications

Conformed standards : IEEE Standards 488-1978

Code used : ASCII code, or binary code for packed format

Logical level : Logical 0 "High" status More than +2.4 V

Logical 1 "Low" status Less than +0.4 V Signal line termination: 16 bus lines are terminated as shown below:

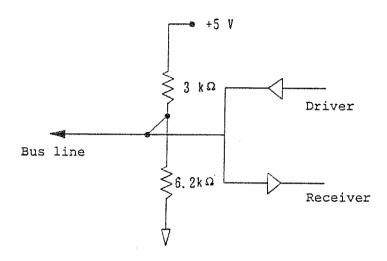


Figure 6-2 Signal Line Termination

Driver specification : Open collector type

Output voltage under the "Low" status

... 48 mA at +0.4 V or less

Output voltage under the "High" status

 \dots -5.2 mA at +2.4 V or more

Receiver specification: "Low" status at +0.6 V or less

"High" status at +2.0 V or more

Length of bus cable : The length of each cable should be less than

4 m and the total length of all bus cables (the number of equipment connected to buses \boldsymbol{x}

2) should not exceed 20 m.

Address specification: 31 types of TALK address/LISTEN addresses can

be set freely using the ADDRESS switch on the

rear panel.

After changing over to the ADDRESS switch, turn OFF the POWER SW once and then ON again.

Connector : 24-pin GPIB connector

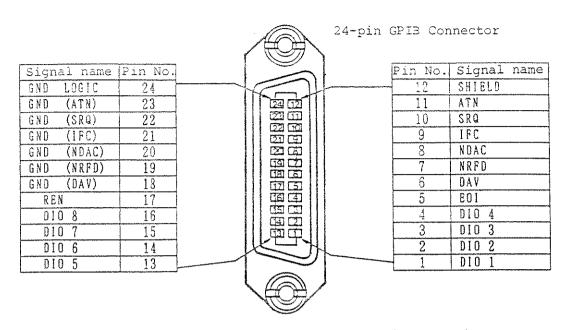


Figure 6-3 GPIB Connector Pins Assignment Diagram

6.2.2 Interface Function

Table 6-1 Interface Function

Cođe	Function and explanation
SH1	Source handshaking function
AH1	Acceptor handshaking function
Т6	Basic talker function, serial polling function, and talker releasing function by listener specification
L4	Basic listener function and listener releasing function by talker specification
SR1	Service requesting function
RL1	Remote function
PP0	No parallel function provided
DC1	Device clearing function provided
DT1	Device triggering function provided
C0	No controlling function provided. However, the controller function is enabled when the plotter is used.
E1	Open collector and bus driver used. However, E2 is used for EOI and DAV (three-state bus driver used).

,3 GPIB Handling Method

6.3.1 For Connection to Component Devices

Since the GPIB system is composed of multiple devices, prepare the entire system while paying attention to the following points especially.

- (1) Before connection, check the condition and operation of each device according to the operation manual for R4131, controller and other peripheral devices, etc.
- (2) Do not make any bus cable connected to each measuring instrument and controller, etc., unnecessarily long. The length of each cable should be less than 4 m and the total length of all bus cables (the number of devices connected to buses x 2) should not exceed 20 m. ADVANTEST provides standard bus cables as shown in Table 6-2.

Table 6-2 Standard Bus Cables (To Be Purchased Separately)

	·
Length	Name
0.5 m	408JE-1P5
1 m	408JE-101
2 m	408JE-102
4 m	408JE-104

- (3) Bus cable connectors are of a piggy back type. Male and female connectors are provided for one connector, which can be used one over the other. Do not pile up three or more connectors when connecting cables. Also, be sure to screw connectors tightly with setscrews.
- (4) Before turning ON the power of the devices connected to the bus lines, check their power supply conditions, grounding status, and setting conditions, too, when necessary. Be sure to set the power of each component unit to ON. If any of them is not set to ON, the overall operation cannot be guaranteed.

3.2 Setting of ADDRESS Switch

The rear panel of this equipment has a ADDRESS switch (Figure 6-4) used to set addresses on the GPIB. By setting bits 1 (the right end) to 5 to 0 or 1, addresses can be set from 0 to 30.

Set the ADDRESS switch before turning on the power.

The relationship between this ADDRESS switch and GPIB addresses is shown in Table 6-3.

GPIB	Bit	GPIB	Bit	GPIB	Bit
address	5 4 3 2 1	address	5 4 3 2 1	address	5 4 3 2 1
0 1 2 3 4 5 6 7 8 9	0 0 0 0 0 0 0 0 0 0 0 0 0 1 0 0 0 1 0 0 0 1 0 0 0 0 1 0 0 0 0 0 1 0 0 0 0 0 1 0 0 1 0 0 0 1 0 0 0 0 0 1 0 0 1 0 0 1 0 0 1 0 0 1 0 0 1 0 0 1 0 0 0 0 0 0 0 0 0 0 1 0 0 0 1 0 0 0 1 0 0 0 1 0 0 0 1 0 0 0 1 0	11 12 13 14 15 16 17 18 19 20	0 1 0 1 1 0 1 1 0 0 0 1 1 0 1 0 1 1 0 1 0 1 1 1 0 0 1 1 1 1	21 22 23 24 25 26 27 28 29 30	1 0 1 0 1 1 0 1 1 0 1 0 1 1 1 1 1 0 0 0 1 1 0 0 1 1 1 0 1 0

Table 5-3 Setting of ADDRESS Switch

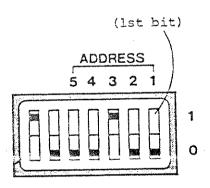


Figure 6-4 ADDRESS Switch

6.3.3 Programming

Programming for GPIB covers the sending of GPIB command codes and data to equipments to be connected, reading of data from devices, execution of bus commands, and I/O commands e.g., serial polling, etc. The arithmetic operation and others shall conform to the program generating procedure in the controller.

The format of GPIB commands to any equipments and I/O statements of data have the configuration as follows:

I/O Part Unit Address ; I/O Command, Code, and Data

Setting of Each Function

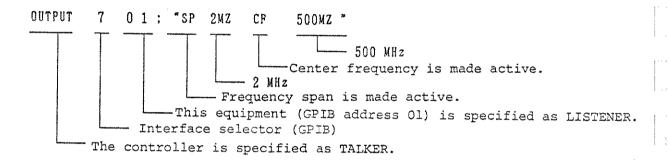
This equipment may be put under remote control for all functions using the GPIB controller.

This section describes the setting of each function of this equipment referring to program examples using a desk-top computer, HP Corporation's $\rm HP200/300$ series.

Program examples are all assumed to be set from their initial status.

Example 6-1: Setting the Center Frequency to 500 MHz and Frequency Span to 2 MHz

HP200, 300 Series



Then programmed and executed as above, this equipment is set to 500 MHz in senter frequency and 2 MHz in frequency span.

 ${\tt F}$, ${\tt SP}$, and ${\tt MZ}$, etc. in the program are all GPIB commands to control this quipment.

ince these commands correspond to keys of this equipment, the programming an be made in the order of pressing keys on the panel.

ee Section 6.9 for a list of GPIB codes.

6.4.1 Setting of Center Frequency

There are two methods available for the setting of center frequency using the GPIB.

One is to make the center frequency increase (or decrease) step by step using the data knob setting command, and, while reading its value sequentially, it is repeated until the frequency is set to the target value. The other method is to set the value of frequency directly.

(1) When the Center Frequency Is Set Using the Command for Setting the TUNING Knob

Example 6-2: Setting the Center Frequency to 1 GHz

HP200/300 Series

- 10 OUTPUT 701; "SP 1GZ"
- 20 OUTPUT 701; "OPCF"
- 30 ENTER 701;F
- 40 IF F=1E9 THEN 70
- 50 OUTPUT 701; "CD"
- 60 GOTO 30
- 70 IF F=1E9 THEN 100
- 80 OUTPUT 701; "CU"
- 90 GOTO 30
- 100 END

Line No.	Meaning
10	Sets the frequency span to 1 GHz.
20	Instructs this equipment to output the value of the center frequency. See the OP Command in 6.5.1.
30	Reads the value of the center frequency.
40	Branches to line No. 70 when the read data is smaller than or equal to 1×10^9 (Hz).
50	Sends the command to turn the data knob counterclockwise for 1 step of COARSE.
60	Returns to line No. 30.
70	Branches to line No. 100 when the read data is equal to 1 \times 109 (Hz).
80	Sends the command to turn the data knob clockwise for 1 step of COARSE.
90	Returns to line No. 30.
100	End of program

Note: Note that the set resolution of the center frequency becomes coarse and the center frequency cannot be set to the desired value when the frequency span is wide.

6.4 Setting of Each Function

(2) When the Value of Center Frequency is Set Directly Example 6-3: Setting the Center Frequency to 1 GHz Directly HP200/300 Series

10 OUTPUT 701: "CF1GZ"

20 END

Line No.	Meaning
	Sets the center frequency to 1 GHz. End of program

6.4.2 Setting of Frequency Span

There are two methods available for the setting of the frequency span using the GPIB. One is to make the frequency span wider or narrower in 1-2-5 steps using the command (NR and WD) corresponding to the key on the front panel. The other method is to set the value of the frequency span directly.

(1) When Using the Command Corresponding to the Key on Front Panel

Example 6-4: Setting the Frequency Span to 20 MHz

HP200/300 Series

- 10 OUTPUT 701; "OPSP"
- 20 ENTER 701;S
- 30 IF S<=20E6 THEN 60
- 40 OUTPUT 701; "NR"
- 50 GOTO 20
- 60 IF S=20E6 THEN 90
- 70 OUTPUT 701; "WD"
- 80 GOTO 20
- 90 END

Line No.	Meaning
10	Instructs this equipment to output the set value of frequency span. Sends the command SP of the SPAN key to light the LED on the key.
20	Reads the data (the value of the frequency span).
30	Branches to line No. 60 when the read data is smaller than or equal
	to 20 x 10^6 (Hz).
40	Sends the command for so of this equipment to make the frequency
	span narrower by 1 step.
50	Returns to line No. 20.
60	Branches to line No. 90 when the read data is equal to 20×10^6
	(Hg)
70	Sends the command for 🖘 of this equipment to widen the frequency
	span by 1 step.
80	Returns to line No. 20.
90	End of program

(2) When the Value of Frequency Span Is Set Directly

Example 6-5: Setting the Frequency Span to 20 MHz Directly HP200/300 Series

10 OUTPUT 701; "SP20MZ" 20 END

Line No.	Meaning
10	Sets the frequency span to 20 MHz.
20	End of program

When the frequency span is set directly, do it using the codes given in the table below.

Frequency Span Set	Value	Codes
--------------------	-------	-------

Cođe	SPAN	Code	SPAN	Code	SPAN
SP50KZ SP100KZ SP200KZ SP500KZ SP5MZ SP2MZ SP5MZ	50 kHz 100 kHz 200 kHz 500 kHz 1 MHz 2 MHz 5 MHz	SP1 0MZ SP20MZ SP50MZ SP100MZ SP200MZ SP500MZ	10 MHz 20 MHz 50 MHz 100 MHz 200 MHz 500 MHz	SPIGZ SP2GZ SP4GZ ZS	1 GHZ 2 GHZ 4 GHZ ZEROSPAN

6.4.3 Setting of Reference Level

There are two methods available for setting the reference level using the GPIB.

One is to set the reference level up and down using the command (LU, LD, or FC) corresponding to the key on the front panel to set it to the desired value. The other method is to set the value of the reference level directly.

Note that the set range of the reference level narrows according to the set value of the input attenuator.

6.4 Setting of Each Function

(1) When Using the Command Corresponding to the Key on Front Panel

Example 6-6: Setting the Reference Level to -30 dBm

HP200/300 Series

- 10 OUTPUT 701; "OM"
- 20 ENTER 701 USING "#,B";A1,A2,A3,A4,A5,A6,A7
- 30 IF A4=1 THEN 50
- 40 OUTPUT 701; "FC"
- 50 OUTPUT 701; "OPRL"
- 60 ENTER 701; L
- 70 IF L<=-30 THEN 100
- 80 OUTPUT 701; "LD"
- 90 GOTO 60
- 100 IF L=-30 THEN 130
- 110 OUTPUT 701; "LU"
- 120 GOTO 60
- 130 END

Line No.	Meaning
10	Instructs the equipment to output the mode string.
20	Reads the mode string.
30	Incorporates a numeric value which indicates the setting COARSE or FINE that the reference level setting switch sets to the numerical variable A4. (COARSE = 0, FINE = 1) Branches to line No. 50.
40	Sends the COARSE/FINE SELECTION key command.
50	Instructs this equipment to output the set value of the reference level.
60	Reads the data.
70	Branches to line No. 100 when the read data is less than or equal to -30 (dBm) .
80	Sends the command of the REFERENCE LEVEL DOWN key 🕏 to lower the reference level by 1 step.
90	Returns to line No. 60.
100	Branches to line No. 130 when the read data is equal to -30 (dBm).
110	Sends the command of the REFERENCE LEVEL UP key (to raise the reference level by 1 step.
120	Returns to line No. 60.
130	End of program

Note: See the mode string in 6.5.3.

6.4 Setting of Each Function

- (2) When the Value of the Reference Level Is Set Directly

 Example 6-7: Setting the Reference Level to -30 dBm Directly

 HP200/300 Series
 - 10 OUTPUT 701: "RL-30DM"
 - 20 END

	Line No.	Meaning	-
	10 20	Sets the reference level to -30 dBm. End of program	
_			

6.4.4 Setting of Marker

There are two methods available for setting the marker.

One is to increase or decrease the marker frequency step by step using the command for the data knob setting, and while reading its value sequentially, this is repeated until the marker is set to the desired value. The other method is to set the value of the marker frequency directly.

(1) When Using the Command Corresponding to the Data Knob

Example 6-8: Setting the Marker Frequency to 1 GHz

HP200/300 Series

- 10 OUTPUT 701; "M1"
- 20 OUTPUT 701; "OPMF"
- 30 ENTER 701;M
- 40 IF M<=1E9 THEN 70
- 50 OUTPUT 701; "FD"
- 60 GOTO 30
- 70 IF M=1E9 THEN 100
- 80 OUTPUT 701; "FU"
- 90 GOTO 30
- 100 END

6.4 Setting of Each Function

Line No.	Meaning
10	Displays the marker. Instructs this equipment to output the value of the marker frequency.
20	Instructs this equipment to output the variety
30	Reads the value of the marker frequency. Branches to line No. 70 when the read data is smaller than or equal
40	Branches to line No. 70 when the read data is small
50	to 1 x 10^9 (Hz). Sends the command to turn the data knob counterclockwise for 1 step of FINE.
60	7. 7. 7.
70	Returns to line No. 30. Branches to line No. 100 when the read data is equal to 1 \times 109
80 90 100	(Hz). Sends the command to turn the data knob clockwise for 1 step of FINE Returns to line No. 30. End of program

(2) When the Value of Marker Frequency Is Set Directly

Example 6-9: Setting the Marker Frequency to 1 GHz Directly

HP200/300 Series

- 10 OUTPUT 701; "MK1GZ"
- 20 END

	Line	Meaning			
ļ	No.		1		
	10	Sets the marker frequency to 1 GHz.	ŀ		
	20	End of program	j		

6.4.6 Setting of VIDEO FiLTeR Band Width

There are two methods available for setting the VIDEO FiLTER band width using the GPIB. One is to set it by making the VIDEO FiLTER band width narrower or wider step by step using the command (VU or VD) corresponding to the key on the front panel. The other method is to directly set the value of VIDEO FiLTER band width.

(1) When Using the Command Corresponding to the Key

Example 6-13: Setting the VIDEO FiLTeR band width to 100 ${\rm Hz}$

HP200/300 Series

- 10 OUTPUT 701; "OPVF"
- 20 ENTER 701; V
- 30 IF V<=1E2 THEN 60
- 40 OUTPUT 701; "VD"
- 50 GOTO 20
- 60 IF V=1E2 THEN 90
- 70 OUTPUT 701; "VU"
- 80 GOTO 20
- 90 END

Line No.	Meaning
10	Instructs this equipment to output the value of VIDEO FiLTER band width.
20	Reads the data.
30	Branches to line No. 60 when the read data is smaller than or equal to 1 \times 10 ² (Hz).
40	Sends the VIDEO FILTER DOWN key 🖸 command to lower the set value of VIDEO FilTeR band width by 1 step.
50	Returns to line No. 20.
60	Branches to line No. 90 when the read data is equal to 1 \times 10 ² (Hz).
70	Sends the VIDEO FILTER UP key ② command to raise the set value of VIDEO FiLTER band width by 1 step.
80	Returns to line No. 20.
90	End of program

- (2) When the Value of VIDEO FiLTER Band Width Is Set Directly Example 6-14: Setting VIDEO FiLTER band width to 100 Hz Directly HP200/300 Series
 - 10 OUTPUT 701; "VF 100HZ"
 - 20 END

Line No.	Meaning
	Sets the VIDEO FiLTeR band width to 100 Hz. End of program

When the value of VIDEO FiLTER band width directly, do it using the codes shown in the table below.

VIDEO FilTeR Band Width Set Value Codes

Cođe	Value of VIDEO FiLTeR Band Width
VF10Hz	10 Hz
VF100Hz	100 Hz
VF1KZ	1 kHz
VF10KZ	10 kHz
VF100KZ	100 kHz
VF300KZ	300 kHz
VF1MZ	1 MHz

6.4.7 Setting of Sweep Time (SWEEP TIME/DIV)

There are two methods available for setting the sweep time using the GPIB. One is to set the sweep by making it long (or short) in steps of 1-2-5 using the command (TU or TD) corresponding to the key on the front panel. The other method is to set the sweep time directly.

6.4 Setting of Each Function

(1) When Using the Command Corresponding to the Key

Example 6-15: Setting the Sweep Time to 200 ms/DIV.

HP200/300 Series

- 10 OUTPUT 701; "OPST"
- 20 ENTER 701; T
- 30 IF T<=0.2 THEN 60
- 40 OUTPUT 701; "TD"
- 50 GOTO 20
- 60 IF T=0.2 THEN 90
- 70 OUTPUT 701; "TU"
- 80 GOTO 20
- 90 END

Line No.	Meaning
10 20 30	Instructs this equipment to output the value of the sweep time. Reads the data (the value of the sweep time). Branches to line No. 60 when the read data is smaller than or equal to 0.2.
40 50	Sends the TIME/DIV DOWN key ② command to lower the sweep time by 1 step (to speed up the sweeping). Returns to line No. 20.
60	Branches to line No. 20.
70	Branches to line No. 90 when the read data is equal to 0.2. Sends the TIME/DIV key of command to raise the value of the sweep time by 1 step (to slowdown the sweeping).
80 90	Returns to line No. 20. End of program

D4454 Amm

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6.4 Setting of Each Function

(2) When the Sweep Time Is Set Directly

Example 6-16: Setting the Sweep Time to 200 ms/DIV Directly ${\tt HP200/300~Series}$

10 OUTPUT 701: "ST200MS"

20 END

Line No. Meaning	
	Sets the sweep time to 200 ms/DIV. End of program

When the value of the sweep time is set directly, do it using the codes shown in the table below.

Sweep Time Set Value Codes

Code	Sweep time	Code	Sweep time
ST5MS	5 ms/	ST500MS	500 ms/
ST10MS	10 ms/	ST1S	1 s/
ST20MS	20 ms/	ST2S	2 s/
ST50MS	50 ms/	ST5S	5 s/
ST100MS	100 ms/	ST10S	10 s/
ST200MS	200 ms/	ST20S	20 s/
		ST50S	50 s/
		ST100S	100 s/

6.5 Output of Setting Conditions

To make the system output the set data of measurement parameters, call it directly using the "OP" command, or make it output the mode strings to detect it.

6.5.1 "OP" Command

When making the measurement parameter output directly, use the "OP" command (Output Interrogated Parameter).

Following the "OP" command, the OP parameter code of the set data to be output is sent to this equipment.

The OP parameters of this equipment are shown below.

OP Parameter Codes

Code	Parameter output	
AT	ATTENUATOR	
CF	CENTER FREQUENCY	
MF	MARKER FREQUENCY	
ML	MARKER LEVEL	
RB	RESOLUTION BAND WIDTH	
RL	REFERENCE LEVEL	
SP	FREQ SPAN	
ST	SWEEP TIME	
VF	VIDEO FILTER BAND WIDTH	
PL	DISPLAY LINE	
OB	OCCUPIED BAND WIDTH (for R4131D)	

Program examples to output the set data are given below.

6.5 Output of Setting Conditions

Example 6-17: Setting the Value of the Center Frequency and Reference Level, and Making These Data Display by Reading It from This Equipment

HP200/300 Series

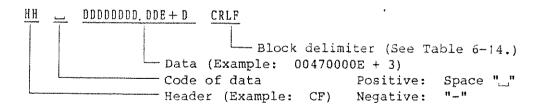
10 OUTPUT 701; "CF470MZ"
20 OUTPUT 701; "RL-30DM"
30 OUTPUT 701; "OPCF"
40 ENTER 701; F
50 OUTPUT 701; "OPRL"
60 ENTER 701; L
70 DISP F,L
80 END

Line No.	Meaning	
10	Sets the center frequency to 470 MHz.	
20	Sets the reference level to -30 dBm.	
30	Instructs this equipment to output the set data of center frequency.	
40	Reads the data and fetches it to variable F.	
50	Instructs this equipment to output the set data of the reference level.	
60	Reads the data and fetches it to variable L.	
70	Displays the value of variables F and L. The value is displayed as "470000000 -30" in this example.	
80	End of program	

After the execution of the above program, the "470000000 - 30" is displayed on the screen.

6.5.2 Format of Output Data

The format of the output data by the "OP" command is as shown below:



The data output from this equipment is all output in this format excluding the trace data and status byte. Since the total number of bytes of data is 17 bytes, make an array declaration with more than 17 bytes when the data is input as a character array variable from the GPIB controller, etc.

The header in the head of output data indicates the type of data and it varies according to the data to be output. See Item (1).

The header may be omitted when not required. The header is set to OFF by the "HD 0" command and to ON by the "HD 1" command.

Header set examples are given below:

(1) Header

The header in the head of output data indicates the type of data, and it varies according to the data to be output.

The table below shows the relation between the output data and header.

Relation Between Output Data and Header

Туре	Type of output data CENTER FREQUENCY			
CENTER				
SPAN			SP	
REFEREN	CE LEVEL	dBm	DM	
		đВµ	טם	
		đBμ∕m	VM	
		LINEAR	LV	
		dBmV	DQ	
SWEEP TIME/DIV			ST	
RESOLUTION BAND WIDTH			RB	
VIDEO FI	VIDEO FILTER		VF	
ATT	ATT		AT	
MARKER	MARKER FREQUENCY		MF	
	LEVEL	dBm	MM	
		đΒμ	MU	
		dBµ/m	ME	
	Target 2 A.	LINEAR	ML	
		dBmV	MQ	

The header may be omitted when not required.

The header is set to OFF by the "HD 0" command and to ON by the "HD 1" command. Header set examples are given below:

6.5 Output of Setting Conditions

Example 6-18: Setting the Header to OFF and Fetching the Value of Center Frequency as a Character String. Next, Setting the Header to ON and Fetching the Value of Center Frequency as a Character String.

HP200/300 Series

- 10 DIM A\$[17]
- 20 OUTPUT 701; "HDO OPCF"
- 30 ENTER 701; A\$
- 40 PRINT A\$
- 50 OUTPUT 701; "HD1"
- 60 ENTER 701; A\$
- 70 PRINT A\$
- 80 END

Line No.	Meaning	
10	Declares the length of character string A\$ to be 17 characters.	
20	Sets the header of output data of this equipment to OFF. Also, instructs this equipment to output the value of the center frequency.	
30	Reads the data and fetches it to character string variable A\$.	
40	Displays the value of character string variable A\$. When the center frequency is 400 MHz, for instance, the value is displayed as " 00400000.00E+3".	
50	Sets the header of output data of this equipment to ON.	
60	Reads the data and fetches it to character string variable A\$.	
70	Displays the value of character string variable A\$. When the center frequency is 400 MHz, the value is displayed as "CF_0040000.00E+3 ".	
80	End of program	

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6.5 Output of Setting Conditions

(2) Block Delimiter

The block delimiter indicates the end of signal.

This equipment provides four types of block delimiters as shown in the table below.

Block Delimiter Specified Codes

Cođe	Block delimiter	
DL 1	Outputs the 1-byte code of "LF".	
DL 2	Outputs the last byte of data and single-wire signal "EOI" at the same time.	
DL 3	Outputs the 2-byte codes of "CR" and "LF".	
DL 0	Outputs the 2-byte codes of "CR" and "LF". Also, outputs the single-wire signal "EOI" simultaneously with "LF".	

When a command or data is sent from the GPIB controller, etc., to this equipment, it accepts the command or data, if the sent command or data is applicable to either one of the above-mentioned block delimiters. When the block delimiter is not applicable to either one of the above four types, the GPIB of this equipment will not operate normally.

When data is fetched from this equipment, the block delimiter of this equipment must be set to that of the data receiving side (GPIB controller, etc.). Select either one of the above four types.

The block delimiter can be changed to a different type of block delimiter by sending the appropriate command for the desired block delimiter from the GPIB controller.

The block delimiter of this equipment is set to DL 3 at power ON.

6.5.3 Mode String

The set value of center frequency and frequency span of this equipment can be output the "OP" command. The setting status of the other keys (e.g., INPUT ATTENUATOR key, etc.) can be checked by the mode string when output.

The mode string is composed of seven bytes of binary code. Each byte indicates the setting status of each function of this equipment.

When the mode string is to be output, use the "OM" (OUTPUT MODE STRING) command. When this command is sent, this equipment outputs the mode string when it is specified to TALKER.

When the mode string is output, the delimiter of the data adds the EOI of the single-wire signal to the last byte (the seventh byte). The CR and LF codes are not used.

The meanings of each byte of the mode string and the functions to be read are as follows:

1st byte: Setting status of MIN INPUT ATTENUATOR

2nd byte: Setting status of 10 dB/, 2 dB/, 5 dB/, LINEAR switches

3rd byte: Setting status of the unit (UNITS switch) of the reference

level

4th byte: Setting of reference level FINE/COARSE SELECTION switch

5th byte: Setting status of trigger mode

6th byte: Definition of whether the setting of data knob is CENTER

FREQ or MARKER

7th byte: Definition of whether the AFC mode is ON or OFF

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6.5 Output of Setting Conditions

Mode String

Byte # 7 6 5 4 3 2 1 0				
0 0 0 0 0 0 0 0 0 1	Byte #		1	Description
20 dB 0 0 0 0 0 0 1 0 1 3 30 dB 0 0 0 0 0 0 1 0 1 5 50 dB 2 0 0 0 0 0 0 1 0 1 5 50 dB 2 0 0 0 0 0 0 1 1 3 3 0 0 0 0 0 0 0 1 0 1 5 50 dB 2 0 0 0 0 0 0 0 1 1 3 5 dB/DIV 0 0 0 0 0 0 0 1 1 3 5 dB/DIV(QP) 0 0 0 0 0 0 0 1 1 3 LINEAR 3 0 0 0 0 0 0 0 1 1 3 LINEAR 3 0 0 0 0 0 0 0 1 1 3 LINEAR 6 0 0 0 0 0 0 0 1 1 1 5 CARSE 0 0 0 0 0 0 0 1 1 1 5 CARSE 0 0 0 0 0 0 0 1 1 1 1 1 1 1 1 1 1 1 1 1	1	00000000	0	INPUT ATTENUATOR: 0 dB
20 dB 0 0 0 0 0 0 0 1 0 2 0 0 0 0 0 0 1 1 3 30 dB 0 0 0 0 0 0 1 0 1 0 4 0 0 0 0 0 0 1 0 1 5 50 dB 2 0 0 0 0 0 0 0 1 0 1 5 50 dB 2 0 0 0 0 0 0 0 1 0 1 5 50 dB 2 0 0 0 0 0 0 0 1 1 3 0 0 0 0 0 0 0 1 1 3 0 0 0 0 0 0 0 1 1 3 0 0 0 0 0 0 0 1 1 3 0 0 0 0 0 0 0 1 1 3 0 0 0 0 0 0 0 1 1 3 0 0 0 0 0 0 0 1 1 3 0 0 0 0 0 0 0 1 1 3 0 0 0 0 0 0 0 1 1 3 0 0 0 0 0 0 1 1 1 3 0 0 0 0 0 0 1 1 1 3 0 0 0 0 0 0 1 1 1 3 0 0 0 0 0 0 1 1 1 3 0 0 0 0 0 0 1 1 1 3 0 0 0 0 0 0 1 1 1 3 0 0 0 0 0 0 1 1 1 5 0 0 0 0 0 0 1 1 1 7 0 0 0 0 0 0 1 1 1 7 0 0 0 0 0 0 1 1 1 7 0 0 0 0 0 0 1 1 1 7 0 0 0 0 0 0 1 1 1 7 0 0 0 0 0 0 1 1 1 7 0 0 0 0 0 0 1 1 1 7 0 0 0 0 0 0 1 1 1 7 0 0 0 0 0 0 1 1 1 7 0 0 0 0 0 0 0 1 1 1 1 7 0 0 0 0 0 0 0 1 1 1 1 7 0 0 0 0 0 0 0 1 1 1 1 1 1 1 1 1 1 1 1 1		00000001	1	10 dB
3 0 0 0 0 0 0 0 1 1 3 4 40 dB 0 0 0 0 0 0 0 1 0 1 5 50 dB 2 0 0 0 0 0 0 0 0 1 1 5 5 50 dB 2 0 0 0 0 0 0 0 1 1 3 3 Tube surface ordinates 10 dB/DIV 0 0 0 0 0 0 0 1 1 3 Tube surface ordinates 10 dB/DIV 0 0 0 0 0 0 0 1 1 3 Tube surface ordinates 10 dB/DIV 0 0 0 0 0 0 0 1 1 3 Tube surface ordinates 10 dB/DIV 0 0 0 0 0 0 0 1 1 3 Tube surface ordinates 10 dB/DIV 0 0 0 0 0 0 0 1 1 3 Tube surface ordinates 10 dB/DIV 0 0 0 0 0 0 0 1 1 3 Tube surface ordinates 10 dB/DIV 0 0 0 0 0 0 0 1 1 3 Tube surface ordinates 10 dB/DIV 0 0 0 0 0 0 0 1 1 3 Tube surface ordinates 10 dB/DIV 0 0 0 0 0 0 0 1 1 3 Tube surface ordinates 10 dB/DIV 0 0 0 0 0 0 0 1 1 3 Tube surface ordinates 10 dB/DIV 0 0 0 0 0 0 0 1 1 3 Tube surface ordinates 10 dB/DIV 0 0 0 0 0 0 0 1 1 3 Tube surface ordinates 10 dB/DIV 0 0 0 0 0 0 0 1 1 3 Tube surface ordinates 10 dB/DIV 0 0 0 0 0 0 0 0 1 1 3 Tube surface ordinates 10 dB/DIV 0 0 0 0 0 0 0 0 0 0 0 0 0 0 Tube surface ordinates 10 dB/DIV 0 0 0 0 0 0 0 0 0 0 0 0 0 0 Tube surface ordinates 10 dB/DIV 0 0 0 0 0 0 0 0 0 0 0 0 0 0 Tube surface ordinates 10 dB/DIV 0 0 0 0 0 0 0 0 0 0 0 0 0 Tube surface ordinates 10 dB/DIV 0 0 0 0 0 0 0 0 0 0 0 0 0 Tube surface ordinates 10 dB/DIV 0 0 0 0 0 0 0 0 0 0 0 0 0 Tube surface ordinates 10 dB/DIV 0 0 0 0 0 0 0 0 0 0 0 0 0 Tube surface ordinates 10 dB/DIV 0 0 0 0 0 0 0 0 0 0 0 0 0 Tube surface ordinates 10 dB/DIV 0 0 0 0 0 0 0 0 0 0 0 0 0 Tube surface ordinates 10 dB/DIV 0 0 0 0 0 0 0 0 0 0 0 0 0 Tube surface ordinates 10 dB/DIV 0 0 0 0 0 0 0 0 0 0 0 0 Tube surface ordinates 10 dB/DIV 0 0 0 0 0 0 0 0 0 0 0 0 0 Tube surface ordinates 10 dB/DIV 0 0 0 0 0 0 0 0 0 0 0 0 Tube surface ordinates 10 dB/DIV 0 0 0 0 0 0 0 0 0 0 0 0 Tube surface ordinates 10 dB/DIV 0 0 0 0 0 0 0 0 0 0 0 0 0 Tube surface ordinates 10 dB/DIV 0 0 0 0 0 0 0 0 0 0 0 0 0 Tube surface ordinates 10 dB/DIV 0 0 0 0 0 0 0 0 0 0 0 0 0 Tube surface ordinates 10 dB/DIV 0 0 0 0 0 0 0 0 0 0 0 0 0 0 Tube surface ordinates 10 dB/DIV 0 0 0 0 0 0 0 0 0 0 0 0 0 0 Tube surface surface surface surface surface surface surface su		00000010	2	
0 0 0 0 0 0 1 0 0 0 0 0 0 0 0 0 0 0 0		00000011	3	
2 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0		00000100	4	
0 0 0 0 0 0 0 1 1 1 axis display: 2 dB/DIV 2 dB/		00000101	5	50 dB
C C C C C C C C C C	2	0 0 0 0 0 0 0 0	0	Tube surface ordinates 10 dB/DIV
0 0 0 0 0 0 0 1 1 3 LINEAR		0 0 0 0 0 0 0 1	1	axis display: 2 dB/DIV
3		0 0 0 0 0 0 1 0	2	5 dB/DIV(QP)
Color Col		00000011	3	LINEAR
C C C C C C C C C C	3	0 0 0 0 0 0 0 0	0	
0 0 0 0 0 0 0 1 1 3 dBμ/m(B) dBμ/m(C) dBμ/m(C) dBμ/m(C) dBμ/m(D) dBμ/m(D) mV, μV 0 0 0 0 0 0 1 1 1 1 7 dBmV		0 0 0 0 0 0 0 1	1	REFERENCE LEVEL:
0 0 0 0 0 1 0 0 4 dB\mu/m(C) dB\mu/m(D) dB\mu/m(D		00000010	2	dBμ/m(A)
0 0 0 0 0 1 0 1 5 dBμ/m(D) mV, μV 0 0 0 0 0 0 1 1 1 1 7 dBmV 4 0 0 0 0 0 0 0 0 1 1 1 1 7 FINE 5 0 0 0 0 0 0 0 0 0 0 0 1 1 1 1 FINE 5 0 0 0 0 0 0 0 1 1 1 1 LINE 0 0 0 0 0 0 0 1 1 1 3 SINGLE 6 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0	TOTAL	00000011	. 3	dBμ∕m(B)
0 0 0 0 0 1 1 0 6 mV, µV 0 0 0 0 0 0 1 1 1 7 dBmV 4 0 0 0 0 0 0 0 1 1 1 FINE 5 0 0 0 0 0 0 0 0 0 TRIGGER MODE: FREE RUN 0 0 0 0 0 0 0 1 1 LINE 0 0 0 0 0 0 1 1 3 SINGLE 6 0 0 0 0 0 0 0 1 1 1 CF		00000100	4	dBµ∕m(C)
0 0 0 0 0 1 1 1 7 dBmV		00000101	5	dBµ∕m(D)
4 0 0 0 0 0 0 0 0 0 0 0 0 FEFERENCE LEVEL: COARSE 0 0 0 0 0 0 0 0 1 1 FINE 5 0 0 0 0 0 0 0 1 1 TRIGGER MODE: FREE RUN 0 0 0 0 0 0 0 1 1 TRIGGER MODE: VIDEO 0 0 0 0 0 0 1 1 3 SINGLE 6 0 0 0 0 0 0 0 0 1 1 1 CF		00000110	6	mV, μV
0 0 0 0 0 0 0 0 1		00000111	7	dBmV
5 0 0 0 0 0 0 0 0 0 0 0 TRIGGER MODE: FREE RUN 0 0 0 0 0 0 0 1 1 LINE 0 0 0 0 0 0 0 1 0 2 VIDEO 0 0 0 0 0 0 1 1 3 SINGLE 6 0 0 0 0 0 0 0 0 0 DATA KNOB: MARKER 0 0 0 0 0 0 0 1 1 1 CF	4	0000000	0	REFERENCE LEVEL: COARSE
0 0 0 0 0 0 0 1 1 1 LINE 0 0 0 0 0 0 1 0 2 VIDEO 0 0 0 0 0 0 1 1 3 SINGLE 6 0 0 0 0 0 0 0 0 0 DATA KNOB: MARKER 0 0 0 0 0 0 0 1 1 1 CF		00000001	1	FINE
0 0 0 0 0 0 1 0 2 VIDEO 0 0 0 0 0 0 1 1 3 SINGLE 6 0 0 0 0 0 0 0 0 0 DATA KNOB: MARKER 0 0 0 0 0 0 0 1 1 CF	5	0 0 0 0 0 0 0	0	TRIGGER MODE: FREE RUN
0 0 0 0 0 0 1 1 3 SINGLE 6 0 0 0 0 0 0 0 0 DATA KNOB: MARKER 0 0 0 0 0 0 0 1 1 CF	***************************************	00000001	1	LINE
6 0000000 0 DATA KNOB: MARKER 00000001 1 CF		00000010	2	VIDEO
0 0 0 0 0 0 0 1 1 CF		00000011	3	!
	6	00000000	0	DATA KNOB: MARKER
		00000001	1	CF
7 0000000 0 AFC: OFF	7	0000000	0	AFC: OFF
0 0 0 0 0 0 0 1 1 ON		00000001	1	ON

6.5 Output of Setting Conditions

Example 6-19: Detecting the Value of Attenuator by Making the Mode String Output

HP200/300 Series

- 10 DIM M(6)
- 20 OUTPUT 701; "OM"
- 30 ENTER 701 USING
- "#, B"; M(*)
 40 DISP M(0)
- 50 END

Line No.	Meaning	
10 20 30 40 50	Secures 7 bytes for variable M. Specifies the output of the mode string. Fetches the mode string. Displays the 1st byte (ATTENUATOR) of the mode string. End of program	

6.6 Input/Output of Trace Data

This equipment can output the trace data (waveform displayed on the screen). It also can input the same data from outside. This function makes it possible to analyze and arithmetically process the waveform data using the controller.

The trace data on the screen of this equipment is composed of 701 points of data on the frequency axis (horizontal axis). For input/output of the trace data, this 701-point data is input or output from the left (lower ones in frequency) sequentially. The trace data of each point is expressed with integers from 0 to 511 (Figure 6-5).

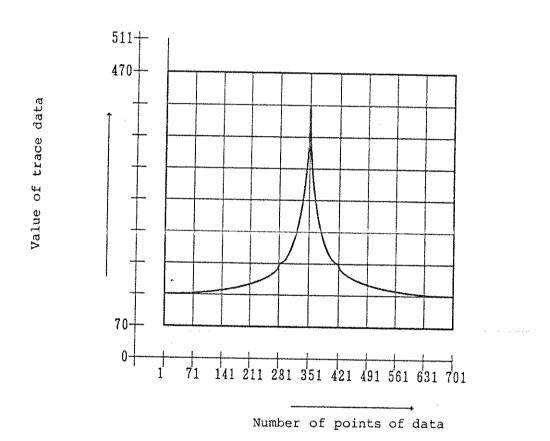


Figure 6-5 Correlation Between Screen Grids and Trace Data

The input/output of trace data can be made in two forms, ASCII code and binary code. Of the two, the ASCII code is convenient when data is input or output point by point. When the data is input or output for one screen (701 points) all together, the binary code is faster in finishing the processing. Use these two ways case by case.

6.6.1 Output of Trace Data

The "OP" command is used for the output of trace data. When the parameter code is sent in succession to the "OP" command, the desired trace data can be output. For the parameter codes of trace data, see the table below.

Trace Data Parameter Codes

Code	Data to be input or output	Type of data
TAA	Trace data of VIEW screen memory	ASCII code
TAW	Trace data of WRITE screen memory	
TBA	Trace data of VIEW screen memory	Binary code
TBW	Trace data of WRITE screen memory	

(1) Method to Output the Trace Data with ASCII Code

OUTPUT 701; "OPTAW"

When this program is executed, this equipment outputs the trace data of the WRITE screen memory with the ASCII code when it is specified to TALKER.

ENTER 701; A

When this program is executed, the trace data for one point is fetched to variable A. When the same ENTER statement is executed, the trace data of the second point, third point ... can be obtained sequentially.

The data format at this time is expressed in 4-digit numerics with no header as shown below:

When the trace data is fetched as a character string variable, declare the array by setting the length of the character string variable used to more than 4 bytes.

6.6 Input/Output of Trace Data

A program example to output the trace data with ASCII code

Example 6-20: Output the trace data in memory with ASCII code, and store in array variable.

HP200/300 Series

- 10 DIM A(700)
- 20 OUTPUT 701; "OPTAW"
- 30 FOR I=0 TO 700
- 40 ENTER 701; A(I)
- 50 NEXT I
- 60 END

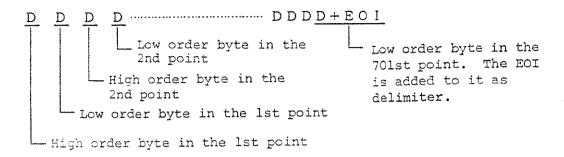
Line No.	Meaning
10	Declares array variable A(I) up to 701 points.
20	Instructs this equipment to output the trace data of the WRITE screen memory with the ASCII code.
30	Instructs this equipment to vary variable I from 0 to 700 one by one. (The loop is repeated 701 times.)
40	Reads the trace data for one point and stores it in array variable A(I).
50	Increments variable I by 1 only, and returns to line No. 40 when I $<$ 700, but runs on to the next line when I \geq 700.
60	End of program

(2) Method to Output Data with the Binary Code

OUTPUT 701; "OPTBW"

When this program is executed, this equipment outputs the trace data of the WRITE screen memory with the binary code when it is specified to TALKER. Since 701 points of trace data (for 1 screen) is output all together at this time, the controller side should be ready to input the 701 points of data at the one time. Also, since the EOI signal is specified to the delimiter when the data is output with the binary code, the controller side should continue the data input until the EOI signal can be detected.

The data output format with the binary code is shown below:



One point of data consists of 9 bits in the binary code. Consequently, one point of data is expressed in 2 bytes which are divided into high order byte and low order byte. When the data is output to the GPIB, the upper byte in the first is output first and then the low order byte in the first point, followed by the high order byte in the second point and so forth, and lastly the low order byte in the 701st point.

Example 6-21: The trace data in the memory is output with the binary code to be stored in an array variable.

HP200/300 Series

- 10 DIM A(700)
- 20 OUTPUT 701; "OPTBW"
- 30 FOR I=0 TO 700
- 40 ENTER 701 USING "#,W"; A(I)
- 50 NEXT I
- 60 END

Line No.	Meaning
10 20	Declares numeric array variable A(I) for as many numbers as required. Instructs this equipment to output the trace data in the WRITE
20	screen memory with the binary code.
30	Instructs this equipment to vary variable I from 0 to 700 one by one. (The loop is repeated 701 times.)
40	Fetches 2-byte binary data, converts it into decimal data, and stores it in numeric array variable A(I). Then, increments variable
50	I by 1 only. When I is < 700 , the program execution returns to the preceding line. When I ≥ 700 , it proceeds to the next line.
60	End of program. (Usually, the trace data execution program is input after this.)

6.6 Input/Output of Trace Data

6.6.2 Input of Trace Data

The "IN" command is used to input the trace data in R4131. When the parameter code of trace code is sent to this equipment after the "IN" command, the desired trace data can be input. The parameter code of trace data used for this input is the same as the code used in its output.

(1) Method to Input the Trace Data with the ASCII Code

OUTPUT 701; "INTAA"

When programmed and executed like this, this equipment enters the input mode of the trace data. When the data is sent to this equipment with the ASCII code after this, that data is stored in the first point of the VIEW screen memory.

When the data is sent further, the trace data is set to the second point, third point ... in the memory, sequentially.

If any data other than the trace data is sent to the equipment under this status, this equipment automatically exits from the trace data input mode and returns to its routine status.

The data format is the same as that when the data is output with the ASCII code.

A program example to input the trace data with the ASCII code

Example 6-22: The trace data is assumed to be provided in numeric array variable A(I). The data in A(I) is then input to the VIEW screen memory of this equipment with the ASCII code.

HP200/300 Series

•

100 OUTPUT 701; "INTAA"

110 FOR I=0 TO 700

120 OUTPUT 701; INT(A(I))

130 NEXT I

140 END

Line No.	Meaning
100	Instructs this equipment to receive the trace data to the VIEW screen memory with the ASCII code.
110	Instructs this equipment to vary variable I from 0 to 700, one by one. (The loop is repeated 701 times.)
120	Converts the data in array A(I) into integers and sends it to this equipment.
130	Increments the value of variable I by 1 only. When I \leq 700, the program execution returns to line No. 120. When I \geq 700, it proceeds to the next line.
140	End of program

When this equipment is set to the VIEW mode after the execution of this program, it is possible to see the tracing waveform by the input data.

(2) Method to Input the Trace Data with the Binary Code

OUTPUT 701; "INTBA"

When programmed and executed like this, this equipment enters the trace data input mode with the binary code. In the binary code, input the trace data for one screen (701 points) all together at a time. Since R4131 continues the data input until the EOI signal is detected, be sure to add the EOI to the last byte of the trace data.

The data format is the same as in the output of the trace data with the binary code. A program example for the input of trace data is as follows:

A program example to input the trace data with the binary code

Example 6-23: The trace data is assumed to be provided in the numeric array variable A(I). The data in A(I) is then input in the VIEW screen memory of this equipment with the binary code.

HP200/300 Series

100 OUTPUT 701; "INTBA"

110 FOR I=0 TO 699

120 OUTPUT 701 USING "#,W"; A(I)

130 NEXT I

140 OUTPUT 701 USING "#,w"; A(I), END

150 END

6.6 Input/Output of Trace Data

Line No.	Meaning
100	Instructs this equipment to receive the trace data in its VIEW screen memory with the binary code, and to make a change so that the
110	EOI is added to the last byte of the delimiter. Instructs this equipment to vary variable I from 0 to 699, one by
	one. (The loop is repeated 701 times.)
120	Converts the data of numeric array A(I) into 2-byte binary code and sends it to this equipment.
130	Increments variable I by 1 only. When I < 699, the program execution returns to the preceding line. When I \geq 699, it proceeds to the next line.
140 150	Adds the EOI signal when the last point data is set. End of program

When this equipment is set to the VIEW mode after the execution of the above program, it is possible to see the trace data input through the input data.

6.7 Service Request

By using the service request function of GPIB, various statuses of this equipment can be detected from the outside.

Contents of the service request can be known from status bytes shown in Table 6-17.

Bit #	Decimal value	Function
7	128	End of sweep
6	64	Service request (SR
5	32	
4	16	CF CAL

8

2

1

3

1

Signal track

Marker search

ZERO CAL

Center frequency set

Status Byte

(1) Status Byte

Each bit of the status byte is set to "1" when the following conditions are met.

Status byte

- Bit 0: "1" is set when ZERO CAL is executed and the calibration is finished.
- Bit 1: "1" is set when the center frequency is set using the "CF" command of GPIB.
- Bit 2: "1" is set if the marker ends the searching when the searching function is executed by the marker.
- Bit 3: This bit is changed from 0 to 1 when the waveform peak position is ended to be set to the center frequency during the execution of the signal tracking function of marker.
- Bit 4: """ is set when the CF CL is executed and the calibration is finished.
- Bit 6: When "1" is set to either bit 0 to bit 5, or bit 7 and the service request (SRQ) is transmitted, this bit also goes to "1" at the same time.
- Bit 7: "1" is set when the sweeping ends.

This service request is turned ON/OFF by GPIB commands "SO" and "S1". When the status byte is read, this equipment clears the status byte.

(2) Output of Status Byte

The status byte can be read when the serial polling is executed as shown in the following example:

Example 6-24: ZERO CAL is judged to be ended by reading the status byte.

HP200/300 Series

- 10 OUTPUT 701; "SHFL"
- 20 S=SPOLL(701)
- 30 IF BIT(S,O) <>1 THEN 20
- 40 OUTPUT 701; "CF200MZ SP100KZ"
- 50 END

Line No.	Meaning
10	Executes the ZERO CAL.
20	Reads the status byte and incorporates it in variable S.
30	Waits until bit #0 becomes 1 after the end of the execution of ZERO CAL.
40	For the next setting after the end of ZERO CAL, the center frequency is set to 200 MHz and spans to 100 kHz in this stage.
50	End of program

6.8 Notes in Programming

(1) Noteworthy Points in Sending a Command

When a command is sent to this equipment, the command can be delimited with a space (_) or comma (,) as shown below:

Example 6-25: A command is delimited with a space $(_)$ or comma (,) and sent to this equipment.

OUTPUT 701; "SO OPCF, HD 1"

(2) Noteworthy Points in Spectrum Analysis When the Frequency Span Is Made Narrower

The center frequency setting accuracy is ± 10 MHz or less when R4131C/CN and the AFC of R4131D/DN are set to OFF. Hence, when the center frequency is set directly by setting the frequency span to less than 10 MHz, no spectrum is displayed on the screen in some cases.

Consequently, when the spectrum is analyzed by making the frequency span narrow, try to program so that narrow the span narrows while always seizing the signal.

Example 6-26: The frequency span is made narrow up to 50 kHz for the 200 MHz reference signal.

HP200/300 Series

- 10 OUTPUT 701; "CF 200MZ, SP20MZ, RL-30DM"
- 20 WAIT 1
- 30 OUTPUT 701; "SHM4"
- 40 S=SPOLL(701)
- 50 IF BIT(S,3)<>1 THEN 40
- 60 OUTPUT 701; "NR"
- 70 OUTPUT 701; "OPSP"
- 80 ENTER 701; A
- 90 IF A <> 50000 THEN 40
- 100 END

Line No.	Meaning
10	Sets the center frequency to 200 MHz, frequency span to 20 MHz, and reference level to -30 dBm.
20	Waits for 1 sec.
30	Sets the signal tracking function to ON.
40	Reads the status byte and incorporates it to variable S.
50	After the end of signal tracking, waits until bit #3 becomes 1.

Line No.	Meaning
60 70 80 90 100	Makes the frequency span narrower by 1 step. Reads the frequency span and sets the mode. Reads the data. Returns to line 40 unless the frequency span is 50 kHz. End of program

(3) Noteworthy Points for the Setting of Center Frequency When the Frequency Span Is Less Than 10 MHz

When the center frequency is changed in the setting of the frequency span to less than 10 MHz, the spectrum shifts after the setting, although varied according to the amount of change. This is caused by the time constant of the frequency stabilization circuit. Note that no correct data is indicated in the case of a program used to read the marker frequency level under this status.

Example 6-27: When the Frequency of the 200 MHz Reference Signal Is Read

HP200/300 Series

10 OUTPUT 701; "CF 3500MZ SP 10MZ"
20 WAIT 1
30 OUTPUT 701; "CF 200MZ"
40 WAIT 10
50 OUTPUT 701; "M4"
60 OUTPUT 701; "OPMF"
70 ENTER 701; F

Line Meaning No. 10 Sets the center frequency to 3500 MHz and frequency span to 10 MHz. 20 Sets the waiting time for 1 sec. 30 Sets the center frequency to 200 MHz. 40 Takes the waiting time here until the spectrum is stabilized (approx. 10 sec. maximum). The waiting time is set to 10 sec. in this example. 50 Executes the PEAK SEARCH. Reads the marker frequency. 60 70 Incorporates the marker frequency to variable F. 80 Displays the marker frequency.

6.9 List of GPIB Codes

Table 6-4 List of GPIB Codes

Setting	lode	Remarks	Setting	Çode	Remarks
(aput of measuring condition		code corresponding to each key gure 6-1.	inpet of trace data	18	Memory, ASCII/ binary specified code is the same as in its output.
Out; 1 0 00 0A condition and state date	1 00GB.		Output of the status byte	0\$	The EOI is added to the last byte of data as a delimiter. (CR LF is not used.)
OP parimeter dode		Output format of output data	Cusput of the mode string	OM	ter is not used.
ATTENLATOR CENTER PROCUENCY MARKER FREQUENCY MARKER LEVEL	AT Symbol formation outside said. SE Symbol of bytes: 17 (except delimiters) MF HH BDBDDDDDDDDD DD E=3 CRuf		Service request Transmitted Not transmitted	\$0 \$1	"Si" at the power ON
RESOLUTION BAND WIDTH REFERENCE LEVEL	RB	Block delimiter	lnitial:zation	19	
FREQ SPAN SWEEP TIME VIDEO FILTER BAND WIGTH DISPLAY LINE OCCUPIED FREQUENCY BAND WIGTH	SP ST VF PL PL	Data index part Data mantissa Positive: Space " "(Blank)	Header OFF ON Heade	HDO HD1 r to be	"HD1" at the power ON output
(R41219 only) Trace memory, and ASCII/ binary specified code Trace data of memory A ASCII output Binary output		Output format of trace data ASC11 DDDD CRLF (Number of bytes: 4 except Jelimiters) Binary DDD-DDD (Number of bytes: 1402)	REFERENCE RESOLUTION DU BAN DU VIDE DU BAN ABB DA VIDE DE BAN	P TIME ILUTION ID WIDTH IO FILTER	ST FREQUENCY RB LEVEL dBm MM dB \(\mu \) MG RU VF dB \(\mu / \mu \) ME
WRITE memory trace data ASCII output Binary output	TAW	except delimiters High/low order bytes in the 701st point High order byte in the 2nd point Low order byte in the 1st point High order byte in the 1st point (1-point data in 2 bytes:	Block delimiter CR. LF760 LF EOI CR. LF	DL1 DL2	"DES" at the power ON

Table 6-5 GPIB Code Corresponding to Each Key

Кеу	Code	Key	Code
INSTR PRESET	IP (SHMO)	ATT 0dB	A0
CTR FREQ DATA KNOB	CF	VIDEO FLTR	
COARSE DOWN	!		VU
()) FINE DOWN	CU FD	DOWN SWEEP TIME/BIV	VD
UP	FU	☐ UP	TU
MARKER ON OFF	M1 M0	☐ DOWN	TD
MKR CF PEAK	м3	TRIGGER	TR
CF CAL	M4 FL	START/RESET	SR
FREQ SPAN	SP	LCL	LC
ZERO SPAN AUTO	ZS (SHSP)	WRITE	WR
RBW	BA RB	STORE VIEW	SE
FREQ SPAN, RBW	19	MAX HOLD	VW MA (SHWR)
	WD	RECALL	RC
∞ NARROW	NR	SAVE	SV (SHRC)
		CF ADJ	SHCF
O UP	LU	OBW AFC	SHM1 *
DOWN	LD	SIG TRK	SHM3 ** SHM4
FINE/COARSE	FC	ZERO CAL	SHFL
10dB/DIV	L1 (SHLD)	NOISE/Hz NORMALIZATION	SHBA SHRB
2dB/DIV	L2 (SHLU)	DSPL LINE	SHWD
QP LINEAR	L3 (SHFC) LN (SHUN)	NORMAL DET	SHVD **
UNITS INPUT ATTENUATOR	UN	POS DET	SHTD
⊕ UP	AU	SAMPLE DET	SHTR
Ø down	AD		

Note: Codes marked with one asterisk (*) are available for R4131D. Codes marked with two asterisks (**) are available for R4131D/DN only.

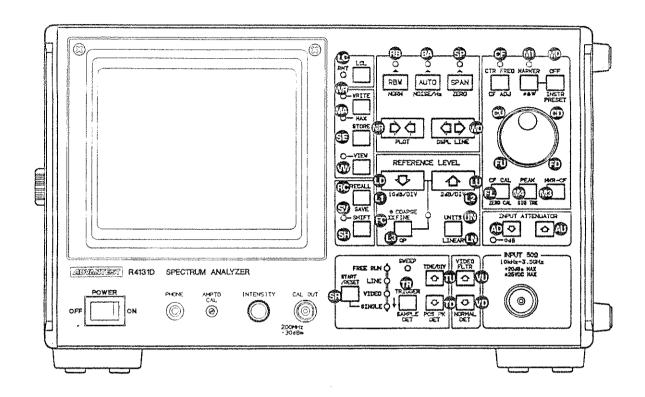


Figure 6 - 6 GPIB Code for each Key

Table 6 - 6 Direct Set GPIB Codes

Cont	Contents		
ПМітć	dBm dB μ dB μ /n (Λ) dB μ /m (Β) dB μ /m (C) dB μ /m (D) dBmV	DM DU D1 D2 D3 D4 DV	
Trigger Mode	FREE RUN LINE VIDEO SINGLE	FR LI VT SI	
Attenuator	0 dB 10 dB 20 dB 30 dB 40 dB 50 dB	A0 A1 A2 A3 A4 A5	
Conte	nts	data Code+ 🗆 🗇	
Center frequence less requency spreading to harker Video filter Sweep time Display line	CF		

Table 6 - 7 Unit Display GPIB Codes

Unit	Code
GHz	GZ
MHz	MZ
kHz	K Z
Ηz	НZ
V	V
m V	MV
μ V	UV
sec	S
msec	MS
d B m	DM
đB μ	DU
dBμ/m (A)	D1
dBμ/m (B)	D2
$dB\mu/m$ (C)	D3
$dB\mu/m$ (D)	D4

Table 6 - 8 Numreric Value Code in Setting Condition Input

Code Set value VF10HZ Video band width VF100HZ VF1KZ 100Hz 1kz VF10KZ VF100KZ VF300KZ VF1MZ 10kz100kz 300kz 1Mz 5 ms/ 10 ms/ 20 ms/ 50 ms/ 100 ms/ 200 ms/ ST5MS ST10MS ST20MS ST30MS ST300MS ST200MS ST300MS ST1S ST3S ST3S ST3S ST3S ST30S ST30S ST30S ST100S Sweep 1 s/ 2 s/ 5 s/ 10 s/ 20 s/ 50 s/ 100 s/ 0 dB 10 dB 20 dB 30 dB 40 dB 50 dB Altenuator A1 A2 A3 A4 Ä5 50 kHz 100 kHz 200 kHz 500 kHz SP50KZ SP100KZ SP200KZ SP500KZ SP1MZ SP2MZ 1 MH2 2 MHz span SPEMZ SP5MZ SP10MZ SP20MZ SP20MZ SP100MZ SP200MZ SP300MZ SP500MZ SP50CMZ SP2GZ SP4GZ 2 MHz 5 MHz 10 MHz 20 MHz 100 MHz 200 MHz 200 MHz Frequency 1 GHz 2 GHz 4 GHz ZEROSPAN 1 kHz 3 kHz 10 kHz 30 kHz RB1KZ RB3KZ Resolution band width RB10KZ RB30KZ RB100K2 RB300KZ RB1MZ 300 kHz 1 MHz

Table 6 - 9 Mode String

*			
Byte	Bit 76543210	Decimal value	Contents
	00000000 00000001 00000010 00000011 00000100 00000101	0-20345	INPUT ATT 0 dB 10 dB 20 dB 30 dB 40 dB 50 dB
2	00000000 0000001 00000010 00000011	0	Tube surface ordinates axis display 10 dB/DIV 2 dB/DIV 5 dB/DIV(QP) LINEAR
3	00000000 00000001 00000010 00000011 00000100 00000101 00000110 00000111	01234567	Ordinates axis unit, dBm dB \(\mu \) (M) dB \(\mu / \mu (A) \) dB \(\mu / \mu (B) \) dB \(\mu / \mu (D) \) mV, \(\mu V \) dBmV
4	00000000 00000001	0	REF LVL STEP SIZE: COARSE FINE
5	00000000 00000001 00000010 00000011	0123	TRIGGER MODE FREE RUN LINE VIDEO SINGLE
6	00000000	0	Data knob Marker CF
7	00000000 00000001	0	AFC OFF ON

Table 6 - 10 Status Byte

Bit	Decimal value	Function (set to 1 when ended)
76543210	128 64 32 16 8 4 2	End of sweeping Service request CF CAL Signal track Marker search Center frequency setting ZERO CAL

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R4131 SERIES SPECTRUM ANALYZER INSTRUCTION MANUAL

7. INSPECTION AND MAINTENANCE

7. INSPECTION AND MAINTENANCE



SPECTRUM ANALYZER INSTRUCTION MANUAL

7.1 Defects and Abnormal Stresses

Defects and Abnormal Stresses

When the R4131C, R4131CN, R4131D, R4131DN is impaired as undermentioned, it is thought that the protective function is damaged.

Before the R4131C, R4131CN, R4131D, R4131DN is used, make sure to find the damage and ensure the safety of this equipment at your nearest support office.

The instruments:

- show visible damage,
- fails to perform the intended measurements,
- has been subjected to prolonged storage under unfavourable conditions,
- has been subjected to severe transport stresses.

7.7	**	7	XT	7	27	\sim
W	А	r	1.₩	. <u>Ł</u>	11	C

To remov the unit case is allowed only for the trained service personnel because there is danger of the electric shock.

- 7.2 Notes in Storing and Shipping this Epuipment
- 7.2.1 Storage of This Equipment

The storage temperature range of this equipment is -20°C to $+70^{\circ}\text{C}$. When this equipment is not used for a long period of time, cover it with vinyl or put in a cardboard box, and store it in a dry place away from direct sunlight.

7.2.2 Cleaning of This Equipment

Periodically take off the filter which protects the CRT display and clean the inside of the filter and CRT display unit with a soft cloth soaked in alcohol. Do not use any cleaner other than alcohol.

The filter can be taken off by removing two screws of the bezel.

CAUTION

Never use any cleaner other than alcohol for the maintenance of this equipment.

Organic solvent such as benzene, toluene or acetone may spoil the plastic parts of this equipment.

7.2.3 Shipment of This Equipment

When shipping this equipment, use the original packing materials. If they are not available, pack the equipment as follows:

- (1) Wrap this equipment in appropriate shock absorbing material and put it in a corrugated cardboard box at least 5 mm thick.
- (2) Wrap its accessories separately in the same shock absorbing material and put them in the same corrugated cardboard box together with this equipment.
- (3) Fasten the corrugated cardboard box with packing strings.

MEMO Ø

R4131 SERIES SPECTRUM ANALYZER INSTRUCTION MANUAL

8. Technical Data of Function and Accessories

8. TECHNICAL DATA OF FUNCTION AND ACCESSORIES

R4131 SERIES SPECTRUM ANALYZER INSTRUCTION MANUAL

8.1 Technical Data of Function

8.1 Technical Data of Function

(1) Frequency Specification

Frequency range

: 10 kHz to 3.5 GHz

Frequency display

: Displayed on the CRT screen

Maximum resolution: 1 kHz (to be changed

according to the frequency span)

Frequency displaying accuracy:

R4131C/CN	Less than ±10 MHz	After ZERO CAL
R4131D/DN	±100 kHz + SPAN 3% or less	After ZERO CAL Within the range of 0 Hz to 2.5 GHz in center frequency and 5 ms to 0.5 S/DIV in sweep time.
	±10 MHz	After ZERO CAL Center frequency 2 GHz or more

Frequency span

: 4 GHz to 100 kHz, ZERO

1-2-5 step

Frequency span accuracy:

: ±5% : R4131C/CN

Frequency stability

Less than 100 kHz/5 min.

Frequency is fixed after warming up for 1

hour under constant temperature.

R4131D/DN

Less than 10 kHz/10 min.

AFC ON

Frequency is fixed after warming up for 1

hour under constant temperature.

(Within the range of 0 Hz to 2.5 GHz in

center frequency, 5 ms to 0.5 S/DIV in sweep

time)

Residual FM

Less than 2 kHzp-p/100 ms

Noise sideband

More than 80 dBc	Where the resolution band width is assumed to be 1 kHz, video filter band width to be
	10 Hz, and 20 kHz to be detuned from signal.

8.1 Technical Data of Function

```
Resolution:
      Resolution band width
         3 dB ...... 1 kHz to 1 MHz with 1-3 step
         6 dB ..... 9 kHz to 120 kHz when QP mode is selected
      Band width selectivity
                         : Less than 15:1 60 dB: 3 ratio of dB
                                                    resolution band width
      Resolution band width accuracy
                          : Less than ±20%
                             Less than the value of CISPR Standards in
                             the OP mode
                        : Can be set freely
   Marker display
      Resolution ...... 1 kHz max. (To be changed according to the
                             SPAN)
      Measuring accuracy ... Center frequency display accuracy +
                             frequency span accuracy
(2) Amplitude Specification
   Tube surface display range
                            LOG 80 dB 10 dB/DIV
                                         2 dB/DIV
                                 20 dB
                                          5 dB/DIV, In the QP mode only
                                 40 dB
                             LIN 10 DIV
                          : LOG ±0.15 dB/1 dB
   Linearity
                                 ±1 dB/10 dB
                                 \pm 1.5 dB/70 dB or more
                             Less than 5% of LIN scale
   Reference level : LOG -69 dBm to +40 dBm: R4131C/D,
                                 40.25 dBµ to 150 dBµ: R4131CN/DN
                                 10 dB, 1 dB step 10 dB/DIV
                              1 dB, 0.25 dB step 2 dB/DIV,
                                                      in the QP mode
                             LIN 72.77 µV to +22.36 V: R4131C/D
                              (102.9 μV to +31.62 V: R4131CN/DN)
    Reference level accuracy
                             Less than ±1 dB in the LOG mode
                              This value is taken after calibrating the
                              level at a frequency of 200 MHz and input
                              ATT of 10 dB within the range of 0 to 59 dBm \,
                              (R4131C/D) and 110 dBµ to 51 dBµ
                              (R4131CN/DN) in reference level.
    Unit of reference level: dBm, dB\mu, dB\mu/m, or dBmV, selectable
    Marker display
                                                 10 dB/DIV
      Resolution ..... 0.2 dB
```

2 dB/DIV

0.05 dB

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8.1 Technical Data of Function

Dynamic range

Average noise level

..... -110 dBm: R4131C/D

-108 dBm: R4131CN/DN

esolution band width 1 kHz, Video filter band

width

10 Hz, Input ATT 0 dB, More than 1 MHz in

frequency

Secondary/tertiary distortion

..... More than 70 dB

Where the input level is assumed to be -30 dBm and frequency to be more

than 1 MHz

Frequency response:

R4131C	100 kHz \leq F \leq 2 GHz ATT 10 dB or more \pm 1 dB or less		10 kHz \leq F \leq 3.5 GHz ATT 10 dB or more ±3.5 dB or less		
R4131D	100 kHz \leq F \leq 2 GHz ATT 10 dB or more ±1 dB or less		10 kHz \leq F \leq 3.5 GHz ATT 10 dB or more ±2 dB or less		
R4131CN/DN	100 kHz ≤ F ≤ 1.5 GHz ±1.5 dB or less			2 kHz \leq F \leq 3.5 GHz \pm 4 dB or less	

Residual response: -95 dBm or less:

R4131C/D

-93 dBm or less:

ATT 0 dB and input 50 Ω When terminated at input

When terminated at input

R4131CN/DN ATT 0 dB and input 75 Ω

Note: At frequency 100 kHz

Video filter band width:

1 MHz, 300 kHz, 100 kHz, 10 kHz, 1 kHz, 100 Hz, or

10 Hz

Resolution selecting accuracy

Gain compression: Less than 1 dB

: Less than ±1 dB

at +20°C to +30°C

at input of -10 dBm

R4131 SERIES SPECTRUM ANALYZER INSTRUCTION MANUAL

8.1 Technical Data of Function

(3) Sweep Specification

Sweep time : 5 ms/div to 100 s/div with 1-2-5 step

Sweep time accuracy

: Less than ±15%

Sweep trigger : FREE RUN, LINE, VIDEO, and SINGLE (Reset/Start)

(4) Input Specification

RF input : Approx. 50 N-type input connector: R4131C/D

Approx. 75 N-type input connector: R4131CN/DN

Maximum input level

: +20 dBm, ±25 VDCmax Input ATT 20 dB or more:

R4131C/D

127 dBu, ±25 VDCmax Input ATT 20 dB or more:

R4131CN/DN

Input ATT : 0 to 50 dB

Input ATT selecting accuracy

: ± 1 dB or less 10 kHz \leq F \leq 2 GHz

(10 dB in standard) 2 GHz < F \leq 3.5 GHz

with a step of 10 dB

(10 dB in standard)

Input VSWR R4131C/D

1.5 or less 2.0 or less

±1.5 dB or less

100 kHz \leq F \leq 2 GHz 2 GHz < F \leq 3.5 GHz

At input ATT 10 dB or

more

R4131CN/DN

1.5 or less 2.0 or less 2.5 or less 100 kHz \leq F \leq 1.5 GHz 10 kHz < F \leq 2 GHz 2 GHz < F \leq 3.5 GHz

At input ATT 10 dB or

more

(5) Display Unit Specification

Display : Waveform, setting conditions, and grid

Trace : 2-screen display of WRITE waveform and VIEW

waveform

WRITE : Memory is rewritten each time sweep and WRITE

waveform is displayed.

STORE : WRITE waveform is stored.

VIEW : Stored waveform data is displayed.

MAX. HOLD : Each time of repetition from the starting point of

this function, the maximum signal level on the

horizontal axis is measured and displayed.

Dictation : This equipment provides the POSI/NEGA (for

R4131D/DN only), POSI, and SAMPLE display and

detection functions.

(6) Output Specification

Output signal for calibration

200 MHz ±30 kHz, -30 dBm ±0.5 dB: R4131C/D 200 MHz ±30 kHz, 80 dBu ±0.5 dB: R4131CN/DN

Monitor output Possible to listen with an earphone (approx. 8 Ω) Recorder output

Analog output only for WRITE waveform

X-axis Approx. -5 V to +5 V (approx. 10 $k\Omega$) Y-axis Approx. 0 V to +4 V (approx. 220 Ω)

IF output Video output

The IF signal, 3.58 MHz, is output at approx. 50 Ω . This output includes the output terminal to

external CRT display and VIDEO plotter, etc., output impedance of approx. 75 Ω , 1 V_{p-p} , and

composite signal.

Probing power terminal ± 15 V

4-pin connector

GPIB data output: Mode operation and I/O are enabled using the GPIB. Plotter interface:

Display screen can be recorded by connecting this equipment directly to the plotter without passing

through the controller.

Output for TG:

1st LOCAL OUT -5 dBm or more Approx. 4 GHz to 7.5 GHz

2nd LOCAL OUT -5 dBm or more Approx. 3.77 GHz

SLOPE OUT; Sweep signal output for TG output level correction 2 V/GHz

(7) General Specifications

Using ambient conditions

: Less than 0°C to 50°C and 85% RH

Storage temperature range

-20 $^{\circ}$ C to +70 $^{\circ}$ C

Power supply 90 V to 132 V or 198 V to 250 V

48 to 66 Hz

Power consumption: Less than 120 VA

External dimensions

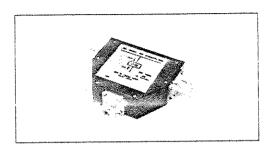
Approx. 300 (W) x 177 (H) x 460 (D) (mm)

Weight Approx. 10 kg : R4131C/CN

Approx. 10.5 kg: R4131D/DN

8.2 Accessories

● TR1625 RF Coupler



Frequency range Maximum input

Degree of coupling : 40 dB ±1 dB

Impedance

V.S.W.R Insertion loss

Connector

: DC-500 MHz

: 50 W

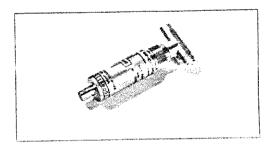
: 50 Ω in both main and auxiliary lines

: Less than 1.5 : Less than 1 dB

: Main line ... N-type for both main and auxiliary

lines

• TR1626 RF Coupler



Frequency range : DC-1500 MHz

Maximum input

Degree of coupling : 40 dB ±1 dB

Impedance V.S.W.R

Insertion loss

Connector

: 50 W

: 50 Ω in both main and auxiliary lines

: Less than 1.5

: Less than 1 dB

: Main line ... N-type, and auxiliary line ... BNC

type

BNCP-FJ Conversion Adaptor

Dielectric strength : 500 VAC/1 min.

Insulation resistance: More than 500 $k\Omega$ at 500 VDC

Contact resistance : Less than 5 $M\Omega$

V.S.W.R : Less than 1.2 at 0.1 GHz

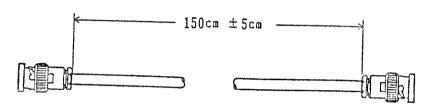
● Earphone for TR16191 Voice Monitor

When the FREQ SPAN is set to 0 (zero) and this spectrum analyzer is tuned with the data knob, the demodulation wave can be observed on the screen, but also listening can be done through the earphone connected to the phone.

Connection cables

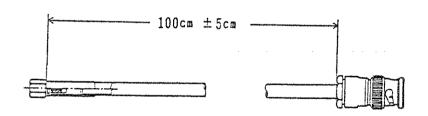
MO-15 Connection cable BNC-BNC $(75\,\Omega)$

Part code: DCB-FF0442



MC-37 Connection cable BNC-SMA

Part code: DCB-FF1130X01

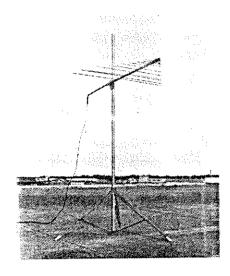


GPIB connection cable

Model name	Length
408JE-1P5	0.5 m
408JE-101	1 m
408JE-102	2 m
408JE-104	4 m

Antenna

● TR1711 Log-periodic Antenna



This is a brad band reception antenna of 8 to 1000 MHz in frequency range. It can be used for monitoring radio waves and for analyzing disturbing waves which occurs in wide bands.

Frequency range

Gain

Front-to-back ratio : More than 14 dB

V.S.W.R

I/O impedance

Weight

Components

80 MHz to 1000 MHz

5 dB (2/2 dipole antenna ratio)

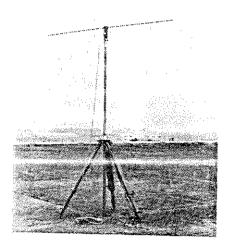
: Less than 2.5

Antenna main body ... Approx. 5 kg

Log-periodic antenna (Element 31 x 2, antenna main body, and balancer), angle adjuster (45° to 00 to 900), tripod, measuring scale (with N-type connector, 10 m), elements container box,

and antenna main body container bag

● TR1722 Half-wave Dipole Antenna



When measuring the field intensity and disturbing wave by using the spectrum analyzer, this antenna is used by changing the length of elements according to the measuring frequency.

Frequency range : 25 MHz to 1000 MHz

Element 1 ... 25 MHz to 80 MHz Element 2 ... 80 MHz to 250 MHz Element 3 ... 250 MHz to 600 MHz Element 4 ... 600 MHz to 1000 MHz

Transmission impedance

: 50 Ω

Polarization : Horizontal polarization/vertical polarization

selected

Antenna ground height: A

Approx. 1 to 4 m

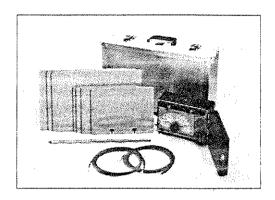
Tripod:

Folding type

Attached coaxial cable

: Attached with 50D, 2W, 10 m, and N-type connector

TR1720 Loop Antenna



Frequency range : 100 kHz to 30 MHz

Antenna tuner unit : 1-band ... 100 kHz to 200 kHz

2-band ... 150 kHz to 300 kHz 3-band ... 300 kHz to 600 kHz 4-band ... 600 kHz to 1400 kHz 5-band ... 1.4 MHz to 3.5 MHz 6-band ... 3.5 MHz to 10 MHz

7-band ... 10 MHz to 30 MHz

Loop antenna section: 7 types of loop antenna for 1-7 bands

Vertical antenna section

: Set to 2 m and 1 m in total length

Impedance : 75 Ω (TR1720N) or 50 Ω (TR1720)

Dimensions and weight:

Tuner unit : Approx. 210 (W) x 140 (H) x 110 (D) (mm); and

2 kg

Loop antenna : Approx. 3 kg in one set

Big) Approx. 360 (W) x 250 (H) x 6 (D) (mm) Small) Approx. 250 (W) x 190 (H) x 6 (D) (mm)

Vertical antenna : 2 m (5 stages in total length)

1 m (expansion and contraction) and 0.2 kg

Container case : Approx. 495 (W) x 290 (H) x 155 (D) (mm)

Aluminum made and approx. 1.9 kg in weight

● TR17201 10 kHz to 30 MHz Active Antenna

This is an antenna used for the measurement of field intensity from 10 kHz to 30 MHz. Since it integrates a low noise and broad band amplifier and the antenna factor is almost contact, the field intensity can be directly read easily.

Frequency range : 10 kHz to 30 MHz
Antenna factor : Approx. 10 to 13 dB

Output impedance : Approx. 50 Ω

Input impedance : More than 1 $M\Omega$ (when measured at the antenna

block)

Amplification gain : 7 dB ±2 dB in nominal gain

Connector : BNC type

Power supply : 12.6 V mercury cell (approx. 20 hours) External dimensions : Approx. 131 (L) x 108 (W) x 77 (H) (mm)

Weight : Approx. 1 kg

● TR17203 25 MHz to 230 MHz Active Dipole Antenna

Since the antenna factor for the measurement of field intensity from 25 MHz to 230 MHz is close to 0 (zero), this antenna can directly read the field intensity in a wide range when used in combination with the spectrum analyzer.

Frequency range : 25 MHz to 230 MHz Antenna factor : Approx. 0 dB Impedance : Approx. 50 Ω

Connecting terminal : N-type

Power supply : 15 VDC (with 1 m long cable)

Weight : Approx. 580 q

TR17204 200 MHz to 1000 MHz Log-periodic Antenna

The antenna can measure a broad band of 200 MHz to 1000 MHz without replacing any element. In addition to its compactness and lightweight, it can be used for transmission and reception. So, it is suitable for immunity measurement in high frequency.

Frequency range : 200 MHz to 1000 MHz

Antenna factor : Approx. 14 dB to 25 dB at 200 MHz to 1000 MHz

Impedance : Approx. 50 Ω

Connecting terminal : N-type

Average V.S.W.R. : Less than 2.0 Average gain : Approx. 7 dB

Antenna dimensions : Approx. 750 (length) x 750 (maximum width)

x 63.5 (thickness) (mm)

Weight : Approx. 2 kg

● TR17205 1 GHz to 10 GHz Log-spiral Antenna

This is an antenna of 1 $\,\mathrm{GHz}$ to 10 $\,\mathrm{GHz}$ which is used to measure $\,\mathrm{EMI}$ conformable to the $\,\mathrm{MIL}$ Standards.

Frequency range : 1 GHz to 10 GHz

Average power gain : 3.75 dB

Average V.S.W.R. : Less than 2.0 Axial ratio : Less than 1 dB

Average beam width : 500

Impedance : Approx. 50 Ω

Polarization : Circular polarization

External dimensions : Approx. 381 (length) x 127 (maximum diameter)

(mm)

Weight : Approx. 3.6 kg

● TR17206 1 GHz to 18 GHz Double-ridged Guide Antenna

This is the most suitable antenna for the EMI measurement. It can measure a wide band of 1 $\,\mathrm{GHz}$ to 18 $\,\mathrm{GHz}$.

Frequency range : 1 GHz to 18 GHz Average power gain : 10.7 dB (Isotropic)

Average V.S.W.R. : Less than 1.5 Impedance : Approx. 50 Ω Average beam width : E Plane 530

H Plane 480

Connector : N-type

External dimensions : Approx. 280 (L) x 245 (W) x 159 (H) (mm)

Weight : Approx. 1.8 kg

Filter

MEP-293/MEP-294/MEP-295/MEP-29, TR14101

Model	name	MEP-292	MEP-293	MEP-294	MEP-295	TR1410;
Filter	name	By-pass filter	By-pass filter	By-pass filter	By-pass filter	Rejection filter
cation	ive communi- equipment ncy band	27 мна	60 MHz	150 MHz	400 MHz	800 MHz to 900 MHz
Working range	frequency	26 MHz to 30 MHz	50 MHz to 80 MHz	120 MHz to 190 MHz	335 MHz to 520 MHz	800 MHz to 900 MHz
Filter Char- acter- istics	Cut-off frequency	40 MHz	100 MHz	240 MHz	670 MHz	1200 MHz
	Attenuation character- istic	More than 35 dB at 28 MHz or less More than 40 dB at 27 MHz	More than 50 dB at 70 MHz More than 30 dB at 80 MHz	More than 50 dB at 170 MHz More than 30 dB at 190 MHz	More than 50 dB at 470 MHz More than 70 dB at 520 MHz	More than 35 dB at 800 MHz to 900 MHz More than 30 dB at 800 MHz or less
	Pass band	40 MHz to 300 MHz	100 MHz to 1000 MHz	240 MHz to 1000 MHz	670 MHz to 1500 MHz	1500 MHz to 3000 MHz
	Insertion loss (within the pass band)	Less than 1 dB	Less than 2 dB	Less than 2 dB	Less than 2 dB	Less than 2 dB
char- acter- istics	Pass band	DC to 300 MHz	-	.		DC to 1000 MHz
	Insertion loss (within the pass band)	Less than 1 dB	-	· _	-	Less than 1 dB
Characte impedanc		50 Ω (BNCJ-BNCJ)	50 Ω (NP-NJ)	50 Ω (NP-NJ)	50 Ω (NP-NJ)	50 Ω (NP-NJ)

Band Pass Filter

TR14201/14202/14203/14204

This filter is used to remove the large signal out of a measurement band in the measurement conforming to the CISPR Standards using the spectrum analyzer.

	TR14201	TR14202	TR14203	TR14204
Pass band	10 kHz to 150 kHz	150 kHz to 30 MHz	25 MHz to 300 MHz	300 MHz to 1000 MHz
Insertion loss within the pass band	Less than 1.5 dB	Less than 1.5 dB	Less than 1.5 dB	Less than 1.5 dB
Attenuation characteristic		More than 35 dB at less than 30 kHz but more than 60 MHz	More than 35 dB at less than 12 MHz but more than 600 MHz	More than 30 dB at less than 150 MHz but more than 1500 MHz
Characteristic impedance (connector)	Approx. 50 Ω (NJ-NP)	Approx. 50 Ω (NJ-NP)	Approx. 50 Ω (NP-NJ)	Approx. 50 Ω (NP-NJ)

External dimensions: Approx. 31 (H) \times 50 (S) \times 100 (L) (mm)

Weight : Approx. 350 g

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R4131 SERIES SPECTRUM ANALYZER INSTRUCTION MANUAL

9. Functional description

9. FUNCTIONAL DESCRIPTION



9.1 Outline

9.1.1 Basic Operations

Figure 9-1 shows the block diagram of this equipment.

(1) When the measuring signal is input to the input connector, the input signal, after passing through the 50 dB RF input attenuator, enters the first mixer where it is mixed with the first local signal sent from the 4 to 7.5 GHz YTO (YIG tuning transmitter), and then it is output as the first IF signal of 4 GHz.

The YTO, under the control of the YTO circuit, sweeps the range of 4 to $7.5~\mathrm{GHz}$ using the RAMP signal and also varies the center frequency with the maximum resolution of $500~\mathrm{Hz}$.

(2) The output first IF signal of 4 GHz enters the second mixer where it is mixed with the second local signal of 3.77 GHz and then enters the third mixer as the second IF signal of 226 MHz. This signal is mixed with the third local signal of 200 MHz and then enters the fourth signal as the third IF signal of 26.4 MHz. This signal is further mixed with the fourth local signal of 30 MHz and converted into the fourth IF signal of 3.58 MHz.

Incidentally, the CAL OUT signal of 200 MHz is generated through the crystal oscillator of the third local signal.

- (3) The fourth IF signal of 3.58 MHz passes through the LC filter second stage and crystal filter second stage, through which the resolution band width is selected in a range from 1 MHz to 1 kHz, and further, the output level is controlled by the resolution of 0.25 dB max. by the STEP AMP. of 50 dB.
- (4) The 3.58 MHz IF signal of which resolution band width and output level are controlled enters the LOG AMP. of the dynamic range 80 dB, and after being subjected to logarithmic companding, the signal enters the detector where it is detected and converted into the DC output. The detection output signal enters video filter circuit where the video filter band width is selected to a range from 1 MHz to 10 Hz and then output as the Y. OUT signal.
- (5) The Y. OUT signal and the X. OUT signal of the RAMP signal are both input to the A/D circuit. The Y. OUT (ordinates axis) is converted from analog to digital signal at 9 bits (512 points) and the X. OUT (quadrature axis) is converted the similarly at 10 bits (1024 points). After being stored in the memory, these signals are controlled by the CPU to display the waveform on CRT through the CRT control circuit.

This equipment has two memories, the WRITE memory which rewrites data at each sweeping and VIEW memory which stores the displayed waveform. It also has a non-volatile memory which stores data even after power OFF.

Furthermore, it performs the MAX. HOLD and normalization processing using the WRITE memory, VIEW memory, and the CPU's arithmetic operation function.

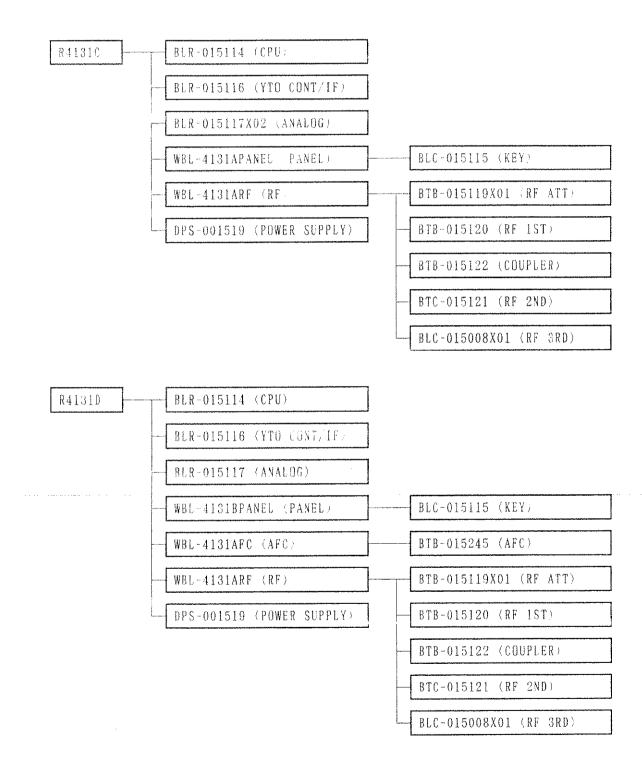
(6) The AFC (Automatic Frequency Control) block is mounted on R4131D/DN only. It applies locking in a range from 4 to 6.5 GHz in the YTO frequency to improve the center frequency setting accuracy.

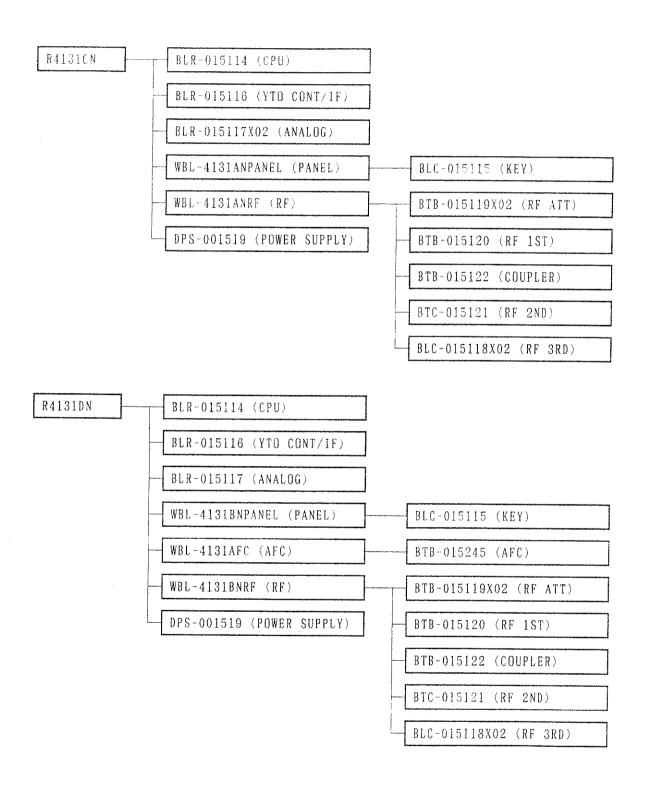
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R4131 SERIES SPECTRUM ANALYZER INSTRUCTION MANUAL



9.1.2 R4131 Series Configuration





9.2 RF Block

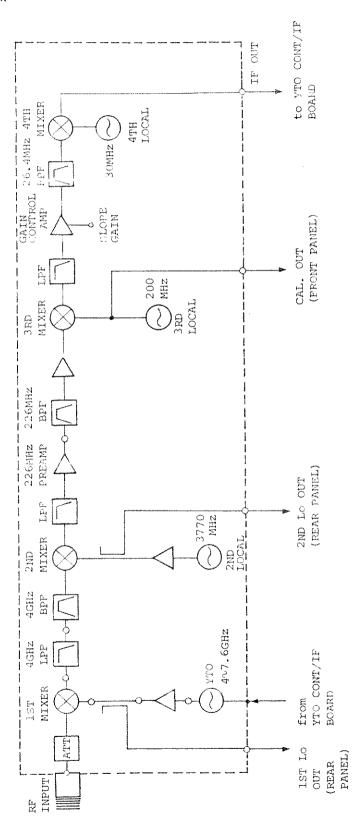


Figure 9-2 RF Block

9.2.1 First Mixer

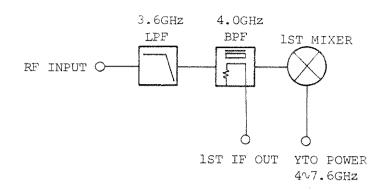


Figure 9-3 First Mixer Block Diagram

(1) 3.6 GHz Low-pass Filter

The 3.6 GHz low-pass filter limits the input frequency band.

(2) 4.0 GHz Band Pass Filter

The $4.0~\mathrm{GHz}$ band pass filter passes only $4~\mathrm{GHz}$ frequency signals of the first IF signals generated by the first mixer.

(3) First Mixer

The first mixer is single-balanced type. It has two ports: one mixes the RF input signals and IF output signals which are isolated by the LPF and BPF in the previous stage.

9.2.2 Second Mixer

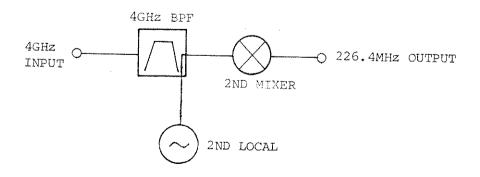


Figure 9-4 Second Mixer Block Diagram

(1) 4.0 GHz Band Pass Filter

The 4.0 GHz band pass filter consists of two dielectric resonators.

(2) Second Local Oscillator

The second local oscillator using a dielectric resonator oscillates the 3770 MHz frequency.

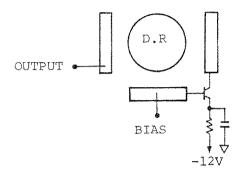


Figure 9-5 Second Local Oscillator

(3) Second Mixer

The second mixer converts the first IF signals (4 GHz) to the second IF signals (226.4 MHz).

9.2.3 Third and Fourth Mixers

The second IF signals (226.4 MHz) are converted to 26.4 MHz (third IF signals) by the third mixer and further converted to 3.58 MHz by the fourth mixer.

The third local oscillator signal is also used as a CAL.OUT signal.

The third IF signal uses a slope signal from the YTO-CONT/IF board to correct the frequency characteristics.

(1) 226.42 MHz Preamplifier

The 226.42 MHz preamplifier has a gain of 20 dB. L3, L4, and C9 are input matching filters. L5, L6, and C13 are output matching filters.

(2) Third Mixer

The third mixer is designed so that it does not input signals outside the band by using the 226.42 MHz BPF. The BPF band width is 4 MHz.

The BPF output is input to the isolation amplifier (Q1) and mixed with 200 MHz signals from the third local oscillator by the third mixer, then converted to $26.4 \, \text{MHz}$. The third mixer is a double-balanced type.

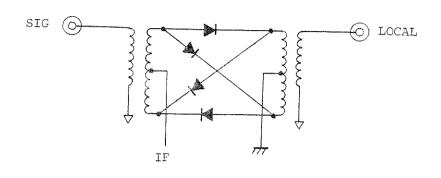


Figure 9-6 Double-balanced Mixer

(3) 200 MHz Crystal Oscillator

The base-ground Colpitts 200 MHz crystal oscillator oscillates a 200 MHz signal. It also oscillates a CAL.OUT signal (200 MHz, $-30~\mathrm{dBm}$).

(4) Gain Control Amplifier

The gain control amplifier changes the resistance of the Q1 emitter and collector to convert the amplifier gain.

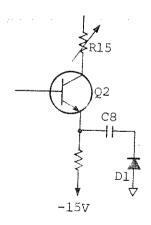


Figure 9-7 Gain Control Amplifier

As the current flowing through the pin diode D1 changes, the resistance changes. Using this characteristic, the gain control amplifier corrects the level. D1 uses a Slope Gain signal to correct the frequency characteristics.

L9 and R20 build a 50-ohm wide band matching circuit so that the gain control amplifier does not affect the $26.4~\mathrm{MHz}$ BPF in later stages.

The 26.4 MHz band pass filter consists of four nelical resonators. The circuit converts the signal frequency to 3.58 MHz by the fourth mixer in the next stage. The double-balanced fourth mixer mixes signals by using a 30 MHz signal generated by the fourth local oscillator.

(5) 30 MHz Crystal Oscillator

The Colpitts 30 MHz crystal oscillator oscillates a 30 MHz local signal. The circuit outputs the signal via a tank circuit (C30 and L13) so that it is not changed by the load.

9.3 YTO-CONT/IF Board

9.3.1 IF Filter

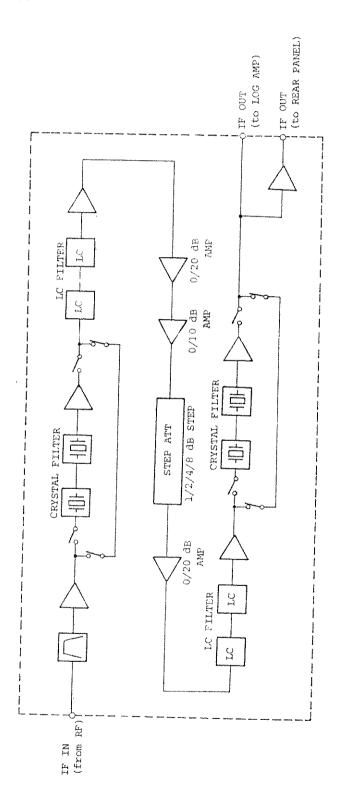


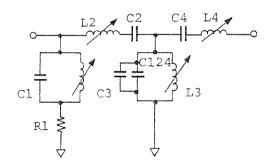
Figure 9-8 IF Filter

The IF filter consists of filters having the resolution bandwidth.

The bandwidth of the filter can be switched by the center frequency of 3.58 MHz according to the setting from the front panel. The filter with narrow bandwidths (1 kHz and 3 kHz) uses four crystal filters; the filter with other bandwidths (1 MHz to 10 kHz) uses four LC filters.

(1) Input 3.58 MHz Band Pass Filter

L2, L3, L4, C2, C3, C4, and C124 form a 3.58 MHz BPF. L1, C1, and R1 form a wide-band impedance matching circuit.



(2) Gain Adjust Amplifier

The gain adjust amplifier is non-inverse type. The circuit changes the total gain by adjusting the variable resistor (AMPTD_CAL) on the front panel.

AMPTD_CAL is used to change the resistance using the FET (Q1) to change the total gain.

R6 is a thermister. It compensates the gain changed by the temperature.

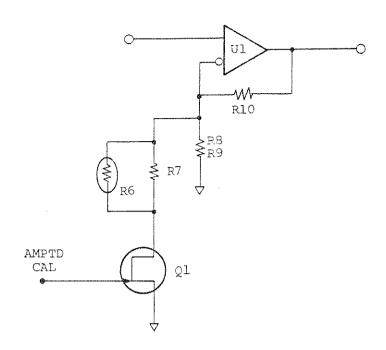


Figure 9-9 Gain Adjust Amplifier

When Q1 = OFF

$$G = 1 + \frac{R10}{R8 + R9} = 1 + \frac{470}{120} = 4.92$$

$$G (dB) = 20 LogG = 14 (dB)$$

When Q1 = ON (10 ohms)

$$G = 1 + \frac{R10}{RT} = 1 + \frac{470}{44.2} = 11.63$$

$$G (dB) = 20 LogG = 21 dB$$

Note: RT is the resistance of R6 to R9 and Q1.

(3) Crystal Filter

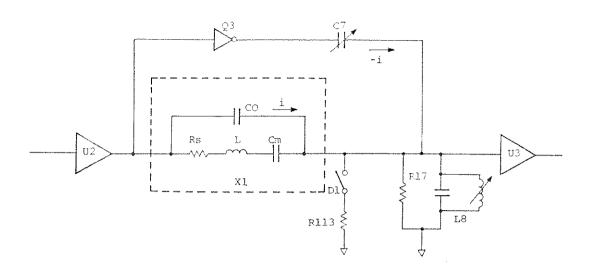


Figure 9-10 Crystal Filter

The bandwidth is selectable with the switch (D1): 1 kHz or 3 kHz. C7 adjusts the symmetry of the filter.

(4) LC Filter

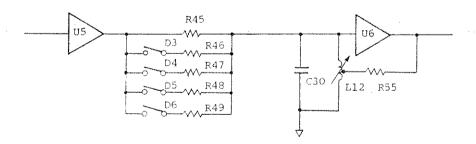


Figure 9-11 LC Filter

The bandwidth is changeable from 10 kHz to 1 MHz by switching the R45 to R49. The bandwidth is narrower as the resistance is larger.

(5) Step Amplifier

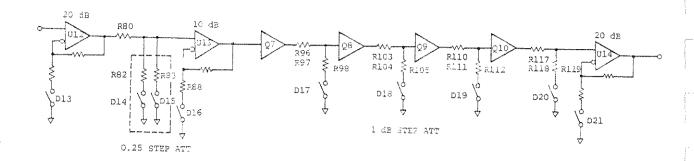


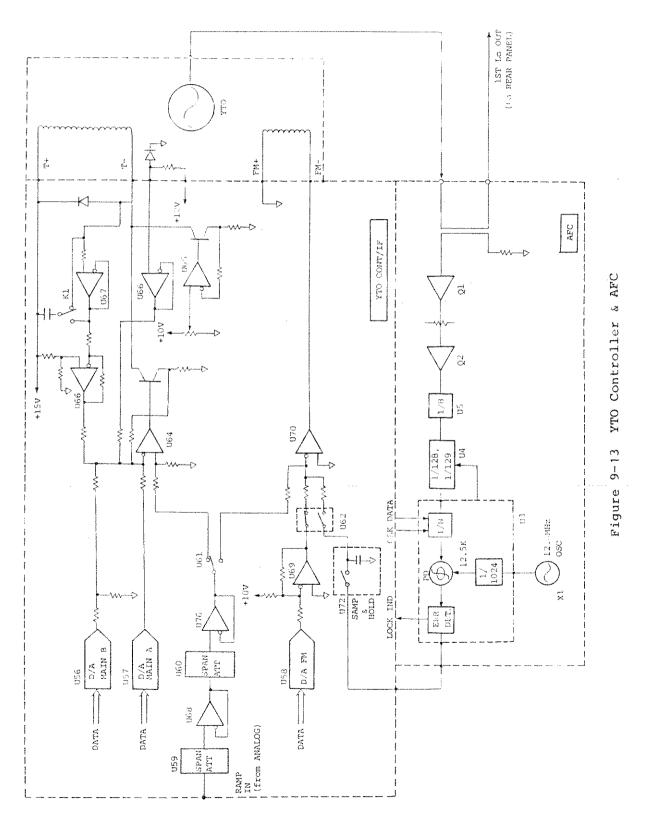
Figure 9-12 Step Amplifier

The step amplifier consists of three step amplifiers (U12 to U14), four 1 dB step attenuators (Q7 to Q10), and a 0.25 dB step attenuator.

U12 and U14 are $0/20~\mathrm{dB}$ step amplifiers and U13 is a $0/10~\mathrm{dB}$ step amplifier.

These step amplifiers and attenuators set the level by steps of 0.25 dB in the range from 0 dB to 59.75 dB.

9.3.2 YTO Controller and AFC



(1) YTO Controller

The YTO controller consists of a controller and a driver.

The tune voltage changes depending on the set center frequency. The YTO controller sets three digital/analog frequency bands and generates a tune voltage by a combination of the three bands. The three D/A converters have different setting ranges.

Tune D/A	Input data	Cent, freq, data	Freq, span
MAIN A (U57)	32 to DE _H	0 to 3.5 GHz	20 MHz to 4.0 GHz
MAIN B (U56)	00 to F9 _H	△ 25.6 MHz	
FM (U58)	00 to F9 _H	△ 128 kHz	100 kHz to 10 MHz

Table 9-1 Tune Voltage Data

For the span voltage, the YTO controller converts the ramp voltage from the ramp generator of the analog board for setting a span by two step attenuators and adds it by the tune voltage in the U64. When the span voltage reaches 10 MHz, a relay (K1) is switched and a noise filter (large-capacity chemical capacitor) is inserted between the main coils. If a charged or discharged current flows through the capacitor, however, the current flowing through the main coil changes, causing a frequency drift. To solve this problem, a charger/discharger is added to charge or discharge at the main T- (See Figure 9-3) even if the noise filter is turned off.

The frequency may also drift because of temperature change. The YTO controller corrects the frequency by the following two methods:

1 Feeds back the voltages at the both ends of the main coil.

When the current flowing through the main coil is increased or decreased to change the YTO oscillation frequency, the temperature inside the YTO controller changes and causes a frequency drift. Temperature change also causes the main coil resistance. The resistance change can be canceled by feeding back the voltages at both ends of the coil.

2 Mounts a diode inside the YTO controller and feed back the on-voltage change of the diode to the U64. As the ambient temperature changes, the on-voltage of the diode changes.

Using the above two circuits, the YTO controller reduces frequency drifts without the PLL.

(2) AFC

The AFC mounted on R4131D/DN operates at the frequency span of 200 MHz or smaller and applies AFC to the YTO. The AFC function is available in the band from 0 to $2.5~\mathrm{GHz}$.

The YTO output (4.0 to 6.5 GHz) is input to the AFC block and converted to the 500 MHz to 812.5 MHz range by the 1/8 divider.

Then, it is compared with the 12.8 MHz oscillation signal by the phase detector and fed back to the tune FM voltage. At this time, if a fault is found in the phase detector output, a pulse is output to the LOCK IND signal line.

The AFC function is executed between sweeps. During AFC, the span is set to 0 and the SAMP/HOLD circuit is closed. It opens when a sweep starts.

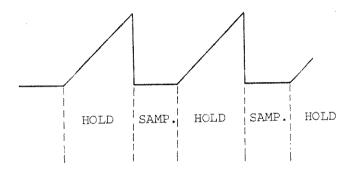


Figure 9-14 SAMP & HOLD

AFC operation sequence is shown below.

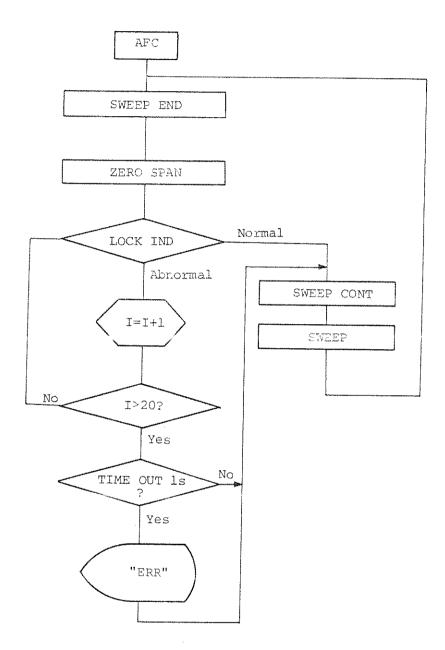


Figure 9-15 Flowchart for AFC

9.4 Analog Board

9.4.1 Log Amplifier

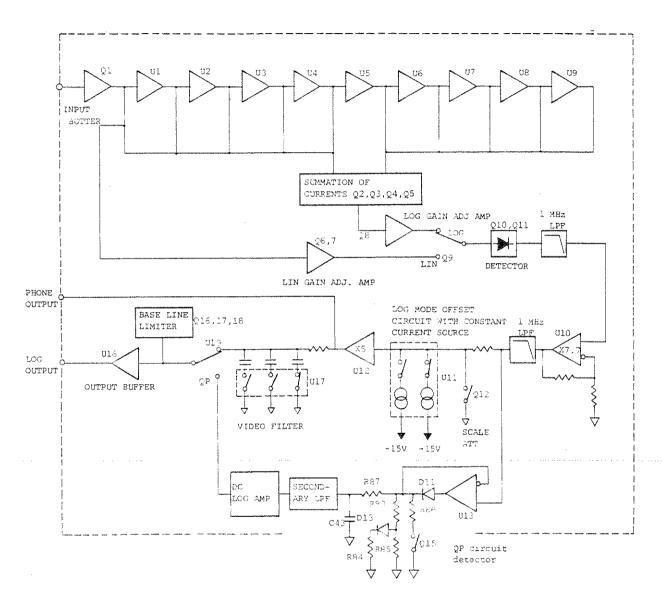
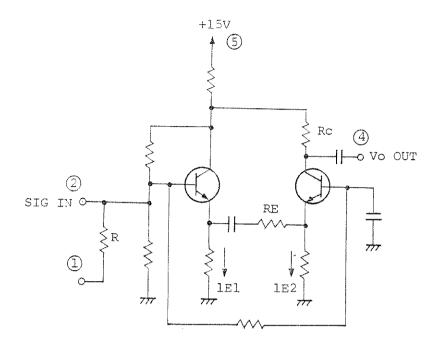


Figure 9-16 Log Amplifier Schematic Diagram

The log amplifier consists of nine saturation amplifiers: each has a gain of 10 dB.

Figure 9-17 shows the saturation amplifier.



$$Gain = 20 Log \frac{R_C}{R_E}$$

 $Vsat p-p = R_C \times (I_{E1} + I_{E2})$

Figure 9-17 One Stage of 10 dB Amplifier

A signal from the IF block is input to the input buffer (Q1) then to the saturation amplifier. $V_{\rm O~OUT}$ is converted to the current $V_{\rm O}/R$ and input to the current amplifier.

To amplify the current, base-ground amplifiers Q3 and Q4 are used with Q2 and Q3, just as for the bias constant current source.

The current amplified by the base-ground amplifier is converted to the voltage by the R19.

When a 3 Vp-p signal is input to the input buffer (Q1), the 10 dB saturation amplifier output is all 3 Vp-p.

The current amplifier output is found as shown below.

$$V_{\rm I} = (3/0.62R + 9 \times 3/R) \times R19$$

Assume that $3/R \times R13 = V$.

$$V_{I} = 10.56 \text{ V}$$

When the input level decreases by 10 dB, the following voltage is output:

$$V_T(-10) = (3/3.16 \times 1/0.62R + 9 \times 3/R) \times R13 = 9.49 V$$

Similarly,

$$V_{I}(-20) = (\frac{1}{10} \times \frac{1}{0.62} + \frac{1}{3.16} + 8) V = 8.47 V$$

$$V_{I}(-30) = (\frac{1}{100} \times \frac{1}{0.62} + \frac{1}{10} + \frac{1}{3.16} + 7) V = 7.43 V$$

$$V_T(-80) = (\frac{1}{100} + \frac{1}{10} + \frac{1}{3.16} + 2) V = 2.43 V$$

As shown above, if the input level changes by 10 dB, the output level changes by approximately 1 dB.

The current amplified by the log gain adjust amplifier (Q8) is sent to the base-ground amplifiers (Q10 and Q11) and shaped to half waves for detection. The output is input to the x7.7 amplifier via the LPF, then to the scale attenuator or QP circuit via the 1 MHz LPF.

The scale attenuator sets the vertical axis mode (10 dB/div., $2 \, dB/div.$) by switching the Q12 on/off.

The U11 is a constant current source used to set the offset in logarithms. It is switched according to the horizontal axis mode selected.

The QP circuit detects an envelope by a detector consisting of the U13 and D13 and a discharger consisting of the R84 to R87, D13, and C43.

The D13 and C84 change for each time constant when repetitive frequency goes high or low.

The Q15 is turned off when the bandwidth is 120 kHz and on when it is 9 kHz.

Signals detected by the QP circuit is input to the LPF then to the DC log amplifier consisting of the U15 and U17.

The LOG or LIN/QP modes is set by the switch consisting of the U19 and output via the U16 and output buffer.

9.4.2 Ramp Generator

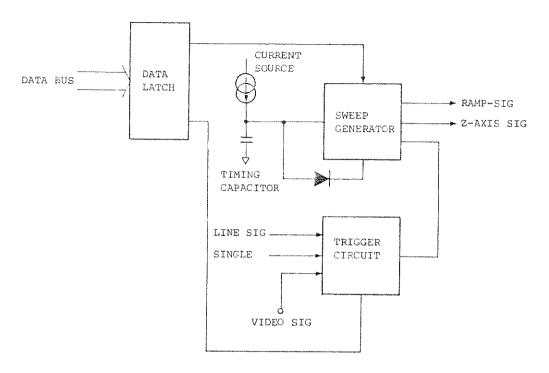


Figure 9-18 Block Diagram

The ramp generator generates a ramp voltage from approximately -5 V to +5 V which is used to sweep the YTO (first local oscillator). The ramp voltage is also used as X-axis data by the A/D converter.

The ramp generator also generates a Z-axis signal which is used to reset the X-axis A/D converter.

The constant current generated from the current source of the ramp generator is applied to the timing capacitor and generates the ramp voltage.

(1) Current Source

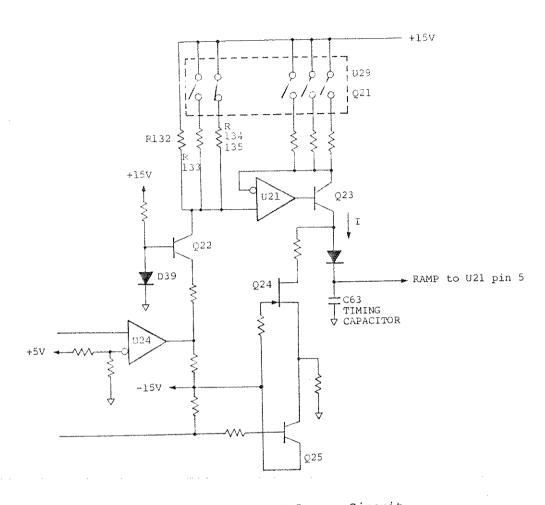


Figure 9-19 Current Source Circuit

The Q22 is a current source that is used to determine the voltage of the U21, pin 5. The voltage is used to correct the temperature of $V_{\rm BE}$ of the Q22.

The voltage of the U21, pin 5 is determined by a combination of the R132 to R135. After the voltage is determined, the emitter current of the Q23 flows until the voltage of the U21, pin 5 is the same as that of the U21, pin 6. The Q23 emitter current is controlled by a combination of the switches (U29 and Q21).

The Q23 collector current is the same as the emitter current because the Q23 current amplifier ratio (hfe) is large.

The constant current determined by the switches (U19 and Q21) flows through the timing capacitor (C63), and then generates a ramp voltage. $V=\frac{1}{C}$ It.

The Q24 and Q25 form a sweep stop controller. When a +5 signal is applied to the base of the Q25, the Q24 and Q25 are switched on and all currents flowing through the C63 flow through the Q24 and Q25. At this time, the ramp voltage is in hold state.

(2) Ramp Generator

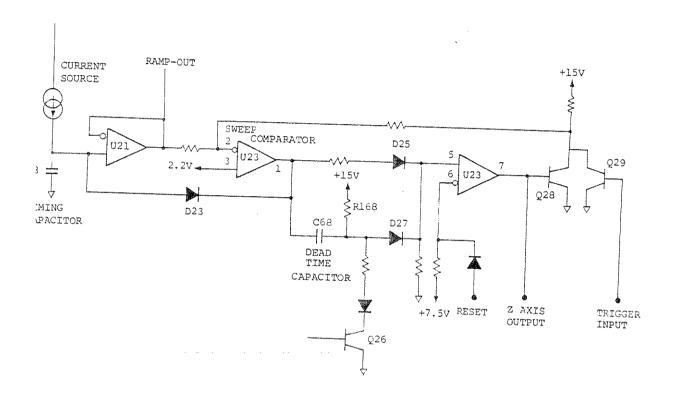


Figure 9-20 Ramp Generator

The ramp voltage from the C63 is input to the sweep comparator U23, pin 2. When the ramp voltage is low, the U23, pin 7 is \pm 15 V and the Q28 is switched on.

When the ramp voltage increases, the voltage of the U23, pin 2 reaches 2.2 V. In other words, when the ramp voltage is 6 V, the U23, pin 1 is inverted and the D25 is switched off. Along with this change, the anode voltage of the D27 also changes via the dead time capacitor. Then, the voltage of the U23, pin 7 becomes -15 V and the Q28 is switched off.

At the same time, the U23, pin 1 is -15 V, the D23 is switched on, and the voltage charged by the C63 is discharged.

When the ramp voltage reaches -6 V, the U23, D23, and U21 form a close loop to keep -6 V. The dead time capacitor (C68) is charged by the R168 because the D27 anode voltage increases. When the voltage of the U23, pin 5 exceeds 7.5 V, the U23, pin 7 becomes +15 V and the Q18 is switched on.

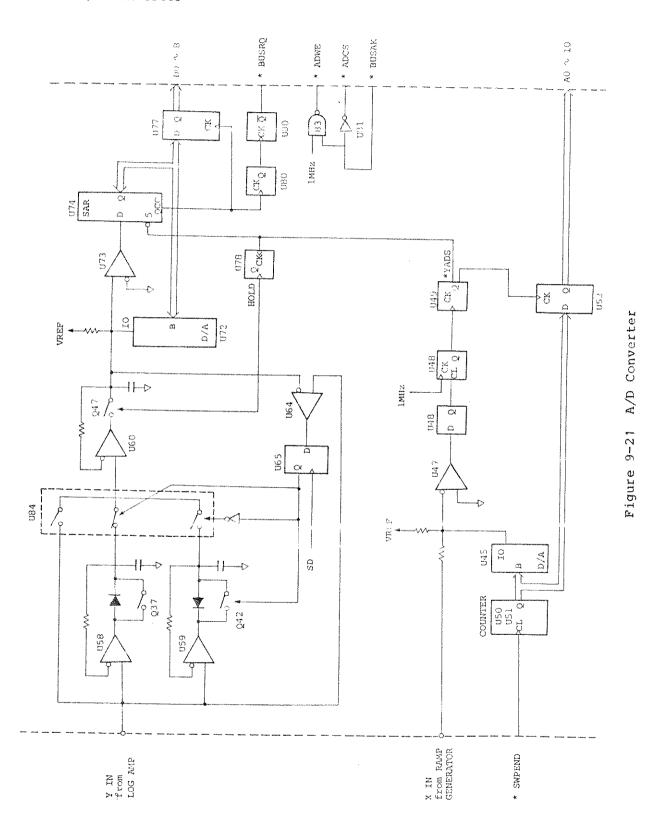
This changes the voltage of the U23, pin 2 and the voltage of the U23, pin 1 to \pm 15 V. The D23 is switched off then the timing capacitor starts charging.

Thus, the ramp generator generates a ramp voltage.

The dead time of the ramp voltage is determined by the R168 and C68. The Q26 is switched on when the trigger mode is set to line, video, or single. Then the D27 anode voltage is set to 7.5 V or less. When the ramp voltage reaches 6 V, the U23, pin 1 is inverted and the Q28 is switched off. When it reaches -6 V, the U23, pin 1 is kept constant.

If the Q29 is switched on by a trigger signal, the voltage of the U23, pin ! becomes +15 V and the D23 is switched off. Then, the timing capacitor C63 starts charging and a ramp voltage is generated.

9.4.3 A/D Converter



(1) X-axis A/D Converter

The X-axis A/D converter compares the voltage generated by the ramp generator and outputs data from the counter with the D/A converted value. The comparator U47 is inverted when the difference between the current generated by VREF and the current generated by the sweep voltage match the current generated by D/A converter input data. At this time the converter latches the counter and at the same time starts Y-axis A/D by *YADS.

(2) Y-axis A/D Converter

The Y-axis A/D converter converts data analog to digital via the peak detector by the successive approximation for display data.

The peak detector mode is selectable using the input waveform: POSI or NEGA.

When a *YADS signal is input to the U74 from the X-axis A/D converter, the Y-axis A/D converter starts Y-axis A/D conversion and outputs QCC from the SAR (U74) successive comparator. Then, converted Y-axis data is latched by the U77.

The converter issues *BUSRQ to the CPU board. When receiving a *BUSAK signal from the board, it selects the fresh memory on the CPU board by *ADCS and transfers it from the A/D board to the CPU board by a direct memory access (DMA).

When a *ADCS signal is input to the OE terminal of the X-axis and Y-axis latch circuits (U52 and U77), the A/D converter is set to the output mode.

Analyzer Test

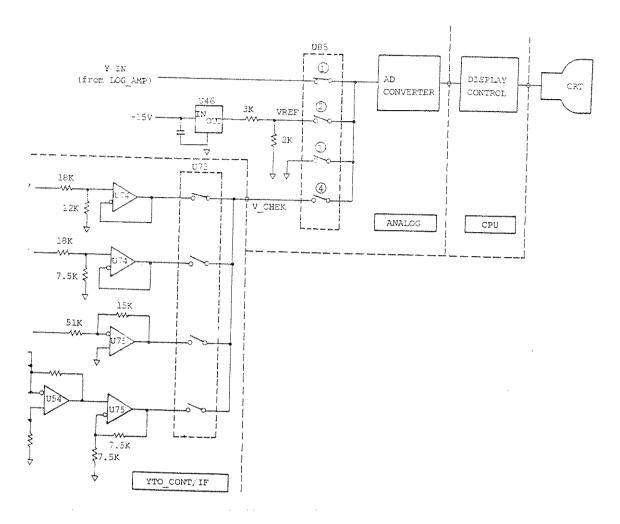


Figure 9-22 Analyzer Test

ne R4131 Series has an adjustment function on the screen display. It enerates a stable reference voltage and divides it into the 4 V eference voltage. The output is sent to the A/D converter and isplayed on the top of the scale. The A/D gain can be adjusted by 1e 4 V power without DVM. The operator simply aligns the displayed ine on the top of the scale. Similarly, adjust the A/D offset by etting the U85 switch to ③ (Figure 9-22) so that the displayed line; on the bottom of the scale.

en the U85 switch is set to 4, the three power sources and slope in of the YTO CONT/IF board can be tested.

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9.4 Analog Boar	ď	Boa	E	100	na	Aı	4	9.
-----------------	---	-----	---	-----	----	----	---	----

These operations can be set by keys. To start the analyzer test mode, press as follows:

The screen shown below appears.

+4	V											
AN	ALYZ	ER 1	EST	:	#	Y,OF	F					
						Y,GA	II	1				
				Ŀ		REF,	+1	3.	5.	V		
				:	ļ	REF,	- 1	3.	5	V	į	V
				:		REF,	+ ;	10	V-			
				:		SLOI	E	0	V			
				:		SLOI	E	2	V/	/GI	Ηz	
		QI	JIT	•	U	NIT						
O	Λ											

Figure 9-23 Analyzer Test Display

Move the mark "#" to the item to be tested with the \bigcirc and \bigcirc keys.

R4131 SERIES SPECTRUM ANALYZER INSTRUCTION MANUAL

10. Calibration and Adjustments

10. CALIBRATION AND ADJUSTMENTS

This section describes the procedures for making basic checks on the R4131 and for calibrating them after performance testing.



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R4131 SERIES SPECTRUM ANALYZER INSTRUCTION MANUAL

10. Calibration and Adjustments

10. CALIBRATION AND ADJUSTMENTS

This section describes the procedures for making basic checks on the R4131 and for calibrating them after performance testing.



10.1 Preparation

Table 10-1 lists the equipment and tools required for calibration and adjustment. Use equipment and tools equivalent or superior in performance to these.

Table 10-1 Equipment and Tools Required for Calibration and Adjustment

Equipment	Performance	Recommended equipment
Digital voltmeter	Range : $\pm 1000 \text{ V}$ Accuracy : $\pm 0.1\%$ Input impedance: $10 \text{ M}\Omega$	TR6846 (Voltage adjustment)
Synthesized signal generator	Frequency range : Frequency accuracy:	TR4511 Adjustment for YTO CONT/IF
10 dB step attenuator	Frequency range: DC to 500 MHz Variable: 0 to 80 dB or more Accuracy: ±0.5 dB or less	Adjustment for LOG AMP
1 dB step attenuator	Frequency range: DC to 500 MHz Variable: 0 to 10 or more Accuracy: ±0.2 dB or less	Adjustment for LOG AMP
Spectrum analyzer	Frequency range : 10 MHz to 4 GHz Frequency accuracy: +100 kHz	R4136 Adjustment for RF
Spectrum analyzer	Frequency range : 10 Hz to 120 MHz Tracking generator output: 10 Hz to 120 MHz T.G. output flatness : ± 1 dB Impedance : 50 Ω and 1 M Ω	TR4171 or R4136 + TR4154 Adjustment for IF FILTER

Table 10-2 Maintenance Tools Required for Calibration and Adjustment

Product name	Stock number	Remarks
Cable (SMA-SMA)	MM-14	
Cable (BNC-UM)	MC-36	2 pcs.
Cable (BNC-BNC)	MI-02	
UM to UM linear adapter	JCF-AC001JX07-1	UM-QA-JJ

1	0		1	P	r	e	g.	a	r	a	t	i	0	ĭ	7
---	---	--	---	---	---	---	----	---	---	---	---	---	---	---	---

(1)	Notes on Adjustment
	Before adjustment, performs the following operations:
1	Before setting the Power switch to OFF, press and .
	This operation sets correct data set by the CPU to zeros when ZERO CAL is executed.
	Corrected data is not erased even if the power is switched off. To reset correction, press these keys again.
2	Adjust the R4131D/DN having the AFC function as follows:
	- Set the Power switch to ON while the key is pressed down.
	- The message "strike any key" appears on the screen.
	- Press the key and the following screen appears:
	<type>:#R4131C (50) R4131D (50) (AFC) R4131CN (75) R4131DN (75) (AFC) <option>: OBW ON</option></type>
-	The Ally
	- Move the mark "#" to the R4131C or R4131CN with the keys.
	R4131D → R4131C R4131DN → R4131CN
	- Press the key.
	- Adjust the values.
	- Return setting to the original type.
	R4131C → R4131D R4131CN → R4131DN

SPECTRUM ANALYZER INSTRUCTION MANUAL

10.2 A/D Adjustment (Analog Board)

(BLR-015117)

10.2 A/D Adjustment (Analog Board) (BLR-015117)

- $\widehat{\text{(1)}}$ Measure the voltage between the TP19 and TP1 (GND) by the DMM and remember the measured value (V_{TP19}) .
- 2 Adjust the variable resistors so that the voltages of the TP20, TP21, and TP22 are as shown in Table 10-3. (This adjustment is available for the R4131D/DN only.)

Table 10-3 TP20, TP21, TP22 Voltage Adjustment Values

TP	Voltage	V R
TP20	V _{TP19} ± 10 mV	R241
TP21		R258
TP22		R277

- \bigcirc Press $\boxed{}$, $\boxed{}$, and $\boxed{}$.
- 4 The following data appears on the screen display:

+4	Λ						T		j	•		
AN	ALYZ	ER	TEST	:	#	Y,01	F	••••	 			-
		***************************************		:		Y,G	II	7	-		1	_
				:		REF	+	13.	5	V		
				:		REF,	 .	13.	. 5	V		
			+			REF,	+	10	V			
			<u> </u>	:		SLO						
				:	ĺ	SLOE	Ε	2	V/	/GI	Ηz	1
		Q	UIT	:	U	TIK						
0	v							i				

- 5 Move the mark "#" to Y.OFF with the \bigcirc and keys.
- 6 Adjust the R308 so that the displayed line aligns with the bottom line on the scale.
- Similarly, move the mark "#" to Y.GAIN with the and keys.

R4131 SERIES SPECTRUM ANALYZER INSTRUCTION MANUAL

10.2 A/D Adjustment (Analog Board)
(BLR-015117)

8	Adjust the R310 so that the displayed line aligns with the top line on the scale.
9	Press the key to initialize the R4131.
10)	Set the local feed-through to the center of the screen at the span 20 MHz.
11)	If the local feed-through is not at the center when the span is returned to 4 GHz, adjust the R233 so that it comes to the center. (X-axis and position adjustment)
12	Set the local feed-through at the center of the screen and change the span to 1 MHz and RBW to 30 kHz.
13)	Set the display detection mode to POSI with the and okeys.
14)	Adjust the R296 so that the waveforms are smoothed.
15)	Set the display detection mode from POSI to NEGA with the and keys.
(16)	Adjust the R302 so that waveforms are smoothed.

10.3 LOG Amplifier Adjustment (Analog Board)
(BLR-015117)

- 10.3 LOG Amplifier Adjustment (Analog Board) (BLR-015117)
 - 1 Disconnect the UM cable from the J4 and press and and to set the X-axis to the linear mode.
 - 2 Adjust the R57 and R72 so that voltage of the TP13 and TP14 is within ±1 mV.

	Voltage	٧R
TP.13	±1 mV	R57
TP.25		R72

- 3 Connect the log amplifier as shown in Figure 10-1.
- (4) Set the signal generator as follows:

Frequency: 3.5789 MHz Amplitude: -1 dBm

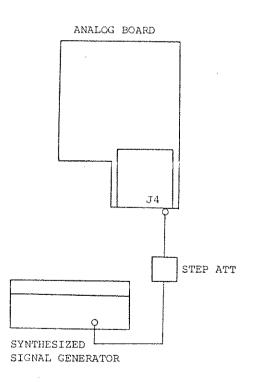


Figure 10-1 Log Amplifier Adjustment

R4131 SERIES SPECTRUM ANALYZER INSTRUCTION MANUAL

10.3 LOG Amplifier Adjustment (Analog Board)

(BLR-015117)

(5) Set the R4131 as follows:

Frequency span: 1 GHz 10 dB/DIV

- (6) Set the step attenuator to 0 dB.
- (7) Adjust the R40 so that the waveform aligns with the top line on the scale.
- (8) Set the step attenuator to 70 dB.
- Adjust the R69 so that the waveform aligns with the second line from the bottom on the scale.
- (10) Repeat steps (6) to (9).
- (11) Set the R4131 to 2 dB/div.
- (12) Set the step attenuator to 0 dB.
- (3) Adjust the R65 so that the waveform aligns with the top line on the scale.
- (14) Set the R4131 to LINEAR.
- (5) Adjust the R38 so that the waveform aligns with the top line on the scale.
- (16) Set the R4131 to QP.
- Adjust the R109 so that the waveform aligns with the top line on the scale.
- (18) Set the step attenuator to 20 dB.
- (19) Adjust the R102 so that the waveform aligns with the middle line on the scale.
- (20) Set the step attenuator to 35 dB.
- (2) Adjust the R96 so that the waveform aligns with the second line from the bottom on the scale.
- (22) Repeat steps (17) to (21).

10.4 IF Filter Adjustment (YTO-CONT/IF Board)

10.4.1 3.58 MHz BPF Adjustment

(1) Set the TR4171 as follows:

INPUT IMPEDANCE : 1 $M\Omega$

MAG mode

CENTER FREQ. 3.5795 MHz

FREQ. SPAN 5 MHz REF. LEVEL : -30 dBm TG LEVEL : -10 dBm

1 dB/DIV.

(2) Connect the units as shown in Figure 10-2.

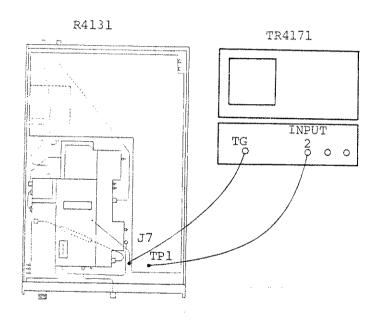


Figure 10-2 3.58 MHz BPF Adjustment

3 Turn the core of the L1 to L4 to adjust the waveform so that its peak is at 3.5789 MHz.

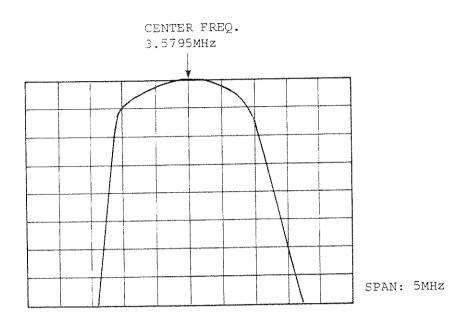


Figure 10-3 Waveform of 3.58 MHz BPF

10.4.2 Crystal Filter Adjustment

 \bigcirc 1) Connect the units as shown in Figure 10-4.

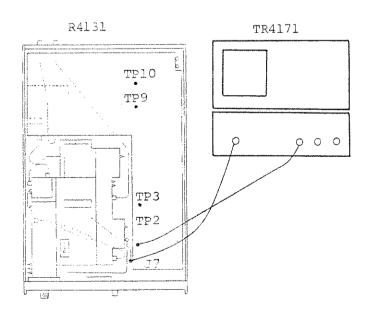


Figure 10-4 Crystal Filter Adjustment

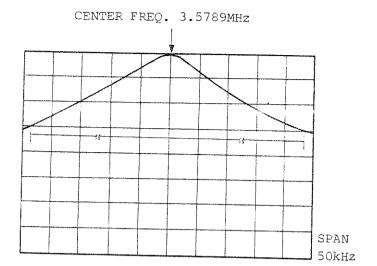


Figure 10-5 Waveform of Crystal Filter

(2) Set the TR4171 as follows:

CENTER FREQ.: 3.5795 MHz FREQ. SPAN : 50 kHz

10 dB/DIV.

(3) Set the R4131 as follows:

RBW: 3 kHz

- 4 Connect the TP1 with the INPUT2 of the TR4:71 and adjust the C9 so that the waveform is symmetrical. Then adjust the L8 so that the peak of the waveform is at its lowest level.
- (5) Connect the TP2 with the INPUT2 of the TR4171 and adjust the C18 so that the waveform is symmetrical. Then adjust the L10 so that the peak of the waveform is at its lowest level.
- 6 Press RBW, , , and set the R4131 as follows:

RBW: <u>QP</u> BW: <u>9 kHz</u>

- 7 Connect the TP9 with the INPUT2 of the TR4171 and adjust the C99 so that the waveform is symmetrical. Adjust the L27 so that the peak of the waveform is at its lowest level.
- (8) Connect the TP10 with the INPUT2 of the TR4171 and adjust the C108 so that the waveform is symmetrical. Adjust the L28 so that the peak of the waveform is at its lowest level.

(9) Adjust the L29 so that the waveform is at its maximum size.

10.4.3 LC Filter Adjustment

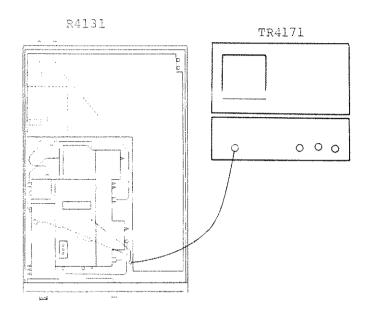


Figure 10-6 LC Filter Adjustment

(1) Set the TR4171 as follows:

CENTER FREQ.: 3.5789 MHz FREQ. SPAN : 100 kHz

2 dB/DIV.

(2) Set the R4131 as follows:

RBW: 10 kHz

- 3 Connect the TP4 with the INPUT2 of the TR4171 and adjust REF.LEVEL so that the waveform appears on the screen.
- $\stackrel{\textstyle ullet}{4}$ Adjust the L12 so that the waveform aligns with the center frequency.
- (5) Connect a probe to the TP5 and adjust REF.LEVEL so that the waveform appears on the screen.
- (6) Adjust the L13 so that the waveform aligns with the center frequency.
- 7 Connect a probe to the TP7 and adjust REF.LEVEL of the TR4171 so that the waveform appears on the screen.

10.4 IF Filter Adjustment (YTO-CONT/IF Board)

- (8) Adjust the L23 so that the waveform aligns with the center frequency.
- (9) Connect a probe to the TP8 and adjust REF.LEVEL of the TR4171 so that the waveform appears on the screen.
- (10) Adjust the L24 so that the waveform aligns with the center frequency.
- 10.4.4 Resolution Bandwidth Level Adjustment
 - (1) Connect the TP5 with the INPUT2 of the TR4171.
 - (2) Set the TR4171 as follows:

CENTER FREQ.: 3.5795 MHz FREQ. SPAN : 100 kHz 2 dB/DIV.

(3) Set the R4131 as follows:

RBW: 300 kHz

- 4 Adjust REF.LEVEL so that the waveform positions at the center on the scale of the TR4171 and store the waveform.
- (5) Set the R4131 as follows:

RBW: 10 kHz

- (6) Adjust the R67 so that RBW is set to the same level as at 300 kHz.
- (7) Set the R4131 as follows:

RBW: 3 kHz

- (8) Adjust the R35 so that RBW is set to the same level as at 300 kHz.
- (9) Connect the J8 with the INPUT2 of the TR4171.
- (10) Set the R4131 as follows:

RBW: 300 kHz

- (11) Adjust REF.LEVEL so that the waveform positions at the center on the scale of the TR4171 and store the waveform.
- (12) Set the R4131 as follows:

RBW: 10 kHz

(3) Adjust the R141 so that RBW is set to the same level as at 300 kHz.

(14) Set the R4131 as follows:

RBW: 3 kHz

- (5) Adjust the R184 so that RBW is set to the same level at 300 kHz.
- 10.4.5 Step Amplifier Adjustment
 - (1) Connect the units as shown in Figure 10-7.

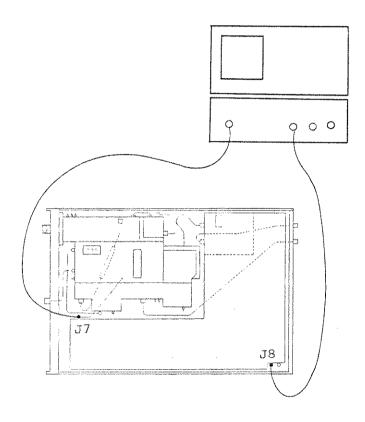


Figure 10-7 Step AMP Adjustment

(2) Set the R4131 as follows:

RBW: 300 kHz

(3) Set the TR4171 as follows:

CENTER FREQ:: 3.5789 MHz
FREQ. SPAN : 200 kHz
REF. LEVEL : -10 dBm
TG LEVEL : -30 dBm

1 dB/DIV.

10.4 IF Filter Adjustment (YTO-CONT/IF Board)

4 Set and adjust R4131 REF.LEVEL and external ATT as shown in Table 10-4 using the R4131 REF.LEVEL as reference.

Table 10-4 Step Amplifier Adjustment

REF.LEVEL	0 dBm	-10 dBm	-20 dBm	-30 dBm	-40 dBm	-50 dBm
External ATT value	0 dB	10 dB	20 dB	30 dB	40 dB	50 dB
VR to be adjusted	Reference	R89	R75	Check	R123	Check

10.5 YTO-CONT Adjustment (YTO-CONT/IF Board)
(BLR-015116)

. 0 5 37m	0.0010 3.44.4.4.4.4.4.4.4.4.4.4.4.4.4.4.4.4.4.	L (3700)	- COM	יור די	T) ~ ~	ا الجرب	מזמו.	-015	1161					
10.5 YT	O-CONT Adjustmen													
(1)	Press [], [],							to O	FF.	$Th\epsilon$	en set	: the	Powe	r
	switch to ON and	d pre	ss 🔲	/	,	and	<u> </u>							
(2)	The following d	ata a	onears	s on	the	scre	en (daif	lav:					
	Inc Lorrowing a	<i></i>	- L- car			501								
		+4 1	7					1			7			
		, ,		<u> </u>	 - -		<u> </u>		-		- Andrews			
		ANA:	LYZER	TEST					ļ 1					
				OVERVIEW OF THE PRESE	:	Y,GA REF,	1	! .5 V	: -					
					:	REF,					1			
	ļ			<u> </u>	-:	REF,	+10	ν	<u> </u>	-				
					:	SLO								
				: !	:	SLOF	}E 2 	V/G	Ηz					
			Ç	UIT	:	UNIT								
	i. Torresta de la companya del companya de la companya del companya de la company	0,	7			<u> </u>								
	<u>ļ</u>			<u> </u>			L		<u>i</u>	L	į			
	,	7iaur	e 10−8	An.	alv	zer T	'est	Dis	plav					
	•				2				2					
3	Move the mark "	#" to	REF.+	·10 V	wi	th th	ie [Φ.	an	d [4	keys.	,	
4	Adjust the R232 scale.	so th	nat th	e di	spl	ayed	line	e al	igns	the	e top	line	on t	he
(5)	Move the mark "	#" to	REF	13.5	V,	with	the		子	and		key	75.	
6	Adjust the R240 scale.	so ti	nat th	ne di	spl	ayed	line	e al	igns	th∈	e top	line	on t	he
7	Move the mark ":	#" to	REF.+	-13 V	wi	th th	ie [♡	an	d [keys.		
8	Check whether the on the scale.	ne dis	splaye	d li	ne	is al	.most	OV:	erla	pped	l on t	he to	p li	ne

- (10) Set the local feed-through to the center of the screen by the encoder.
- 11) Adjust the R355 so that the local feed-through does not shift horizontally even if the frequency span is set to 10 MHz.

(9) Set the offset of the R4(3) as follows:

CENTER FREQ.: 0 MHz FREQ. SPAN : 20 MHz

10.5 YTO-CONT Adjustment (YTO-CONT/IF Board)

(BLR-015116)

(12) Main Span

Connect the units as shown in Figure 10-9.

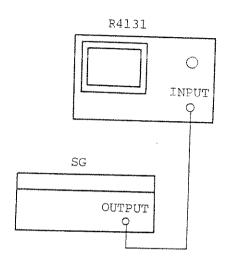


Figure 10-9 Adjustment for Main Span

13 Set the SG as follows:

FREQUENCY: 800 MHz AMPLITUDE: +10 dBm

(14) Set the R4131 as follows:

CENTER FREQ.: 2 GHz FREQ. SPAN : 4 GHz

- 15 Adjust the R308 so that the spectrum aligns the scale.
- (6) Set the SG of FM span as follows:

FREQUENCY: 80 MHz AMPLITUDE: +0 dBm

(17) Set the R4131 as follows:

FREQ. SPAN: 10 MHz

(18) Adjust the R319 so that the spectrum aligns the first vertical line from both ends of the scale.

10.5 YTO-CONT Adjustment (YTO-CONT/IF Board)
(BLR-015116)

(19) Set the SG of 0M tune A as follows:
FREQUENCY: 800 MHz AMPLITUDE: +0 dBm
20 R4131 as follows:
CENTER FREQ.: 0 MHz FREQ. SPAN : 20 MHz CF CAL
(2) Adjust the R287 so that the local feed-through is 0 MHz ±2 MHz.
22) Set the R4131 as follows:
CENTER FREQ.: 3200 MHz FREQ. SPAN : 20 MHz CF CAL
23) Adjust the R270 so that the spectrum is 3200 MHz ±2 MHz.
24 Repeat steps 20 to 23 .
25) Tune B
Set the Power switch of the R4131 to OFF.
$\stackrel{\circ}{(26)}$ Set the Power switch to ON while the $\stackrel{\circ}{\Box}$ key is pressed down.
27 The following data appears on the screen display:
A: 96 B: 32 FM: 32
01,Dec,87
28) Set the R4131 as follows:
CENTER FREQ.: 0 MHz FREQ. SPAN : 20 MHz
29 Turn the encoder so that B: 05 is set.
30 Press and .
31) Turn the encoder so that B: CD is set.
32) Adjust the R269 so that the current waveform aligns the stored waveform.

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10.5 YTO-CONT Adjustment (YTO-CONT/IF Board)
(BLR-015116)

33 Tune FM	
Set the R4131 as follows:	
CENTER FREQ.: 0 MHz FREQ. SPAN : 200 kHz SWEEP TIME : 5 ms/	
34) Turn the encoder so that FM: F8 is set.	
35) Press and .	
36) Turn the encoder so that FM: 32 is set.	
37 Adjust the R317 so that the spectrum aligns the stored waveform.	
38 Slope	
Press , and and data shown in Figure 2-8 appears.	
39 Move the mark "#" to SLOPE_0 V with the 🗘 and 👽 keys.	
$\stackrel{\textstyle \bigodot}{40}$ Adjust the R261 so that the displayed line aligns with the bottom line on the scale.	
41) Similarly, move the mark "#" to SLOPE_2 V/GHz with the 1 and	
keys.	
(42) Adjust the R257 so that the displayed line aligns with the top line on the scale.	

10.6 RF Block Adjustment

- 10.6.1 Third Local Oscillator Adjustment
 - (1) Connect the R4136 INPUT to the CAL.OUT connector.
 - (2) Set the R4136 as follows:

CENTER FREQ.: 200 MHz FREQ. SPAN : 20 kHz REF. LEVEL : -25 dBm RBW : 1 kHz

10 dB/DIV.

- 3 Adjust the C20 so that spectrum positions at the center of the oscillating start frequency and stop frequency.
- 4 Set the R4136 as follows:
 1 dB/DIV.
- (5) Adjust the R27 so that the CAL.OUT level is -30 dBm ±0.5 dB.
- 10.6.2 Second Local Oscillator Adjustment
 - (1) Connect 2ND LOCAL OUT on the rear panel of the R4131 to R4136 INPUT,
 - (2) Set the R4136 as follows:

CENTER FREQ.: 3770 MHz FREQ. SPAN : 2 MHz

- 3 Turn the adjusting bar on the upper cover of the second local block so that the frequency is 3770 MHz.
- 10.6.3 Fourth Local Oscillator Adjustment
 - (1) Remove a shorting pin from the J3 and connect a probe to the J3, pin 2.
 - (2) Set the R4136 as follows:

CENTER FREQ.: 30 MHz FREQ. SPAN : 500 kHz REF. LEVEL : 0 dBm 2 dB/DIV.

(3) Adjust the L13 so that the peak of the waveform is set.

10.7 Location Diagram of YTO CONT/IF Board

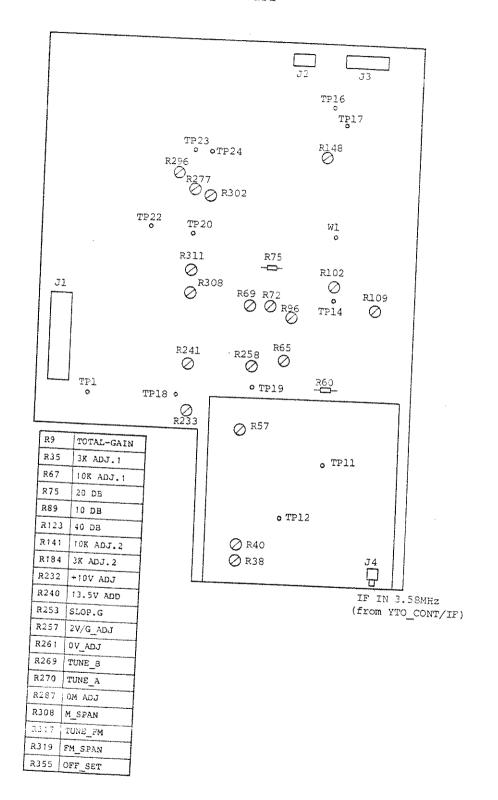


Figure 10-10 Location Diagram of YTO CONT/IF Board

10 - 20

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10.8 Location Diagram of Analog Board

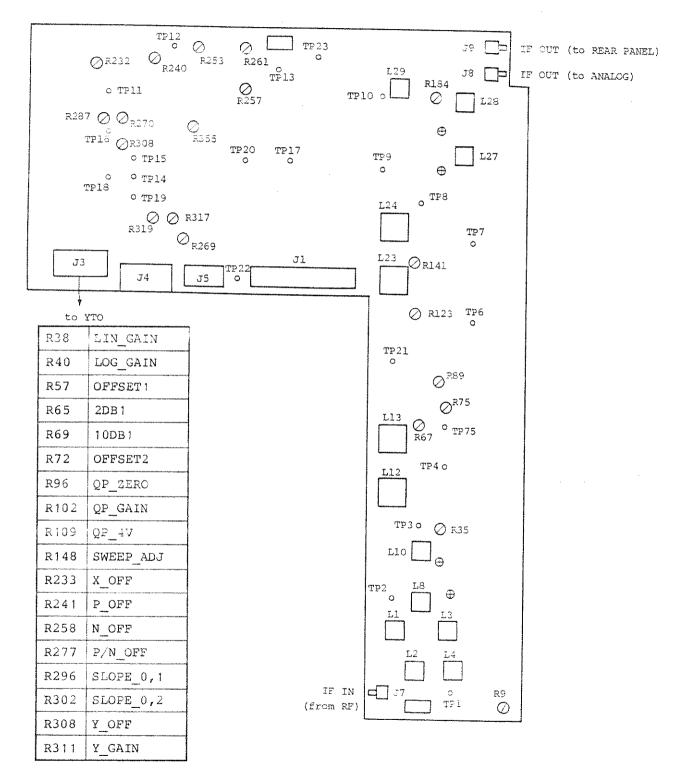


Figure 10-11 Location Diagram of Analog Board

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11. Performance Testing

11. PERFORMANCE TESTING

This section describes performance test procedures for the R4131.

11.1 Preparation

The equipment for the performance testing are listed in Tables 11-1.

Table 11-1 Equipment Required for Performance Testing

Equipment	Specifications	Recommended model
<pre>(1) Synthesized signal source</pre>		TR4511
(2) Function generator	Frequency accuracy: 0.5% or less	
(3) 10 dB step ATT 1 dB step ATT	Accuracy: ±0.5 dB or less, 0 to 70 dB or more Accuracy: ±0.1 dB or less, 0 to 12 dB or more	
(4) Power meter	Frequency range: 10 MHz to 8 GHz	
(5) Power sensor		
(6) Sweep oscillator	Frequency range: 10 MHz to 8 GHz	TR4515
(7) Sweep adapter		TR13211
(8) Impedance converter		ZT301

11.2 General Precautions

11.2 General Precautions

- (1) Always operate the instrument at the specified voltage. Refer to Section 1.3 for the power line voltage.
- (2) The operating temperature range should be 0°C to 50°C, and the relative humidity less than 85%.
- (3) Warm up the instrument for about 30 minutes before starting the performance test.

11.3 Frequency Span Accuracy

Specification: The frequency span between two arbitrary points on the

display screen must be ±5% or less.

Equipment used: Synthesized signal source, function generator

(1) Description

Test the accuracy of frequency span by using the synthesized signal source and function generator. Use the 800 MHz radio frequency of the synthesized signal for the frequency span of 4 GHz to 1 GHz. For the frequency span of 500 MHz to 500 kHz, use the reference synthesized signal subtracted by the span width frequency. For the frequency span of 200 kHz to 50 kHz, use the pulse modulation synthesized signal of the function generator.

(2) Procedure

(1) Set the R4131 as follows:

FREOUENCY SPAN : 4 GHz

RESOLUTION BANDWIDTH : AUTO (1 MHzw)

REFERENCE LEVEL : COARCI, 10 dB/DIV, -10 dBm

INPUT ATTENUATOR : 0 dB
TRACE : WRITE
VIDEO FILTER BAND WIDTH: 1 MHz
SWEEP TRIGGER : FREE RUN

- 2 Test frequency spans from 4 GHz to 1 GHz Referring to Figure 11-1, connect the output of TR4511 synthesized signal source to the INPUT connector of the spectrum analyzer.
- 3 Set the output of TR4511 synthesized signal sourse to -5 dBm, 800 MHz, modulation off.
- 4 Turning the TUNING dial on the spectrum analyzer, adjust the local feedthrough (zero carrier wave) to position it on the leftmost graticule on the display screen. Check that the 4th signal (3.2 GHz) from the local feedthrough (without counting the feedthrough itself) is positioned on or within ±0.4 division of the eighth graticule from the left most graticule (without counting the leftmost graticule itself). (See Figure 11-1.)

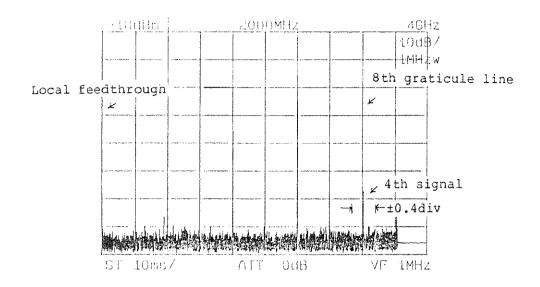


Figure 11-1 Frequency Span 4 GHz Test

- (5) With the spectrum analyzer SPAN switch set to 2 GHz, turn the TUNING dial to position the local feedthrough on the leftmost graticule on the display screen. Check that the second signal (1.6 GHz) from the local feedthrough is positioned within ±0.4 division of the eighth graticule from the left.
- 6 Next, with the spectrum analyzer SPAN switch set to 1 GHz, turn the TUNING dial to position the local feedthrough on the leftmost graticule on the display screen. Check that the first signal (800 MHz) from the local feedthrough is positioned within ±0.4 division of the eighth graticule from the left.

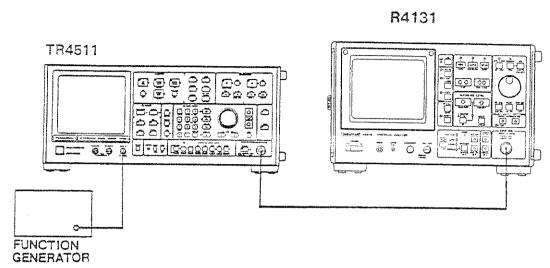


Figure 11-2 Frequency Span Test Setup

- 7) Test frequency spans 500 MHz to 500 kHz. Set the spectrum analyzer INPUT ATTENUATOR switch to 10 dB and the SPAN switch to 500 MHz.
- 8 Set the output of TR4511 synthesized signal source to -10 dBm, 1 GHz modulation off.
- 9 Turning the TUNING dial, adjust the 1 GHz input signal to the leftmost graticule on the display screen.
- Set the output frequency of the TR4511 synthesized signal source to 1.4 GHz. Check that the signal is positioned on the eighth graticule from the leftmost graticule on the display screen (or within ± 0.4 division of the eighth graticule). (See Figure 11-3.)

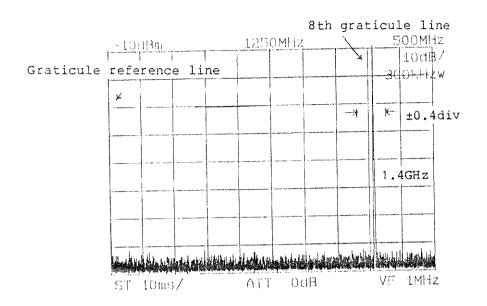


Figure 11-3 Frequency Span 500 MHz Test

Perform similar tests by reducing the frequency span to 200 MHz, 100 MHz, and finally to 500 kHz. For each frequency span, adjust the 1 GHz signal to be on the leftmost graticule on the display screen; then, apply a signal having a frequency equal to 1 GHz + 0.8 x span, checking that the input signal is positioned on the eighth graticule from the leftmost graticule on the screen (or within ±0.4 division of the eighth graticule).

Table 11-2 Frequency Span 500 MHz to 500 kHz Test

Frequency span	Signal adjusted to be on the leftmost graticule on the display screen	Second in- put signal	Tolerance
500 MHz	1 GHz	1.4 GHz	
200 MHz	1 GHz	1.16 GHz	Check that the second input signal is posi-
100M	1 GHz	1.08 GHz	tioned on the eighth
50M	1 GHz	1.04 GHz	graticule from the leftmost graticule on
20M	1 GHz	1.016 GHz	the display screen for
10M	1 GHz	1.008 GHz	within ±0.4 division of the eighth graticule.)
5м	1 GHz	1.008 GHZ	gracicate.)
2M	1 GHz	1.0016 GHz	
1 M	1 GHz	1	
500k	1 GHz	1.0008 GHz	

- 12) Next, perform frequency span 200 kHz to 50 kHz tests using the same setup as shown in Figure 11-2.
- 13 Set the output of the TR4511 synthesized signal source as follows:

Modulation: External pulse modulation

Output level: -10 dBm

Set the function generator as follows:

Waveform: Square wave Output amplitude: 0 to +5 V

(4) Set the output frequency of the function generator to 20 kHz. Turn the TUNING dial to bring the reference spectrum to the leftmost graticule on the display screen. Check that the eighth signal from the reference spectrum is positioned on the eighth graticule from the leftmost graticule on the display screen (or within ±0.4 division of the eighth graticule). (See Figure 11-4.)

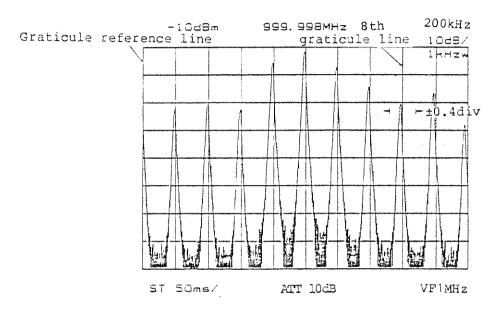


Figure 11-4 Frequency Span 200 kHz Test

(15) Similarly, test frequency span 100 kHz and 50 kHz by referring to Table 11-3.

Table 11-3 Tests for Frequency Spans of 200 kHz or Less

Span	Function generator output frequency	Eighth span position
200 kHz	20 kHz	Within ±0.4 division of the eighth graticule from the
100 kHz	10 kHz	leftmost graticule on the display screen
50 kHz	5 kHz	* *

11.4 Center Frequency Readout Accuracy

Specification: R4131C/CN ...

Less than ±10 MHz After ZERO CAL R4131D/DN ...

Less than ±100 kHz + SPAN 3% or less

after ZERO CAL

Within the range of 0 Hz to 2.5 GHz in center frequency

and 5 ms to 0.5 S/DIV in sweep time.

Less than ±10 MHz After ZERO CAL

Center frequency 2.5 GHz or more.

Equipment used: TR4511

(1) Description

Display the signal applied from the TR4511 synthesized signal source to the R4131 in the center of the display screen and test this center frequency as displayed.

NOTE: Perform zero calibration before performing the center frequency readout accuracy test. (See Section 4-3)

- (2) Procedure
- (1) With the spectrum analyzer INPUT connector open, press the ZERO CAL switch to perform zero calibration.
- ② Set the spectrum analyzer as follows:

FREQUENCY SPAN

: 4 GHz

RESOLUTION BANDWIDTH

: AUTO (1 MHzw)

REFERENCE LEVEL

: COARSE, 10 dB/DIV, 0 dBm

INPUT ATTENUATOR

: 10 dB

TRACE

: WRITE

VIDEO FILTER BAND WIDTH: 1 MHz

SWEEP TRIGGER

: FREE RUN

R4131

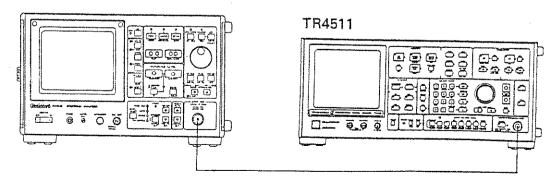


Figure 11-5 Center frequency readout accuracy test setup

- 3 Set the frequecny to test the TR4511 synthesized signal source. An example of 1 GHz.
- 4 Set the dial of spectrum analyzer to 1000 MHz, gradually decrease the frequency span from 4G, 2G, 1G and so on, and set the frequency span so that the waveforms can be displayed within the screen.
- (5) Make sure that the shift from the center frequency is within the range of specifications (see Figure 11-6).

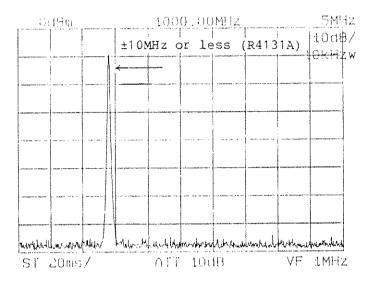


Figure 11-6 Center Frequency Readout Accuracy Test

11.5 Residual FM

Specification: Less than 2 $\rm kHz_{p-p}/100\ ms$

(1) Description:

The calibration signal with a stabilized frequency from this spectrum analyzer is used to perform the residual FM test. The test is performed by FM demodulation by using the R4131 as a fixed tuned receiver with its frequency span set to zero span.

Demodulation is accomplished by using the slope of the spectrum analyzer IF bandpass filter.

NOTE: When performing the residual FM test, install the spectrum analyzer in a place free from vibration, because accuracy of measurement is extremely susceptible to vibrations.

(2) Procedure

Set the spectrum analyzer as follows:

FREQUENCY SPAN : 100 MHz CENTER FREQ : 200 MHz

RESOLUTION BANDWIDTH : AUTO (300 kHzw)

REFERENCE LEVEL : COARSE, 2 dB/DIV, -40 dBm

INPUT ATTENUATOR : 10 dB
TRACE : WRITE
VIDEO FILTER BAND WIDTH: 1 MHz
SWEEP TRIGGER : FREE RUN

2) Connect the spectrum analyzer CAL OUT connector and the INPUT connector with the supplied cable as shown in Figure 11-7.

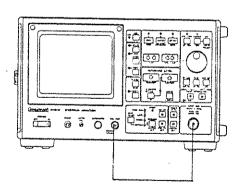


Figure 11-7 Residual FM Test Setup

- 3 Reduce the spectrum analyzer frequency span to 100 kHz. If the 200 MHz signal moves from the center of the display screen, center it again by turning the TUNING dial. The resolution bandwidth is set to 10 kHz.
- 4 Set the spectrum analyzer to the ZERO SPAN mode, and turn the TUNING dial to bring the signal level closer to the center line on the display screen.
- (5) With the sweep time/division set to 0.1 second, press the STORE switch twice to keep the waveform still.

 Check that the peak-to-peak level change in any division (that is, 0.1 second) on the horizontal axis is 1.2 divisions or less as shown in Figure 11-8.

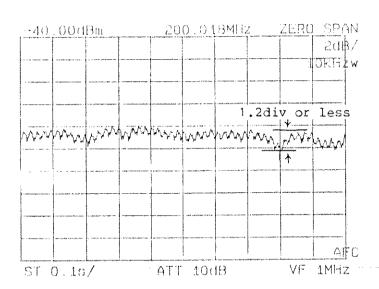


Figure 11-8 Residual FM Test

The value of 1.2 divisions has been acquired for the following reason: The 10 kHz bandwidth filter of the spectrum analyzer is used to allow the residual FM to be displayed on the display screen. The residual FM can be visually observed when the spectrum analyzer is set to a resolution bandwidth of 10 kHz. (See Figure 11-9.) As can be seen from this figure, a 2 kHz change in the frequency axis moves the level about 1.2 divisions.

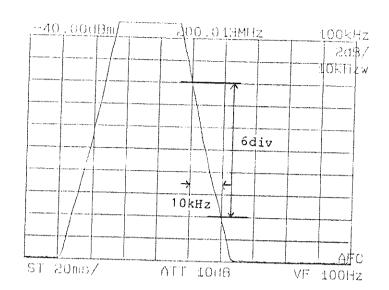


Figure 11-9 Residual FM to AM Conversion Display

Therefore, if the peak-to-peak level change as shown in Figure 11-8 is less than 1.2 divisions, it follows that the residual FM is less than $2 \, \mathrm{kHz}$.

11.6 Noise Sidebands

Specification: -80 dBc or less with a resolution bandwidth of 1 kHz and

10 Hz video filter at the position which is 20 kHz from

the carrier

Equipment used: Synthesized signal source

(1) Description

The noise sidebands test is performed using stable, high-purity 1 GHz, -10 dBm signals.

- (2) Procedure
- 1) Connect the spectrum analyzer and the synthesized signal source to each other as shown in Figure 11-10.
- 2) Set the output of the synthesized signal source to 1 GHz (carrier wave) and -10 dBm.
- (3) Set the spectrum analyzer as follows:

FREQUENCY SPAN : 1 GHz

CENTER FREQ : 1 GHz

RESOLUTION BANDWIDTH : AUTO (300 kHzw)

REFERENCE LEVEL : COARSE, 10 dB/DIV, -10 dBm

INPUT ATTENUATOR : 10 dB
TRACE : WRITE
VIDEO FILTER BAND WIDTH: 1 MHz

SWEEP TRIGGER : FREE RUN

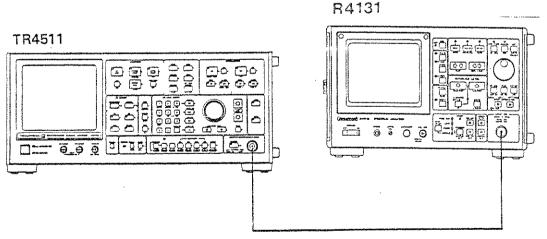


Figure 11-10 Noise Sidebands Test Setup

- 4 Reduce the span to 100 kHz. If the waveform peak moves from the center of the display screen, center it again by turning the TUNING dial.
- (5) If the peak moves from the center of the display screen, center it again by turning the TUNING dial.

- $\stackrel{\textstyle \leftarrow}{}$ Set the reference level to -30 dBm and the video filter to 10 Hz.
- 7 Measure the noise sidebands at the position which is 2 divisions (20 kHz) from the center of the display screen. Check that the noise sidebands is lower than the reference level by 60 dB or more as shown in Figure 11-11.

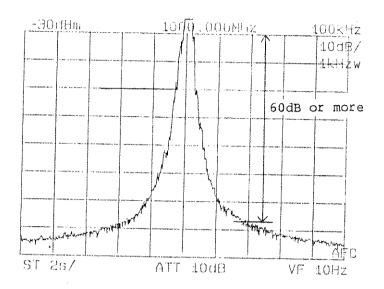


Figure 11-11 Noise Sidebands Measurement

11.7 Resolution Bandwidth Accuracy

11.7 Resolution Bandwidth Accuracy

Specification: Resolution bandwidth between -3 dB points from the signal

peak must be calibrated to ±20% or less.

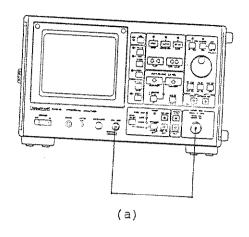
Equipment used: Synthesized signal source

(1) Description

The resolution bandwidth is tested by setting the spectrum analyzer vertical axis to the 2 dB/division mode and measuring the width between two points -3 dB from the signal peak.
Resolution bandwidths narrower than 3 kHz are tested by applying 3.58 MHz signals to the spectrum analyzer IF FILTER IN connector.

(2) Procedure

① Connect the calibration signal of the spectrum analyzer to the INPUT connector as shown in Figure 11-12 (a).



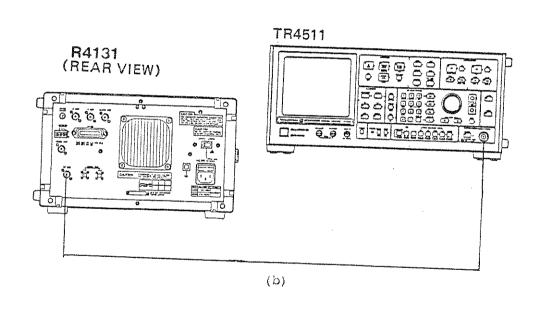


Figure 11-12 Resolution Bandwidth Accuracy Test Setup

② Set the spectrum analyzer as follows:

FREQUENCY SPAN : 1 GHz
CENTER FREQ : 200 MHz
RESOLUTION BANDWIDTH : AUTO

REFERENCE LEVEL : COARSE, 2 dB/DIV, -23 dBm
INPUT ATTENUATOR : 10 dB

TRACE : 10 dB
TRACE : WRITE
VIDEO FILTER BAND WIDTH: 1 MHz

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SWEEP TRIGGER: FREE RUN SWEEP TIME/DIV: 10 ms

- 3 Set the span to 2 MHz. If the signal peak moves from the center of the display screen, center it again by turning the TUNING dial.
- (4) Set the resolution bandwidth to 1 MHz.
- 5 Turning the spectrum analyzer AMPTD CAL control, adjust the signal peak to be 1.5 divisions (3 dB) above the horizontal axis in the center of the display screen. (See Figure 11-13.)

 Then, measure the width of the two points on the horizontal axis traversed by the signal. This width is taken as the 3 dB bandwidth.

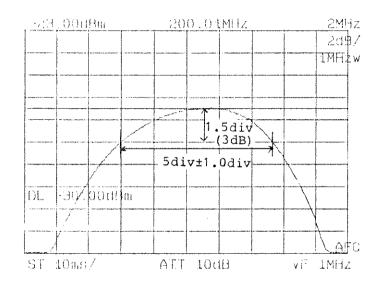


Figure 11-13 Resolution Bandwidth Accuracy Test

- 6 Move the signal to left and right by turning the TUNING dial to determine the order of the graduation in which the measured bandwidth falls. Check that this width is between 4 and 6 divisions (5 ±1 divisions).
- 7 Change the spectrum analyzer frequency span and resolution bandwidth to the values specified in Table 11-4, and repeat steps (5) and (6) above.

Table 11-4	Resolution	Bandwidth	Test	1	MHz	to	10	kHz	

Resolutio	quency	3 dB down width		
bandwidth	span	min.	max.	
1 MHz	2 MHz	4 div	6 div	
300 kHz	500 kHz	4.8 div	7.2 div	
100 kHz	200 kHz	4 div	6 div	
30 kHz	100 kHz	2.4 div	3.6 010	
10 kHz	50 kHz	1.6 div	2.4 div	

- (8) In testing resolution bandwidths 3 kHz to 1 kHz, remove the top cover of the spectrum analyzer and apply 3.58 MHz, -20 dBm signals to the IF FILTER IN connector from the synthesized signal source. (See Figure 11-12 (b).)
- 9 Set the spectrum analyzer resolution bandwidth to 3 kHz and adjust the output frequency of the synthesized signal source for the maximum waveform peak by varying the output frequency at the 10 Hz place.
- 10 Adjust the output level of the synthesized signal synthesized source to bring the spectrum analyzer display level to 1.5 divisions above the horizontal axis in the center of the display screen.
- Reduce the output frequency of the synthesized signal source until the waveform peak displayed on the display screen coincides with the horizontal axis in the center of the display screen. Record this output frequency as f1.
- 12) Next, increase the output frequency of the synthesized signal source until the waveform peak rises once above the horizontal axis in the center of the display screen, and then correspondingly falls. Record this output frequency as f2.
- \bigcirc Determine the 3 dB bandwidth by calculating f2 minus f1. Check that this value falls between 2.4 and 3.6 kHz (3 ±0.6 kHz or less).
- 14) Test resolution bandwidths 1 kHz according to Table 11-5. Keep records of the resultant 3 dB resolution bandwidth values for use in the resolution bandwidth selectivity test described in Section 11.8.

11.7 Resolution Bandwidth Accuracy

Table 11-5 Resolution Bandwidth Accuracy Test 3 kHz to 1 kHz

	TR4511 output			
Resolution bandwidth	frequency variation place	min.	max.	
3 kHz	10 Hz	2.4 kHz	3.6 kHz	
1 kHz	10 Hz	0.8 kHz	1.2 kHz	

3 Resolution Bandwidth Selectivity

Specification: 60 dB/3 dB resolution bandwidth ratio: 15:1 Equipment used: Synthesized signal source

(1) Description

The 60 dB bandwidth of the spectrum analyzer is determined first, and is then compared with the 3 dB bandwidth obtained in Section 11.7 to determine resolution bandwidth selectivity. As in Section 11.7, the resolution bandwidth selectivity is tested in two parts: 1 MHz to 10 kHz, and 3 kHz or less resolution bandwidths.

(2) Procedure:

1) Set the spectrum analyzer as follows:

FREQUENCY SPAN : 4 GHz CENTER FREQ : 200 MHz RESOLUTION BANDWIDTH : 1 MHzw

REFERENCE LEVEL : COARSE, 10 dB/DIV, -10 dBm

INPUT ATTENUATOR : 10 dB TRACE : WRITE VIDEO FILTER BAND WIDTH: 10 kHz SWEEP TRIGGER : FREE RUN SWEEP TIME/DIV : 10 ms

R4131

R4511

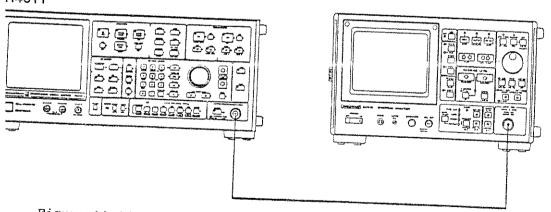


Figure 11-14 Resolution Bandwidth Selectivity Test Setup

Set the synthesized signal source to 200 MHz (CW), -10 dBm. Connect the spectrum analyzer and the synthesized signal source to each other as shown in Figure 11-14.

Press the SPAN switch to activate the frequency span. Reduce the span while turning the TUNING dial to adjust the signal to be in the center of the display screen. Select the minimum span that allows the two points 60 dB lower than the signal peak to be observed on the screen.

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11.8 Resolution Bandwidth Selectivity

- 4 Turn the AMPTD CAL control to bring the signal peak to the top graticule on the display screen.
- (5) Turn the TUNING dial to position the 60 dB point for the best reading.
- (6) Measure and record the 60 dB bandwidth. Check that the ratio of the 60 dB bandwidth to the 3 dB bandwidth measured in Section 11.7 is 15 or less.
- 7) Repeat steps 3 to 6 for resolution bandwidths of 300 kHz to 10 kHz as well.
- 8 Connect the output of the synthesized signal source to the spectrum analyzer IF FILTER IN connector as shown in Figure 11-12 (b).
- 9 Set the output frequency of the synthesized signal source to 3.58 MHz (CW), -20 dBm.
- Adjust the output frequency of the synthesized signal source for a maximum reading on the R4131 display screen, and set the signal to be on the reference graticule.
- Increase the output frequency of the synthesized signal source until the signal level is reduced 60 dB (6 graticules).

 Now measure and record this frequency as f1.
- Reduce the output frequency of the synthesized signal source until the signal level is up 60 dB (6 graticules). Again, measure and record this frequency as f2.
- 13 Determine the 60 dB bandwidth by calculating f1 minus f2. Check that the following relation holds: 60 dB bandwidth/3 dB bandwidth ≤ 15.
- (4) Repeat steps 10 to 13 for resolution bandwidth of 1 kHz.

11.9 Resolution Bandwidth Switching Accuracy

.9 Resolution Bandwidth Switching Accuracy

Specification: ±1 dB (referenced to 300 kHz bandwidth)

(1) Description

The amplitude readout error associated with switching of the resolution bandwidth is measured using a CAL signal.

(2) Procedure

1) Set the R4131 as follows:

FREQUENCY SPAN

: 1 GHz

CENTER FREQ

: 200 MHz

RESOLUTION BANDWIDTH: 1 MHz REFERENCE LEVEL : COARSI

: COARSE, 2 dB/DIV, -28 dBm

INPUT ATTENUATOR

: 10 dB

TRACE

: WRITE

VIDEO FILTER SWEEP TRIGGER

: 10 kHz : FREE RUN

SWEEP TIME/DIV

: 10 ms

- 2 Connect the CAL input to the INPUT connector. (See Figure 11-15.)
- 3 Set the span to 2 MHz, while turning the TUNING dial to center the waveform on the display screen.
- 4 Pressing the RBW switch, set the resolution bandwidth to 300 kHz.

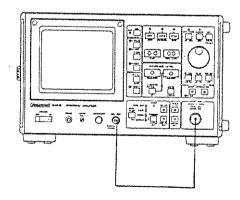


Figure 11-15 Resolution Bandwidth Switching Accuracy Test Setup

5) Turn the AMPTD CAL control to adjust the signal peak to be 1 division lower than the reference graticule on the display screen.

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11.9 Resolution Bandwidth Switching Accuracy

- (6) Set the resolution bandwidth to 1 MHz. Check that the maximum amplitude point is ± 1 dB (± 0.5 division) or less when compared to the 300 kHz resolution bandwidth.
- (7) Similarly, set the span and the resolution bandwidth to 100 kHz. Check that the maximum amplitude point is ±1 dB or less when compared to the 300 kHz resolution bandwidth.
- (8) Also test resolution bandwidths 30 kHz to 1 kHz at the settings specified in Table 11-6.

Table 11-6 Bandwidth Switching Uncertainty

Resolution bandwidth	Frequency span/division	Amplitude readout change
1 MHz	2 MkHz	±1 dB
300 kHz	2 MkHz	0 dB (REF.)
100 kHz	1 MkHz	±1 dB
30 kHz	200 kHz	±1 dB
10 kHz	100 kHz	±1 dB
3 kHz	50 kHz	±1 dB
1 kHz	50 kHz	±1 dB

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11.10 LOG Linearity and LIN Linearity

11.10 LOG Linearity and LIN Linearity

Specification: LOG linearity: ±1 dB/10 dB, ±0.15 dB/1 dB, ±1.5 dB/70 dB

LIN linearity: ±5% of full scale

Equipment used: Synthesized signal source

10 dB step ATT 1 dB step ATT

(1) Description

Linearity test is performed by utilizing the marker on the display screen when the aid of the external signal and the attenuators.

(2) Procedure LOG linearity

(1) Set the R413! as follows:

FREQUENCY SPAN : 1 GHz
CENTER FREQ : 200 MHz
RESOLUTION BANDWIDTH : AUTO
REFERENCE LEVEL : -10 dB
INPUT ATTENUATOR : 10 dB
TRACE : WRITE
VIDEO FILTER BAND WIDTH: 1 MHz
SWEEP TRIGGER : FREE RUN

(2) Set the output frequency of the synthesized signal source to 200 MHz (CW), -10 dBm, and connect the synthesized signal source to the spectrum analyzer INPUT connector using attenuators as shown in Figure 11-16.

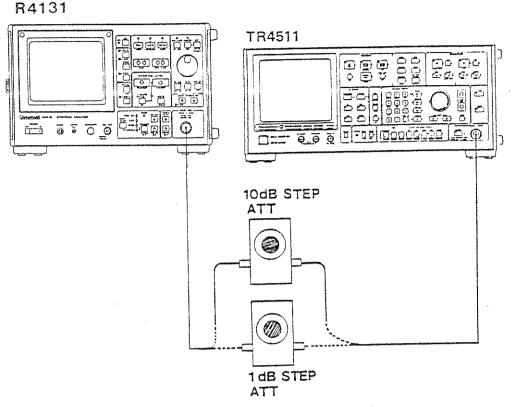


Figure 11-16 LOG/LIN Linearity Test Setup

- (3) Set the 10 dB step ATT to 0 dB.
- 4 Set the span to 2 MHz while turning the TUNING dial to position the signal peak in the center of the display screen. Then, make the following settings:

Resolution bandwidth : 30 kHz Sweep time/division : 20 ms Video filter band width: 10 kHz

- (5) Press the MARKER switch and turn the TUNING dial to position the marker at the signal peak.
- 6 Adjust the AMPTD CAL control to set the marker level reading to -10.0 dBm.
- 7 Vary the 10 dB step ATT 10 dB at a time, checking that the marker level values conform to the values of Table 11-7. With an attenuator setting of 70 dB, set the video filter to 100 Hz and the sweep time/division to 0.1 s in order to prevent noise being superimposed on the signal.

NOTE: If the marker moves off the signal peak during measurement, position it at the signal peak again by turning the TUNING dial.

Table	1	1 - 7	LOG	Line	ar	ity
-------	---	-------	-----	------	----	-----

ATT setting	Marker level readout	Video filter	Sweep time/div
0	-10 dBm (REF)	10 kHz	20 ms
10	-20 ±1 dBm	10 kHz	20 ms
20	-30 ±1 dBm	10 kHz	20 ms
30	-40 ±1 dBm	10 kHz	20 ms
40	-50 ±1 dBm	10 kHz	20 ms
50	-60 ±1 dBm	10 kHz	20 ms
60	-70 ±1 dBm	10 kHz	20 ms
70	-80 ±1.5 dBm	100 Hz	0.1 s

- 8 Connect the 1 dB step ATT to the spectrum analyzer and set the video filter to 10 kHz and the sweep time/division to 20 ms.
- (9) Set the ATT to 0 dB.
- (10) Set the R4131 reference level to 2 dB/division and the resolution bandwidth to 300 kHz. Turn the AMPTD CAU control to adjust the marker level to be -10.0 dBm.
- ① Set the ATT to 2 dB. Check that the resultant marker level reading is -12 dBm ±0.3 dB, or less.

 Next, set the ATT to 10 dB. Check that the resultant marker level reading is -20 dBm ±1 dB, or less.

LIN linearity

- (2) Set the ATT to 0 dB, and set the output level of the synthesized signal source to -10 dBm (70.71 mV).
- (3) Set the R4131 to the LIN mode, and position the marker at the signal peak. Turn the AMPTD CAL control until the marker level is set to 70.71 mV (on the reference graticule).
- (4) Set the ATT to 6 dB. Check that the marker level reading is 35.4 mV ± 3.5 mV, or less.

11.11 Reference Level Accuracy

Specification: The reference level as varied with MIN INPUT ATT 10 dB

(fixed) must be accurate to within 1 dB.

Equipment used: Synthesized signal source

10 dB step ATT 1 dB step ATT

(1) Description

The reference level accuracy can be determined by testing the IF GAIN accuracy in the LOG display mode.

(2) Procedure

(1) Set the R4131 as follows:

FREQUENCY SPAN : 1 GHz

CENTER FREQ : 200 MHz
RESOLUTION BANDWIDTH : AUTO

REFERENCE LEVEL : FINE, 2 dB/DIV, 0 dBm

INPUT ATTENUATOR : 10 dB
TRACE : WRITE
VIDEO FILTER BAND WIDTH: 1 MHz
SWEEP TRIGGER : FREE RUN

2 Set the output frequency of the synthesized signal source to 200 MHz (CW), -10 dBm, and connect the source to the spectrum analyzer INPUT connector using attenuators as shown in Figure 11-17.

R4131C/D

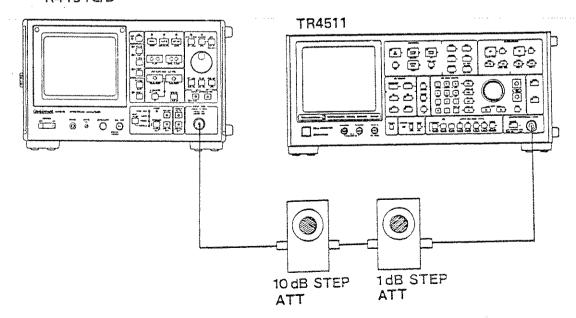


Figure 11-17 Reference Level Accuracy Test Setup

8.12 REFERENCE LEVEL ACCURACY

- (3) Set both the 10 dB and 1 dB step ATTs to 0 dB.
- 4 Set the span to 2 MHz while turning the TUNING dial to position the signal peak in the center of the display screen.
- 5 Then, make the following settings:
 Resolution bandwidth: 300 kHz
 Video filter : 1 kHz
 Sweep time/division : 50 ms

-40 dBm

-50 dBm

- (6) Press the MARKER switch and turn the TUNING dial to position the marker at the signal peak.
- 7 Adjust the AMPTD CAL control to set the marker level reading to -10.0 dBm.
- (8) With the 1 dB step ATT at 1 dB, set the reference level to -1.00 dBm. Check that the marker level reading is -11.00 ±1 dB or less.
- 9 Proceed with further testing with the settings specified in Table 11-8.

Reference level ATT setting Marker readout level setting 0 dBm 0 dB -10.00 dBm (REF.) -1 dBm 1 dB $-11.00 \pm 1 \text{ dBm}$ -2 dBm 2 dB -12.00 ± 1 dBm -13.00 ± 1 dBm -3 dBm 3 dB -4 dBm 4 dB $-14.00 \pm 1 \text{ dBm}$ -5 dBm 5 dB -15.00 ± 1 dBm -6 dBm -16.00 ± 1 dBm 6 dB -7 dBm 7 dB $-17.00 \pm 1 \text{ dBm}$ -8 dBm 8 dB $-18.00 \pm 1 \text{ dBm}$ -9 dBm 9 dB -19.00 ± 1 dBm -10 dBm 10 dB -20.00 ± 1 dBm -20 dBm 20 dB -30.00 ± 1 dBm -30 dBm 30 dB -40.00 ± 1 dBm

Table 11-8 Reference Level Accuracy

 -50.00 ± 1 dBm

 $-60.00 \pm 1 \text{ dBm}$

40 dB

50 dB

11.12 Residual Responses

Specification:

R4131C/D ...

-95 dBm or less (at an input attenuator setting of 0 dB)

R4131CN/DN ...

-93 dBm or less (at an input attenuator setting of 0 dB)

(1) Description

Residual responses refers to the signal displayed on the display screen in the absence of input. Testing is performed at 100 MHz intervals in the range 100 kHz to 3.5 GHz.

(2) Procedure

distant. . Substitute Section (Section)

 $\widehat{\mbox{\scriptsize (1)}}$ After terminating the spectrum analyzer INPUT connector with a 50 Ω terminator (R4131C/D) and a 75 Ω terminator (R4131CN/DN), set the spectrum analyzer as follows:

: 100 MHz FREQUENCY SPAN : 50 MHz CENTER FREQ : 30 kHz

RESOLUTION BANDWIDTH : COARSE, 10 dB/DIV, -50 dBm REFERENCE LEVEL

: 0 dB INPUT ATTENUATOR : WRITE TRACE VIDEO FILTER BAND WIDTH: 1 kHz : FREE RUN SWEEP TRIGGER : 1 s SWEEP TIME/DIV

- 2 Set the TRIGGER MODE switch to SINGLE and press the START switch to test residual responses in the range of 0 to 100 MHz. Check that the residual responses is -95 dBm or less (R4131C/D), -93 dBm or less (R4131CN/DN).
- 3 Turn the TUNING dial to set the center frequency to 150 MHz. Press the START switch to test residual responses in the range of 100 to 200 MHz. Check that the residual responses is -95 dBm or less (R4131C/D), -93 dBm or less (R4131CN/DN).
- 4 Similarly, test residual responses up to 3.5 GHz at 100 MHz intervals.

11.13 Gain Compression

Specification: *MIX input end must be 1 dBm or less for a -10 dBm input.

[*: (Input signal level) - (MIN INPUT ATT)]

Equipment used: Synthesized signal source

Power meter Power sensor 10 dB step ATT

(1) Description

The gain compression is tested by checking to see if the reading level rises 10 dB when the MIX input end level is increased from -20 dBm to -10 dBm.

(2) Procedure

(1) Set the R4131 as follows:

FREQUENCY SPAN : 100 MHz
CENTER FREQ : 200 MHz
RESOLUTION BANDWIDTH : AUTO

REFERENCE LEVEL : COARSE, 10 dB/DIV, -10 dBm

INPUT ATTENUATOR : 10 dB
TRACE : WRITE
VIDEO FILTER BAND WIDTH: 1 MHz
SWEEP TRIGGER : FREE RUN

- 2 Set the output frequency of the synthesized signal source to 200 MHz (CW) and connect it to the power meter, adjusting the synthesized signal source for 0 dBm output.
- 3 Set the 10 dB step ATT to 10 dB and connect it to the spectrum analyzer as shown in Figure 11-18.

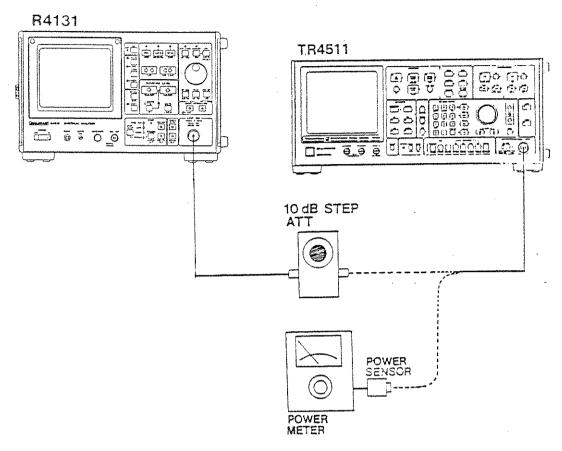


Figure 11-18 Gain Compression Test Setup

- (4) Set the span to 1 MHz while turning the TUNING dial to position the 200 MHz signal in the center of the display screen. Pressing the RBW switch, set the resolution bandwidth to 300 kHz, and set the reference level to 2 dB/DIV.
- (5) Turn the AMPTD CAL control to bring the signal peak to the reference graticule (top graticule) on the display screen.
- (6) Set both the reference level and the 10 dB step ATT to 0 dB. Check that the signal peak falls within 0.5 division (1 dB) of the top graticule (reference graticule) on the display screen.

11.14 Frequency Response

Specification: Frequency response (MIN INPUT ATT: 10 dB)

R4131C	100 kHz≦F≦ ±1 dB or lea	 10 kHz ≦ F ≤ ±3.5 dB or	
R4131D	100 kHz≦F≦ ±1 dB or le	10 kHz ≦ F ≤ ±2 dB or le	
R4131CN/DN	100 kHz \leq F \leq 1.5 GHz \pm 1.5 dB or less	2 GHz 2 GHz≤F≤ less ±4 dB or le	

Equipment used: Sweep oscillator

Power meter Power sensor Sweep adapter

(1) Description

Testing is performed by setting the R4131 to the full span mode and a sweep oscillator to the external sweep mode and observing changes of the amplitude reading on the display screen. Since sweep oscillator frequency responses are included in the measurement results, measure the sweep oscillator response with a power meter prior to testing for later correction of the measurements.

(2) Procedure

1) Set the R4131 as follows:

FREQUENCY SPAN : 4 GHZ
CENTER FREQ : 2000 MHZ

RESOLUTION BANDWIDTH : AUTO

REFERENCE LEVEL : COARSE, 10 dB/DIV, 0 dBm

INPUT ATTENUATOR : 10 dB

TRACE : WRITE, POSI PEAK

VIDEO FILTER BAND WIDTH: 1 MHz
SWEEP TRIGGER : FREE RUN
SWEEP TIME/DIV : 10 ms

(2) Set the sweep oscillator output to 200 MHz (CW), -10 dBm and connect it to the power meter using the A01002 cable. Adjust the output level of the sweep oscillator to -10 dBm. (See Figure 11-19.)

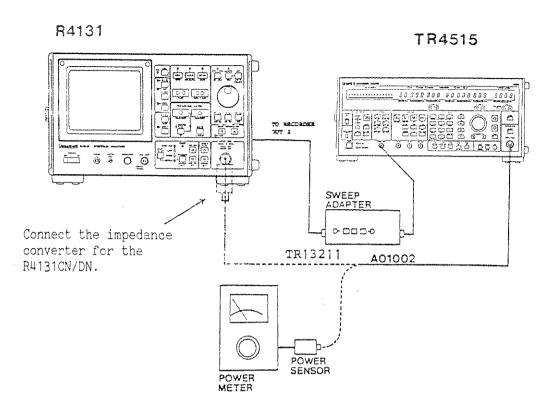


Figure 11-19 Frequency Response Test Setup

- 3 Connect the sweep OSC output to the spectrum analyzer INPUT connector. Connect the impedance converter for the R4131CN/DN. (See Figure 11-19) With its amplitude set to 2 dB/division, set the refrence level to display a 200 MHz signal on the center axis of the display screen.
- (4) Set the sweep oscillator to the external sweep mode, and set the start and stop frequencies to 10 MHz and 4 GHz, respectively.
- (5) Press the sweep adapter START switch, and adjust the START dial to display the signal at the leftmost position on the display screen. Next, press the STOP switch and adjust the STOP dial to display the signal at the rightmost position on the display screen.
- (6) When the SWEEP switch is pressed after the STOP dial has been adjusted, the waveform, shown in Figure 8.20 (a) appears. When a uniform spectrum waveform is not displayed, finely adjust the START and STOP dials.
- (7) Set the sweep time/division to 1 s, and the frequency characteristics will be displayed on the display screen. (See Figure 11-20 (b).) Make sure that the ripple current is within the range of the specifications.

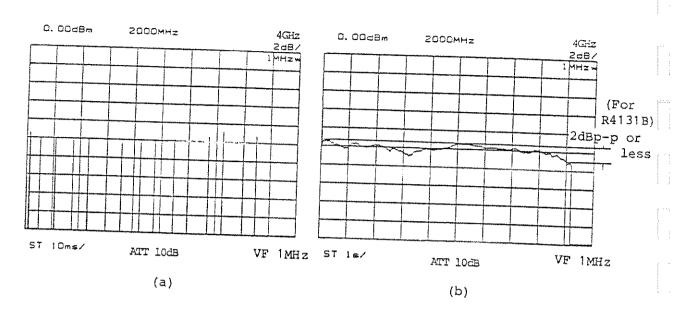


Figure 11-20 Frequency Response (100 kHz - 3.6 GHz)

11.15 Average Noise Level

Specification:

R4131C/D ... -110 dBm or less R4131CN/DN ... -108 dBm or less

(Resolution bend width 1 kHz, Video filter 10 Hz, Input ATT 0 dB, More than 1 MHz in frequency.)

(1) Description

The average noise level is the maximum value of the average noise levels in the 1 kHz resolution bandwidth with an input ATT setting of 0 dB.

 $\underline{\underline{\text{Note}}}$: Be sure to perform amplitude calibration (see Section 4.7) before performing this test.

(2) Procedure

(1) Set the R4131 as follows:

FREQUENCY SPAN : 4 GHz CENTER FREQ : 2000 MHz RESOLUTION BANDWIDTH : 1 MHz : 0 dB INPUT ATTENUATOT REFERENCE LEVEL : -50 dBm TRACE : WRITE VIDEO FILTER BAND WIDTH: 1 kHz SWEEP TRIGGER : FREE RUN SWEEP TIME/DIV : 1 s MARKER : ON

2 Turning the TUNING dial, position the marker at the maximum noise level point. (See Figure 11-21)

③ Press the MKR + CF switch. (Set the center frequency to the marker frequency.) Set the frequency span to zero span and set the resolution bandwidth to 1 kHz.

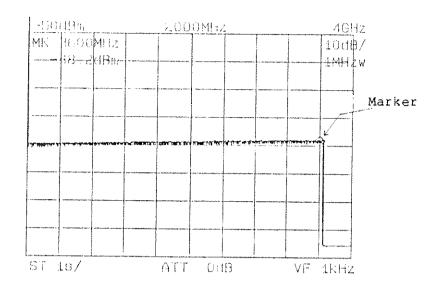


Figure 11-21 Maximum Noise

4 Set the video filter to 10 Hz. (See Figure 11-22) Check that the marker level reading is -110 dBm or less (R4131C/D), and -108 dBm or less (R4131CN/DN).

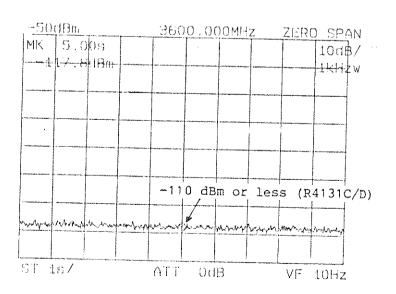


Figure 11-22 Average Noise Level Test

11.16 Sweep Time Accuracy

Specification: ±15%

Equipment used: Synthesized signal source

Function generator

(1) Description

Sweep time accuracy is tested by demodulating signals in the R4131 zero span mode after they are amplitude modulated by the function generator and measuring the periods of the demodulated waves.

(2) Procedure

1 4

(1) Set the R4131 as follows:

FREQUENCY SPAN : 100 MHz
CENTER FREQ : 50 MHz
RESOLUTION BANDWIDTH : 1 MHz

REFERENCE LEVEL : 2 dB/DIV, -10 dBm

INPUT ATTENUATOR : 10 dB
TRACE : WRITE
VIDEO FILTER BAND WIDTH: 10 kHz
SWEEP TRIGGER : FREE RUN
SWEEP TIME/DIV : 10 ms

- Set the output frequency of the synthesized signal source to 50 MHz, -10 dBm, EXT AM mode.
- (3) Set the function generator to generate sine waves at 200 Hz ±0.5%.
- 4 Connect the instruments as shown in Figure 11-23. Turn the R4131 TUNING dial to position the signal in the center of the display screen. Further, set the frequency span to zero span and adjust the TUNING dial to obtain the maximum signal level.
- \bigcirc Adjust the function generator output level to obtain demodulated waves in the order of 3 DIV_{p-p}.
- (6) Adjust the reference level to position the signal at an easily viewed position on the display screen.
- (7) Set the TRIGGER MODE switch to VIDEO.
- 8 Set the sweep time/division to 5 ms and store the resultant waveform. Check that five periods of the demodulated waves have a duration of 25 ±3.75 ms, or less. (See Figure 11-24)
- Similarly, test other sweep time/division with the settings specified in Table 8-11.

Table 11-9 Sweep Time Accuracy

Sweep time/div	Function generator frequency	Duration of five periods
5 ms	200 Hz ±0.5%	25 ms ±3.75 ms
10 ms	100 Hz ±0.5%	50 ms ±7.5 ms
20 ms	50 Hz	100 ms ±15 ms
50 ms	20 Hz	250 ms ±37.5 ms
0.1 s	10 Hz	0.5 s ±75 ms
0.2 s	5 Hz	1 s ±150 ms
0.5 s	2 Hz	2.5 s ±375 ms
1 s	1 Hz	5 s ±0.75 s
2 s	0.5 Hz	10 s ±1.5 s
5 s	0.2 Hz	25 s ±3.75 s
10 s	0.1 Hz	50 s ±7.5 s

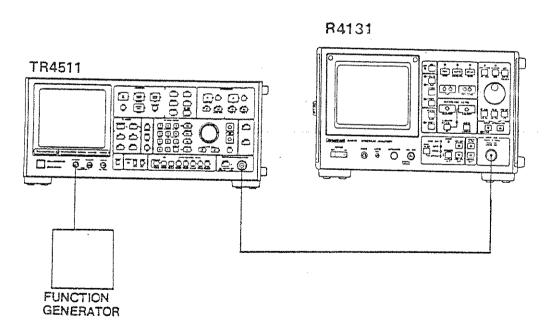


Figure 11-23 Sweep Time Accuracy Test Setup

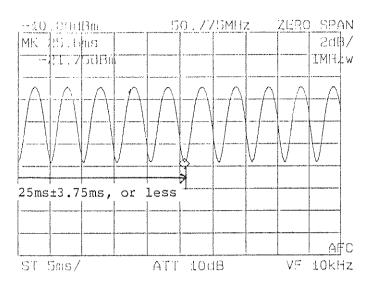


Figure 11-24 Sweep Time Accuracy Test

11.17 Calibrated Output Accuracy

Specification: 200 MHz ±30 kHz, -30 dBm ±0.5 dB:R4131C/D

200 MHz ±30 kHz, 80 dBµ ±0.5 dB :R4131CN/DN

Equipment used: Synthesized signal source

Power meter

(1) Description

Test the accuracy of CAL signal frequency by using the synthesized signal source. Test the accuracy of signal level by connecting the power meter directly to the CAL signal line.

(2) Procedure

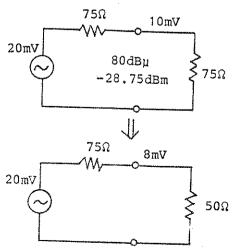
Frequency Test

- 1 Press the R4131 ZERO CAL switch.
- 2 Set the synthesized signal source to 200 MHz, -30 dBm.
- 3 Connect the synthesized signal sourse to the spectrum analyzer INPUT connector. Set the span to 100 kHz while turning the TUNING dial to position the 200 MHz signal in the center of the display screen.
- 4 Next, connect the CAL signal to INPUT connector. (See Figure 11-25) Check that the center frequency is 200 MHz ±30 kHz, or less.

Amplitude Test

- ① Directly connect the power meter to the CAL OUT signal line.
- 2) Make sure that the CAL OUT output signal level is $-30~\mathrm{dBm}~\pm0.5~\mathrm{dB}$ (R4131C/D) or $-28.93~\mathrm{dBm}~\pm0.5~\mathrm{dB}$ (R4131CN/DN).

The reason why the R4131CN/DN has the -28.93 dBm signals when the 80 dB μ CAL OUT signal is measured on the 50Ω power meter:



80 dB
$$\mu$$
, -28.75 dBm is:
10 logP = -28.75 dBm

$$\frac{V^2}{D}$$
 = 1.334 x 10⁻³ (mW)

As R =
$$75\Omega$$
:
V = 10.00 (mV)
Therefore, if the 50Ω power meter is connected:
P = $\frac{V^2}{R} = \frac{(8mV)^2}{50} = 1.28 \times 10^{-3}$ (mW)
10 logP = $\frac{28.93}{8}$ dBm

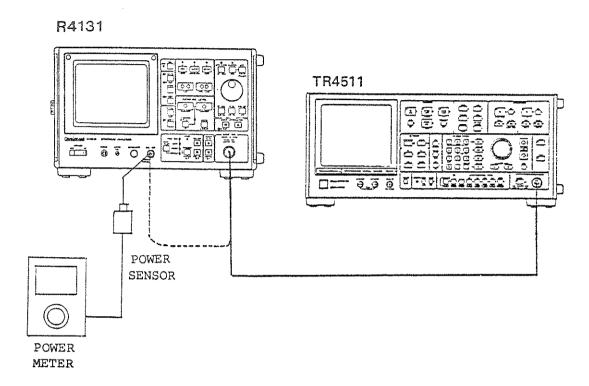


Figure 11-25 Calibrated Output Accuracy Test Setup

R4131 SERIES SPECTRUM ANALYZER INSTRUCTION MANUAL

12. Maintenance Data

12. MAINTENANCE DATA



12.1 Preparation

The equipment and tools necessary for troubleshooting are listed in Table 12-1. The equipment must have equivalent or better performance ratings than those in the table.

Table 12-1 Equipment and Tools Required For

Equipment Digital voltmete	Performance r Range : +1000 H	Recommended equipment
	Range : $\pm 1000 \text{ V}$ Accuracy : $\pm 0.1\%$ Input impedance: $\pm 1000 \text{ M}$	TR6846
High frequency power meter	Frequency: 100 kHz to 8 GHz Sensitivity: -30 dBm to +20 dBm Accuracy: ±0.5 dB	
DC power supply	Output voltage: +10 V Accuracy : +0.03%	TR6142
Oscilloscope	Frequency range: DC to 100 MHz Input impedance: 1 $M\Omega$	
Signal generator	Frequency range : 100 kHz to 1800 MHz Output level : ± 10 dBm or more Output impedance : 50 Ω Frequency accuracy: 2 E-8/day Variable frequency: 1 Hz step	TR4512
FET probe	Frequency range: DC to 500 MHz Input impedance: 1 M Ω or more, 2 pF or less	
Spectrum analyzer	Frequency range : 10 MHz to 8 GHz Frequency accuracy: +100 kHz	R4136
Spectrum analyzer	Input frequency range : 100 kHz to 1.8 GHz Tracking generator output: 400 kHz to 1.8 GHz T.G. output flatness : +1 dB	TR4171 or R4136 + TR4154
	Frequency: 100 kHz to 1500 MHz Sensitivity: -30 dBm to +20 dBm Accuracy: ±0.5 dB	

Table 12-2 Maintenance Tools Required for Troubleshooting

Product name	Stock number	Remarks
Cable (UM-UM)	MM-17	Tool And The Control of the Control
Cable (SMA-SMA)	MM-14	
Cable (BNC-BNC)	MI-02	}
Cable (BNC-UM)	MC-36	2 pcs.
UM to UM Linear Adapter	JCF-AC001JX07	
SMA to SMA Adapter	JCF-AA001JX28	A THE PERSON NAMED IN COLUMN N

12.2 Location Diagram (Top & Bottom)

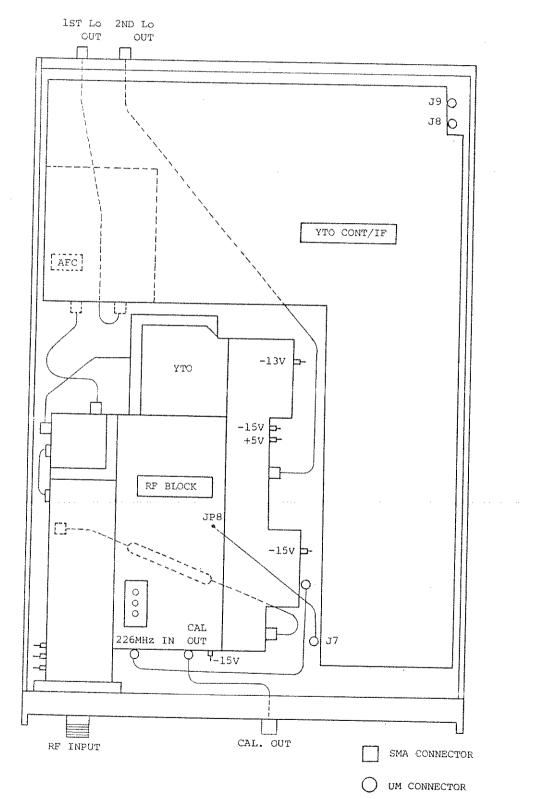


Figure 12-1 Location Diagram (Bottom View)

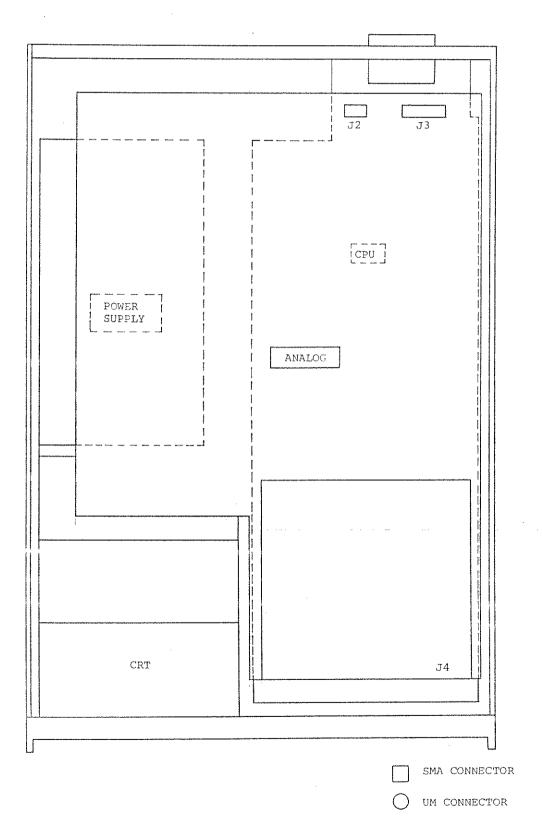


Figure 12-2 Location Diagram (Top View)

12.3 Location Diagram for RF

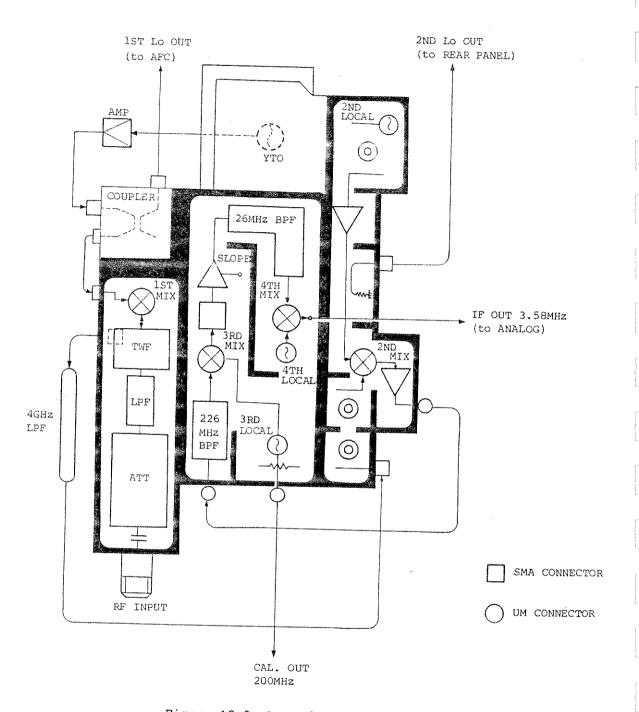


Figure 12-3 Location Diagram for RF

12.4 Block Diagram

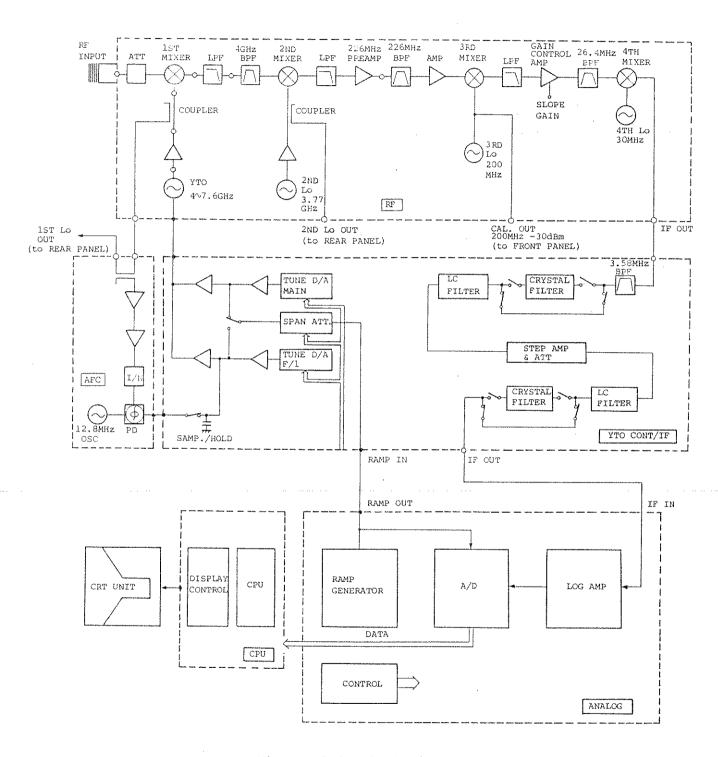


Figure 12-4 Block Diagram

12.5 Self Test

The R4131 performs SELF TEST for the RAM and ROM on the CPU board when power is turned on.

In the case there is a failure RAM or ROM, the following error message is displayed on the CRT.

Message	Mean
 RAM error	Failure RAM U26 or U32 (SMM-8464C-5) on the CPU board (BLR-015114)
ROM error	Failure ROM U21 (SMM-27C25-1) on the CPU board (BLR-015114)

R4131 SERIES SPECTRUM ANALYZER INSTRUCTION MANUAL

A - 1

Appendix

APPENDIX



Oct 20/89

A.1 Explanation of Terminologies

IF Bandwidth

In this spectrum analyzer, a band pass filter (BPF) is used to analyze each frequency component included in input signals. The 3 dB bandwidth of this BPF is called the IF band (see Figure A-1 (a)). The BPF characteristic should be set to the appropriate size according to the sweep width and sweep speed. In this equipment, it is set to the maximum value according to the sweep width. Since this bandwidth can generally improve the resolution (a degree of separation) more and more when it is set narrower, the resolution of the spectrum analyzer is expressed in the narrowest IF bandwidth in some cases (see Figure A-1 (b)).

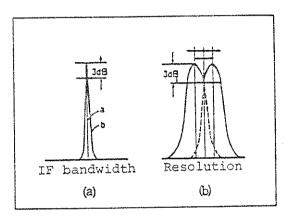


Figure A-1 IF Bandwidth

Gain Compression

In case the input signal becomes larger than a certain value, no correct value is displayed on the CRT screen and a somewhat compressed phenomenon occurs even when the input signal is increased. This is called the gain compression. It expresses the linearity of the input signal range. In general, a level range is used until 1 dB is compressed.

Input Sensitivity

This means the highest capacity of a spectrum analyzer to detect minor signals. The sensitivity is related to the noise generated from the spectrum analyzer itself and it depends on the IF bandwidth used. Generally, the input sensitivity expresses the average noise level in the minimum IF bandwidth of that spectrum analyzer.

Maximum Input Level

This is the maximum allowable level of the input circuit of a spectrum analyzer. The allowable level can be changed by the input attenuator.

R4131 SERIES SPECTRUM ANALYZER INSTRUCTION MANUAL

A.1 Explanation of Terminologies

Residual FM

This is a method to express a short term frequency stability of the local oscillator groups integrated in a spectrum analyzer. The frequency straying per unit time is expressed in p-p. This also indicates the critical value when the residual FM of a measured signal is measured.

Residual Responses

This defines to what level value the spurious signal generated in a spectrum analyzer is suppressed when calculated in terms of the input level. This signal is caused when a particular signal, e.g., the local oscillator output, etc., inside the spectrum analyzer is leaked. Care should be taken in this respect when a very small input signal is analyzed.

Quasi Peak Value Measurements

Disturbing noise received in radio communication often appears in an impulsive state. As an objective evaluation of this disturbance, the disturbing noise component is evaluated with a value proportional to its peak value. Such prerequisite factors as the measuring bandwidth and detection time constant for this measurement are used as the quasi peak values. This is represented by the JRTC Standards in Japan and by the CISPR Standards internationally.

Frequency Response

Frequency response is usually used as a term to indicate the amplitude characteristic with frequency (frequency characteristic). In spectrum analyzer, this term means the frequency characteristic (flatness) of an input attenuator, mixer, etc. at each input frequency. It is represented by \pm dB.

Frequency Span

This means the display range of the ordinates axis (frequency axis) on the Braun tube. The frequency span is set arbitrarily from a broad band to narrow band with the frequency scale which is calibrated accurately.

Zero Span

A spectrum analyzer does not sweep the frequency in this mode. Instead, it sweeps an arbitrary frequency taking the ordinates axis as the time axis.

Spurious

The spurious means unnecessary signals. They are classified into the following categories according to the properties of each signal:

Harmonic spurious: This is defined to indicate the harmonic level to be generated by the spectrum analyzer itself (to be generated in the mixer circuit in general) when no-distortion signal is applied to it. At the same time, it means the capacity of the harmonic wave distortion measurement.

Neighborhood spurious: A small spurious generated in the neighborhood of the spectrum analyzer when a pure single spectrum signal is applied to it.

Non-harmonic spurious: Apart from the above two, the spectrum analyzer generates a certain proper frequency as a spurious. This is also called the residual response.

Noise Bandwidth

This is used widely as performance to express the oscillation purity of an oscillator, etc. In the spectrum analyzer itself, the noise is generated in the vicinity of the spectrum on the Braun tube from local oscillator and phase lock loop, thus lowering the analyzing capacity of the analyzer. To compensate, the analyzer defines its own sideband range enabling it to analyze the incoming signal noise sidebands within this range. The spectrum analyzer expresses the noise sideband characteristic as follows:

Example:

-70 dB apart from the carrier by 20 kHz where the IF bandwidth is assumed to be 1 kHz. It is also expressed with the energy which exists within the 1 Hz bandwidth in general (Figure A-2 (b)).

Since this value is -70 dB at the 1 kHz bandwidth when expressed with a 1 Hz bandwidth, the signal within the 1 Hz bandwidth becomes a value which is lower than it by approx. 10 log 1 Hz/1 kHz (dB), approx. 30 dB. It is then expressed as -100 dB/Hz apart from the carrier by 20 kHz when the IF bandwidth is 1 kHz.

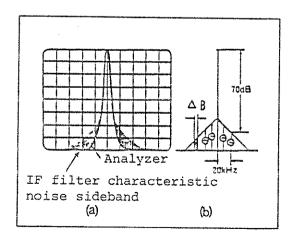


Figure A-2 Noise Sideband

Bandwidth Selectivity

The characteristic of a band-pass filter is not the so-called rectangular characteristic, but it is generally given an attenuation characteristic like a gauss distribution. When two large and small signals are mixed close by, the small signal is concealed behind the large signal (Figure A-3). It is therefore necessary to define the bandwidth in a certain attenuation area (60 dB). For this purpose, the ratio of 3 dB width vs. 60 dB width is expressed as the bandwidth selectivity.

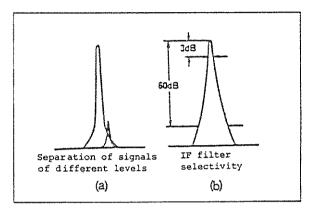


Figure A-3 Bandwidth Selectivity

Bandwidth Accuracy

This is the performance to express the bandwidth accuracy of the IF filter. It is expressed as a deviation of the nominal value at a 3 dB lowering point. Although this performance little affects the level measurement of ordinary continuous signals, it should be taken into consideration for the level measurement of a noise signal.

Bandwidth Switching Accuracy

For dissolving a signal into spectrums, not one but several IF filters are used to obtain the optimum resolution for the scan width. Even when measuring the same signal, an error occurs when the IF filter is switched for a portion having different loss. This is defined as the bandwidth switching accuracy.

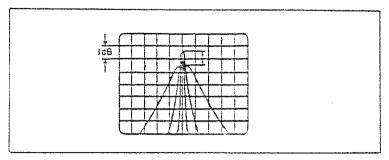


Figure A-4 Bandwidth Switching Accuracy

Reference Level Display Accuracy

In the spectrum analyzer, the absolute level of an input signal is obtained by reading how much the dB is lowered from the upper-most scale on the tube surface as a standard. The level set on this upper-most stage is called the reference level. The reference level is changed by the IF GAIN key and input attenuator and it is expressed in dBm or dB μ . The absolute accuracy of this display becomes the reference level frequency.

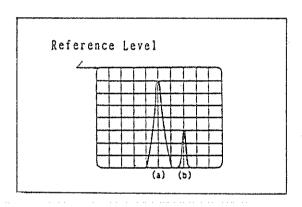


Figure A-5 Reference Level

VSWR: Voltage Standing Wave Ratio

This is a constant which expresses the impedance matching status. It is expressed as the ratio of the maximum value vs. minimum value of the standing wave caused by the composition of the progressive wave and reflected wave, where the spectrum analyzer is loaded to the ideal and nominal impedance source. This is expressed in a different form by the reflection coefficient and reflection loss.

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A.1 Explanation of Terminologies

When signal E_0 sent from the transmission side is completely transmitted to the reception side (the spectrum analyzer input section) without miss-matching in the impedance in Figure A-6, signal E_1 received in the reception side is equivalent in value to E_0 . When not all the signal is transmitted owing to the miss-matching on the reception side and returned by reflection to the reception side, the reflected ratio (the reflection coefficient) can be expressed as follows where the size of the reflected wave is taken as E_R :

Reflection coefficient m = Reflected wave E_R / progressive wave E_0

The ratio of reflected wave $\textbf{E}_{\textrm{R}}$ vs progressive wave $\textbf{E}_{\textrm{O}}$ becomes the reflected attenuation.

Reflected attenuation = 20 log
$$E_R$$
 / E_0 (dB) VSWR = $(E_0 + E_R)$ / $(E_0 - E_R)$

Its relation with the reflection coefficient becomes a range of 1 to in VSWR where the VSWR is assumed to be VSWR = (1+|m|) / (1-|m|). The closer to 1, the better the matching condition.

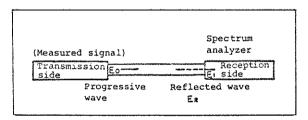


Figure A-6 V.S.W.R.

Spurious Response

When the signal level becomes larger, the harmonic wave is distorted in the input mixer circuit. A range usable with no distortion varies according to the fundamental wave input level. In the example in Figure A-7, it becomes -70 dB for the -30 dBm. When the input signal level is larger, the signal applied to the mixer is made smaller by the input attenuator so that it becomes an optimum input level.

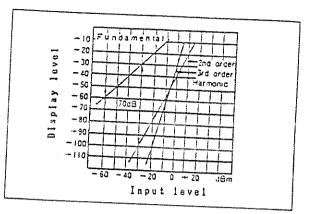


Figure A-7 Spurious Response

YIG-turned Oscillator

This was reported by Griffiths for the first time in 1946. The garnet-series ferrite which represents the (Yttrium Iron Garnet) monocrystal shows a quite sharp electronic spin resonant phenomenon and its resonant frequency has a linear proportional relationship throughout a broad frequency band for the applied DC magnetic field. It is known from this that the broad band electronic tuning is enabled by varying the exciting current of electromagnet which forms the AC magnetic field. This is applied to the spectrum analyzer and to the local sweep generator of the automatic microwave frequency counter of ADVANTEST.

A.2 Level Conversion Table

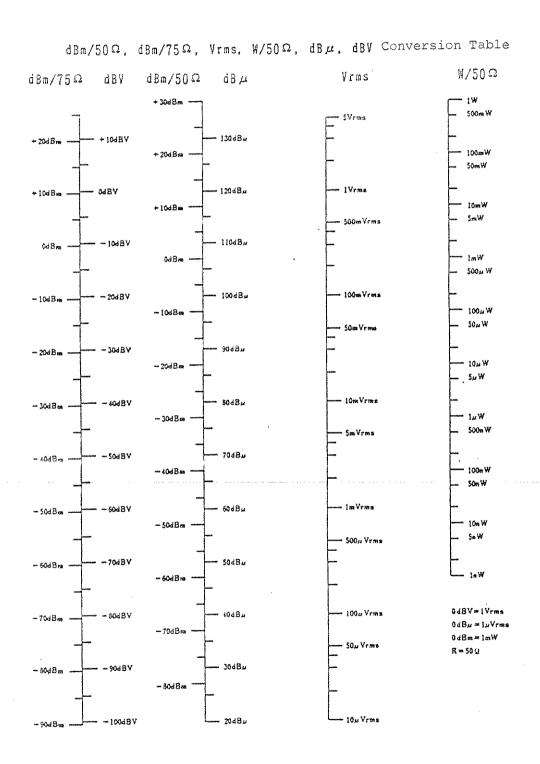


Figure A-8 Level Conversion Table

A4131 SERIES SPECTRUM ANALYZER INSTRUCTION MANUAL

A. I Part	s Location		- •	
	a nocation	200	Circuit	To 3
		C4 4 1 C4	OTT CHIE	1112ar:

A.3 Parts Location and Circuit Diagrams

A - 10

Parts No.	ADVANTEST Stock No.	Parts No.	ADVANTEST Stock No.
U34 U35	SIT-75160		
U36 -37	SIT-75161 SMM-2018B	The state of the s	
U38 U39	SIM-74HCO4 SIM-74HC74		
J40 J41	SMM-27128A SIM-8254C	7,000	
J42 J43	SMM-2864 SIM-74HC393		
J44 <1	SIM-74HC04 DXC-000109		
	57.0 000109	The second secon	
		The state of the s	
U. TYPYAAA MARAATINA MARAA			
			×
And the second second		Control of the Contro	
Andrivery and the second and the sec			
	100 mm		
THE PARTY OF THE P		-	
Laborate transmission		The second secon	
And the state of t		Management of the second of th	
V-04-			

arts No.	ADVANTEST Stock No.	Parts No.	ADVANTEST Stock No.	
4 5	JCP-BH010PX01	R4	DCD AUGOV	
5	JCR-AF010PX01	R 5	RCP-AH22K RCP-AH470K	
5 7 -9	JCP-BH003PX01	R6	DSP-000015	1
,>	JCF-ACOO1JXO1	R7	RCP-AH100	11
Ĺ	KRL-000874 LCL-C00554	R8	RCP-AH68	
2	LCL-C00490	R10 R11	RCP-AH470	•
\$	LCL-C00673	R12	RMF-AC470QFJ RCP-AH100	1
F	LCL-CC0490	R13	RCP-AH15	
;	LCL-C00124	R14	RCP-AH33	
,	LCL-C00012	R15	RCP-AH22K	- 1
}	LCL-C00010 LCL-C00672	R16	RCP-AH4R7K	ĺ
)	LCL-C00010	R17 R18	RMF-AC100QFJ	
0	LCL-C00672	R19	RCP-AH560 RMF-AC1KFJ	1
1	LCL-C00012	R20	RCP-AH3R9K	Į.
2 -13 4 -15	LCL-C00549	R21	RCP-AH2R2K	
6	LCL-C00012 LCL-B01024	R22	RCP-AH18K	
8	LCL-801024	R23	RCP-AH15	
0 -22	LCL-C00012	R24	RCP-AH33	
3 -24	LCL-C00549	R27	RCP-AH4R7K RCP-AH22K	
5 -26	LCL-C00010	R28	RMF-AC150QFJ	
7 -28	LCL-C00672	R29	RCP-AH560	, .
0 -32	LCL-C00554	R30	RMF-AC1KFJ	
3	LCL-C00012 LCL-B01024	R31	RCP-AH3R9K) '
5	LCL-B01024	R32	RCP-AH2R2K	7
9 -44	LCL-T00084A	R33	RCP-AH18K RCP-AH470	
	SFN-SST4859	R36	RCP-AH330	į
	STN-2SC1815	R37	RMF-AC1KFJ	1.
	STN-2SC2712	R38	RCP-AH100	
-10	STN-2SC1815 STN-2SC2712	R39	RCP-AH10K	:
l	STN-2SC1815	R40	RCP-AH100	4
5	STN-2SC2712	R41 -42 R43 -44	RCP-AH2R2K RMF-AC2R2KFJ	
5	STN-2SC1815	R45	RCP-AH6R8K	
, -27	STN-2SC2712	R46	RCP-AH3R3K	
)	STN-FN1A4P STP-2SA1162	R47	RCP-AH750	
32	STP-25A1162 STP-25A1015	R48	RCP-AH220	
,	STN-2SC1815	R49 R50 -52	RCP-AH56	
· -35	SFN-SST4393	R53	RCP-AH120 RCP-AH390	
-37	STN-2SC1983	R54	DSP-000017	
,	STN-FA1A4P	R55	RCP-AH470	
	STP-2SA1162 STN-2SC2712	R56	RCP-AH100	
	STN-25C2/12 STN-2SC1815	R57 -58	RCP-AH2R2K	
-44	STP-2SA1015	R59 -60 R61	RMF-AC2R2KFJ	
	RCP-AH39	R62	RCP-AH6R8K RCP-AH3R3K	
	RCP-AH56	R63	RCP-AH750	
	RCP-AH10K	R64	RCP-AH220	

Parts No.	ADVANTEST Stock	No. Parts	No. ADVANTEST Stock	c No.
R65 R668 R774 R777 R780 R882 R883 R884 R877 R780 R882 R887 R890 R891 R92 R93 R94 R995 R997 R998 R900 R1003 R1004 R1005 R1007 R1007 R110 R1112 R112 R112 R112 R112 R112 R11	RCP-AH156 RCP-AH1500 RCP-AH157 RCP-AH157 RCP-AH157 RCP-AH157 RCP-AH157 RCP-AH157 RCP-AH157 RCP-AH157 RCP-AH157 RCP-AH157 RCP-AH157 RCP-AH157 RCP-AH157 RCP-AH157 RCP-AH157 RCP-AH157 RCP-AH157 RCP-AH168	R13333334444445555689012345678901234567890123456789012345678901234567890123456777777777777777777777777777777777777	RCP-AH6R8K RCP-AH3R3K RCP-AH1R2K RCP-AH680 RCP-AH56 RCP-AH560 RCP-AH560 RCP-AH560 RCP-AH100 RCP-AH2R2K RMF-AC2R2KFJ RCP-AH6R3K RCP-AH6R3K RCP-AH6R3K RCP-AH156 RCP-AH15 RCP-AH15 RCP-AH15 RCP-AH15 RCP-AH15 RCP-AH15 RCP-AH15 RCP-AH15 RCP-AH22K RCP-AH560 RMF-AC150QFJ RCP-AH22K RCP-AH22K RCP-AH15 RCP-AH1680 RMF-AC150QFJ RCP-AH18K RCP-AH2R2K RCP-AH22CK RCP-AH220K RCP-AH28C RCP-AH28C RCP-AH28C RCP-AH28C RCP-AH28C RCP-AH280 RCP-AH280 RCP-AH280 RCP-AH270 RMF-AC680QFJ RCP-AH390 RCP-AH470 RMF-AC470QFJ RCP-AH10K	

Parts No.	ADVANTEST Stock No.	Parts No.	ADVANTEST Stock No.
R196-203	RCP-AH47K		
R204-205		R286	RMF-AC8R2KFJ
R206-211		R288	RMF-AC6R2KFJ
R212	RCP-AH150	R290	NUF-ADIUK
R213	RCP-AH62K	R291	RCP-AH220
R214-219	RCP-AH220	R292	RCP-AH1K
R220	RCP-AH1R5K	R293	RCP-AH100
R221	RCP-AH47K	R294	RCP-AH270
R222	RCP-AH51	R295-298	
R223-224	RCP-AH39	R299	RCP-AH4R7K
R227-229		R300	RCP-AH1K
3230	RCP-AH3R3K	R301	RMF-BJ30KFJ
R231	RMF-AC4R7KFJ	R302	RMF-BJ30KFJ RMF-BJ15KFJ
233		R303-304	RMF-BJ7R5KFJ
2234	RMF-AC8R2KFJ	R305	RMF-BJ10KFJ
235-237	RMF-AC510QFJ RCP-AH10K	R306	RMF-BJ1KFJ RMF-BJ110QFJ
₹238	RCP-AHIOK RCP-AH3R3K	R307	
₹239		R309	RMF-BJ10KFJ
₹241	RMF-BJ8R2KFJ	R310	RMF-BJ220QFJ
₹242-243	RMF-BJ6R8KFJ	R311	RMF-BJ220QFJ RMF-BJ2R7KFJ
₹244	RMF-BJ10KFJ	R312	RMF-BJ7R5KFJ
₹245	RCP-AH10K	R313	RMF-BJ15KFJ
1246-247	RCP-AH3R3K	R314	RMF-BJ15KFJ RMF-BJ7R5KFJ RMF-BJ11KFJ
1248-249	RMF-BJ33KFJ	R315	RMF-BJ11KFJ
1250	RCP-AH47K	R316	RMF-BJ56QFJ
251	RCP-AH18K	R318	RMF-BJ4R7KFJ RMF-BJ3KFJ
1252	RCP-AH5R1K	R320	RMF-BJ3KFJ
254	RCP-AH10K	R321	RCP-AH100K
:255	RCP-AH3R3K	R322	RMF-BJ5R1KFJ
256	RCP-AH15K	R323	RCP-AH1R5K
258	RCP-AH1OK	R324-332	RCP-AH220
259	RMF-BJ5R1KFJ	R333-335	RMF-AS330QFK
260	RMF-BJ10KFJ	R336	RMF-BJ20KFJ
262	RMF-BJ68KFJ	R337	RMF-BJ10KFJ
263	RCP-AH820K	R338-340	RCP-AH680
264	RCP-AH1M	R341	RMF-BJ10KFJ
265	RMF-BJ5R6KFJ	R342	RMF-BJ12KFJ
266	RMF-BJ100QFJ	R343	RMF-BJ10KFJ
267	RCP-AH10K	R344	RMF-BJ12KFJ
268	RCP-AH12K	R345	RCP-AH12K
271	RCP-AH1K	R346	RCP-AH3R3K
272	RMF-BJ11KFJ	R347	RMF-BJ18KFJ
273	RMF-BJ56QFJ	R348	RMF-BJ12KFJ
273	RMF-BJ10KFJ	R349	RMF-BJ18KFJ
275	RMF-AC10KFJ	R350	RMF-BJ7R5KFJ
276	RMF-AC7R5KFJ	R351	RMF-BJ51KFJ
277	RMF-BJ330KFJ	R352	RMF-BJ15KFJ
1	RMF-BJ1KFJ	R353	RMF-BJ7R5KFJ
1	RMF-BJ30KFJ	R354	RMF-BJ7R5KFJ
1	RMF-BJ1KFJ	R356	RCP-AH3R3K
	RMF-BJ5R1KFJ	U1 -2	SHB-001655
1	RCP-AH220	U3 -4	SHB-001658
-02	RCP-AH1K	U 5	SHB-001656
			0.0000000000000000000000000000000000000

Parts	No.	ADVANTEST Stock No.	Parts No.	ADVANTEST Stock No.
112		SHB-001657		
U6 U8	-7	SHB-001544		
	-10	SHB-001543		
U11		SHB-001544		
	-14	SHB-001655 SHB-001656		
U15 U16	-17	SHB-001657		
U18		SHB-001655		
U19	-20	SHB-001658		
U21		SHB- 001 655 SHB-001656		
U22 U23		SHB-001636 SHB-001543		
U24		SHB-001544		
U25		SHB-001543	A CONTRACTOR OF THE CONTRACTOR	
U26		SHB-001544		
U29 U33	-32	SIM-74HC138 SIM-74HC273		,
U34		SIM-74HC174	WAR AND	
U35		SIM-74HC273		
U36		SIM-74HC174		
U38	-40	SIM-74HC74 SIM-74HC4538		
U41 U42		SIM-74HC04	A CANADA A A CANADA A	
U43	-45	SIT-74LS06		
U46		SIA-4558		
U47		SIA-324	And the state of t	
U48		SIM-74HC273 SIA-OP77P		
U52		SIA-TL082		
3	-54	SIA-4558		
U55	5 0	SIA-393		
1	-58 -62	SIA-DA7524-4 SIA-DG201		
	-65	SIA-OP77P		
U66		SIA-TL072	THE PARTY OF THE P	
U67		SIA-811		
U68		SIA-811 SIA-TL072		
U70		SIA-812	all the same of th	
U71		SIA-4558		
U72		SIA-398		
U73		SIA-DG201 SIA-4558		
U74		SIA-4558		
U76		SIA-811		
X1	-4	DXD-001059		
			Table 1	

rts No.	ADVANTEST Stock No.		Parts i	No.	ADVANTEST Stock No	o.
1 -4	CCD DAROALES					
7	CCP-BAR01U50V CCP-BBR1U50V		C104-1		CCK-CD10U25V	
-11	CCP-BAR01USOV		C106-1	.07	CCP-BBR1U50V	
.2	CMC ADZZADAZU		C108		CCP-BA33P5OV	
3 -15	CMC-AP330PR5K		C109		CFM-AS1000P50V	
6 -17	CCP-BAR01U50V		C110-1	11	CCK-CD10U25V	
7 -24	CMC-AP470PR3K		C112-1	15	CCP-BBR1U50V	
7 -28	CCP-BARO1U50V	1	C116		CCP-BA33P50V	
, -20 9	CCP-BBR1U50V CCP-BA15P50V		C117		CFM-AS2200P50V	
Ó	CCP-BBR1U5OV		C118-1		CCK-CD10U25V	
1 -32	CCP-BAR01U50V		0120-1		CCP-BBR1U50V	
3 -37	CCP-BBR1U5OV		0122-1		CCP-BA1000P50V	
3	CTA-AC10U16V		2124-1		CCP-BBR1U50V	
7	CTA-AC1U35V	(126-1	27	CCP-BA1000P50V	
Ó			128-1	29	CCP-BBR1U5OV	
	CFM-ASRO1U5OV CMC-AP100PR5K		2130		CCP-BA100P50V	
2	CCP-BA330P50V		131	ļ	CCK-CD10U25V	
·	CFM-AHR47U100V		132		CCP-BA47P50V	
-45	CCP-BBR1U50V		133		CCK-CD22U25V	
-47	CTA-AC10U16V		134-1		CCP-BBR1U50V	
-49	CCP-BAR01U50V		141-1		CCK-CD47U25V	
-55	CCP-BBR1U50V	C	149-1	50	CCK-CD47U1OV	
	CCP-BAR01U5OV	C	151-1	92	CCP-BBR1U5OV	
	CCP-BAR01050V CCP-BA15P50V	D	1 -2	*	SDS-1SS270	
	CCK-CD22U16V	D	3 -4		SDS-1SS286	
	CCP-BBR1U5OV	D	5 -9		SDS-1SS270	
	CFM-AH1U100V	D	10 .		SDS-1SS286	
-66	CCP-BBR1U50V	12	11		SDS-LD1	
00	CCP-BARO1U5OV	17	12 -1	7	SDS-1SS270	
	CFM-ASRO22U50V	- 11	20		SDZ-MO30	
-	CCP-BBR1U5OV	12	21 -2		SDS-LD1	
	CCP-BA1000P50V	15	24 -3		SDS-1SS270	
-72	CCP-BBR1U5OV	. 16	35	L L	SDZ-M051	
' -	CCK-CD2R2U5OV	- 11	36 -39		SDS-1SS270	
-75	CCK-CD220U25V	13	41 -45		SDS-1SS270	
	CCP-BBR1U50V	11	47		SDS-LD1	
	CCK-CD10U25V	11	48 -50	0 :	SDS-1SS270	
	CCP-BBR1U5OV	18	52		SDS-LD1	
	CCK-CD10U16V	12	53 -50	6 :	SDS-1SS270	
-81	CCP-BBR1U5OV	- 11	59 -60	0 :	SDZ-MO51	
	CCP-BA1000P50V	13	61 -62	2 3	SDS-1SS286	
	CCP-BA220P50V	J			JCR-AF050PX02	
1	CCP-BA1000P50V	J			JCP-BH002PX02	
-86	CCP-BBR1U5OV	J			JCP-BH010PX02	
-95	CCP-BBR1U50V	J			JCF-ACOO1JXO1	
	CCP-BA47P5OV	L		L	LCL-T00084A	
	CCK-CD22U25V	L.			LCL-C00014	
	CCP-BBR1U5OV	Q			STN-2SC2757	
	CCP-BA330P50V	Q:	-		STN-2SC2712	
-101	CCP-BBR1U50V	Q			STN-2SC2757	
	CCP-BA33P50V	Q:	-		STP-2SA1462	
	CFM-AS1000P50V	Q			STN-FA1A4P	
1	ALL MATORIAN	Q:	10 -11		STN-2SC2757	

Parts No.	ADVANTEST Stock N	0.	Parts No	o. ADVANTEST Stock	< No
R115	Den Allena				···········
R116	RCP-AH2R2K		R188	RCP-AH4R7K	
R117	RCP-AH47K		R189	RCP-AH15K	
	RCP-AH10K		R190		
R118	RCP-AH220		R191	RCP-AH1K	
R119	RCP-AH1M		R192	RCP-AH180K	
R120 -121	RCP-AH1OK		R193-19	RCP-AH1K	
R122	RCP-AH1K			DOCENT	
R123	RCP-AH150	J	R197-19	1 11141111	
2124 - 127	RCP-AH680		R200	RCP-AH470	
R128	RCP-AH1K	[R201-20	2 RCP-AH10K	
R131			R203	RCP-AH4R7K	
2132	RCP-AH47K		R205	RCP-AH47K	
133	RCP-AH1OK		R206	RCP-AH39K	
j	RCP-AH3R9K		R207-218	8 RCP-AH47K	
2134 -135	RCP-AH3R3K		R232	1111-11	
2136 -137	RCP-AH1OK		R234	RMF-BJ4R7KFJ	
138	RCP-AH100K	13	R235	RCP-AH1R8K	
139 -140	RCP-AH1M	1		RCP-AH4R7K	
141	RCP-AH200K		R236	RCP-AH22	
142 -143	RCP-AH1M		R237-238	RMF-BJ10KFJ	
144	RCP-AH200K	F	R239	RCP-AH10K	
145			R240	RCP-AH1K	
146 -147	RCB-AK10M		R242	RCP-AH2R2K	
	RCP-AH27K		R243	RCP-AH100	
149 -150	RCP-AH10K	11	R244	RCP-AH6R8K	
151	RCP-AH270K		R245		
152	RCP-AH47K		246	RCP-AH150	
153 -156	RCP-AH10K	11	R247	RCP-AH6R8K	
157	RCP-AH330	11		RCP-AH150	
158	RCP-AH1K		248-249	1	
159	RCP-AH220	7.1	250	RCP-AH1K	
160	RCP-AH15K		251-252	RCP-AH180	
161	RCP-AH10K		253	RCP-AH82K	
162		R	254-255	RCP-AH2R2K	
163	RMF-BJ10KFJ	R	256	RCP-AH4R7K	
64	RMF-BJ12KFJ	R	257	RCP-AH1K	
	RMF-BJ5R6KFJ		259		
.65	RMF-BJ2R2KFJ	[1	260	RCP-AH2R2K	
.66	RCP-AH1M	: :	261	RCP-AH100	
67	RCP-AH180K		262	RCP-AH6R8K	
168	RCP-AH220K			RCP-AH150	
.69	RCP-AH270K	12	263	RCP-AH6R8K	
70-171	RCP-AH15K		264	RCP-AH150	
.72	RCP-AH100K		265-266	RCP-AH33	
.73	RCP-AH3R9K		267-268	RCP-AH180	
74-175		R	269	RCP-AH100K	
76	RCP-AH100K	R	270	RCP-AH3R3K	
77	RCP-AH47K	- 1	271	RCP-AH2R2K	
78	RCP-AH100K		272		
*	RMF-BJ10KFJ	11	273	RCP-AH4R7K	
79	RCP-AH47K		274	RCP-AH1K	
80	RCP-AH10K			RCP-AH100	
81	RCP-AH180	11	276	RCP-AH100	
82-184	RCP-AH47K	11	278	RCP-AH2R2K	
85	RCP-AH100		79	RCP-AH100	
86-187	RCP-AH47K	11	280	RCP-AH6R8K	
	NOF-884/K	R 2	81	RCP-AH150	
1		12			

Parts No.	ADVANTEST Stock No.	Parts No.	ADVANTEST Stock No.
R282 R283 - 285 R284 - 285 R286 - 288 R287 - 288 R2890 - 294 R2992 - 300 R2992 - 300 R3005 - 306 R3005 - 313 R311 - 9 U111 - 16 U113 - 16 U113 - 22 U224 - 337 U117 U123 U234 - 34 U338 - 37 U345 - 42 U446 - 42 U447 U448 - 51 U448 U456 - 51 U455 - 42 U448 U456 - 51 U557 - 42 U448 U456 - 51 U556 - 7	RCP-AH150 RCP-AH33 RCP-AH1K RCP-AH180 RCP-AH180 RCP-AH3R3K RCP-AH3R3K RCP-AH4R7K RCP-AH10K RCP-AH3R3K RCP-AH10K RCP-AH15K RCP-AH15K RCP-AH15K RCP-AH15K RCP-AH15K RCP-AH15K RMF-BJ10KFJ RCP-AH15K RMF-BJ1R2KFJ RCP-AH16 SHA-TL072 SIA-TL072 SIA-TL072 SIA-TL072 SIA-TL072 SIA-TL072 SIA-TL072 SIA-TL072 SIA-TL072 SIA-TL073 SIM-74HC03 SIM-74HC03 SIM-74HC03 SIM-74HC174 SIA-4066 SIM-74HC174 SIA-4066 SIM-74HC174 SIT-DN8650 SIT-74LS06 SIM-74HC175 SIM-74HC125 SIM-74HC125 SIM-74HC02 SIM-74HC02	U58 -64 -62 U63 -66 U63 -66 U667 U068 U771234 U77778 U77778 U107778 U107778 U10883 U10883 U10883	SIA-2525D SIA-393 SIA-311N SIM-74HC74 SIM-74HC139 SIM-74HC00 SIM-74HC08 SIA-6012 SIA-311N SIM-74C905 SIM-74HC08 SIM-74HC07 SIM-74HC107 SIM-74HC107 SIM-74HC107 SIM-74HC175 SIM-74HC04 SIM-74HC00 SIA-DG201

Parts No.	ADVANTEST Stock No.	Parts No.	ADVANTEST Stock No.
C1 -4	CCP-BARO1U5OV `	C103	
C5 -7	CCP-BBR1U5OV	C104-105	CFM-AS1000P50V
C8 -11	CCP-BAR01U5OV	C112-113	1 00 20 0 20 0
C12	CMC-AP330PR5K	C130	
C13 -15	CCP-BARO1U5OV	C131	CCP-BA100P50V
016	CMC-AP470PR3K	C132	CCK-CD10U25V
C17 -24	CCP-BAR01U5OV	C133	CCP-BA47P50V
C25 -26			CCK-CD22U25V
027 -28	CCP-BBR1U5OV	C134-136	
029	CCP-BA15P5OV	C141-148	
030	CCP-BBR1U5OV	C149-150	
031 -32	CCP-BAR01U50V	C151-193	CCP-BBR1U50V
C33 -37	CCP-3BR1U5OV	C194	CFM-AS2200P50V
C38	CFM-ASRO22U5OV	D1 -2	SDS-1SS270
C39		D3 -4	SDS-1SS286
C40	CFM-AS2200P50V	D5 -9	SDS-1SS270
C41	CMC-AP820PR3K	D10	SDS-1SS286
C42	CMC-AP220PR5K	D11	SDS-LD1
C42	CCP-BA330P50V	D12 -13	SDS-1SS270
	CFM-AHR47U100V	D15 -17	SDS-188270
C44 -45	CCP-BBR1U5OV	D20	SDZ-M030
C46 -47	CTA-AC10U16V	D21 -23	SDS-LD1
C48 -49	CCP-BAR01U50V	D24 -34	SDS-1SS270
C50 -55	CCP-BBR1U5OV	D35	SDZ-M051
C56	CCP-BAR01U50V	D36 -39	SDS-1SS270
C57	CCP-BA15P5OV	D41 -45	
C61	CCK-CD22U16V	D47 -43	SDS-1SS270
C62	CCP-BBR1U5OV	D60	SDS-LD1
C63	CFM-AH1U100V	11	SDZ-MO51
C64 -66	CCP-BBR1U50V	D61 -62	SDS-1SS286
C67	CCP-BAR01U5OV	J1	JCR-AF050PX02
C68	CFM-ASR022U5OV	J2	JCP-BH002PX02
C69	CCP-BBR1U5OV	J3	JCP-BH010PX02
C70	CCP-BA1000P50V	J4	JCF-ACOO1JXO1
C71 -72	CCP-BBR1U5OV	L2 -4	LCL-T00084A
C73	CCK-CD2R2U5OV	L5 -6	LCL-C00014
C74 -75	CCK-CD220U25V	Q1	STN-2SC2757
C76	CCP-BBR1U50V	Q2 -5	STN-2SC2712
77	CCK-CD10U25V	Q6	STN-2SC2757
78		Q7 -8	STP-2SA1462
79	CCP-BBR1U5OV	Q9	STN-FA1A4P
80 -81	CCK-CD10U16V	Q10 -11	STN-2SC2757
82	CCP-BBR1U5OV	Q12	SFN-SST4859
1	CCP-BA1000P50V	Q13 -14	STN-2SC2712
83	CCP-BA22OP5OV	Q15	SFN-SST4393
84	CCP-BA1000P50V	Q16	STP-2SA1162
85 -86	CCP-BBR1U50V	Q17	STN-2SC2712
91 -95	CCP-BBR1U5OV	Q19	STN-2SC2712
96	CCP-BA47P5OV	Q20	STP-2SA1162
97	CCK-CD22U25V	Q21	
98	CCP-BBR1U5OV	13	SFN-SST4393
99	CCP-BA330P50V	Q22.	STN-2SC2712
10C-101	CCP-BBR1U5OV	Q23	STP-2SA1162
102	CCP-BA33P5OV	Q24	SFN-SST4393
[Q25 -31	STN-2SC2712

BLR-015117X02 (2/4)

Parts No.	ADVANTEST Stock No.	Parts No.	ADVANTEST Stock No.
QQQQQQQQQQQQQQQQQQQQQQQQQQQQQQQQQQQQQQ	STP-2SA1162 SFN-SST4859 STN-2SC2712 STP-2SA1162 SFT-2SA1162 SFT-SST406S STP-2SA1162 RCP-AH82 RCP-AH15K RCP-AH15K RCP-AH185 RCP-AH185 RCP-AH180 RCP-AH181 RCP-AH150 RCP-AH150 RCP-AH150 RCP-AH150 RCP-AH150 RCP-AH150 RCP-AH150 RCP-AH150 RCP-AH150 RCP-AH12K RCP-AH12K RCP-AH12K RCP-AH12K RCP-AH12K RCP-AH160 RCP-AH182 RCP-AH170 RCP-AH180 RCP	R678 R71	RMF-BJ6R8KFJ REE-AR510-1 RCP-AH3R9K RCP-AH15K RMF-BJ15KFJ RMF-BJ20KFJ RMF-BJ68KFJ RCP-AH15K RCP-AH110K RCP-AH100 RCP-AH2R2K RCP-AH2R2K RCP-AH10K RCP-AH10K RCP-AH10K RCP-AH220 RCP-AH220 RCP-AH1M

Parts No.	ADVANTEST Stock No.	Parts No.	ADVANTEST Stock No.
R124-127	RCP-AH680	R197-199	RCP-AH4R7K
R128	RCP-AH1K	R200	RCP-AH470
R129	RCP-AH100K	R201-202	RCP-AH10K
R130	RMF-BJ680QFJ	R203	RCP-AH4R7K
R131	RCP-AH47K	R205	RCP-AH47K
1	RCP-AH10K	R205	RCP-AH39K
R132	RCP-AH3R9K	R207-211	RCP-AH47K
2133	RCP-AH3R3K	R213-218	RCP-AH47K
2134-135	RCP-AH10K	;i	RMF-BJ4R7KFJ
(136 - 137		R232	RCP-AH1R8K
2138	RCP-AH100K	R234	
139-140	RCP-AH1M	R235	RCP-AH4R7K
8141	RCP-AH2OOK	R236	RCP-AH22
3142-143	RCP-AH1M	R237-238	RMF-BJ10KFJ
144	RCP-AH2OOK	R239	RCP-AH10K
2145	RCB-AK10M	R240	RCP-AH1K
2146-147	RCP-AH27K	R242	RCP-AH2R2K
R149-150	RCP-AH10K	R243	RCP-AH100
2151	RCP-AH270K	R244	RCP-AH6R8K
2152	RCP-AH47K	R245	RCP-AH150
R153-156	RCP-AH1OK	R246	RCP-AH6R8K
2157	RCP-AH330	R247	RCP-AH150
R158	RCP-AH1K	R248-249	RCP-AH33
R159	RCP-AH220	R250	RCP-AH1K
1	RCP-AH15K	R251-252	RCP-AH180
R160	RCP-AH10K	11	RCP-AH82K
R161	RMF-BJ10KFJ	R253	RCP-AH2R2K
R162		R254	RCP-AH1K
R163	RMF-BJ12KFJ	R255	
R164	RMF-BJ5R6KFJ	R256	RCP-AH4R7K
R165	RMF-BJ2R2KFJ	R304	RCP-AH15K
R166	RCP-AH1M	R305-306	
R167	RCP-AH180K	R307	RCP-AH22
R168	RCP-AH220K	R309	RMF-AC16KFJ
R169	RCP-AH270K	R311	RMF-BJ1R2KFJ
R170-171	RCP-AH15K	R312-313	RCP-AH1K
R172	RCP-AH100K	R314	RMF-BJ3KFJ
R173	RCP-AH3R9K	R315	RMF-BJ2KFJ
R174-175	RCP-AH100K	R318	RCP-AH22
R176	RCP-AH47K	U1 -9	SHB-001464
R177	RCP-AH100K	U10	SIA-318C
R178	RMF-BJ10KFJ	U11	SIA-TL072
R179	RCP-AH47K	U12	SIA-318C
R180	RCP-AH10K	U13 -16	SIA-TL072
R181	RCP-AH180	U17	SIA-HA1127
	RCP-AH47K	13 .	SIA-4558
R182-184	RCP-AH100	U18	SIA-4066
R185	RCP-AH47K	U19	·
R186-187	RCP-AH4R7K	U20	SIA-4558
R188		U21 -22	SIA-TL082
R189	RCP-AH15K	U23	SIA-4558
R190	RCP-AH1K	U24	SIA-393
R191	RCP-AH180K	U25	SIM-74HC4538
R192	RCP-AH1K	U26	SIM-74HCO3
R193-196	RMF-BJ22KFJ	U27	SIM-74HCOO

BLR-015117X02 (4/4)

Parts No. ADVANTEST S	Stock No. Parts No	. ADVANTEST Stock No.
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Parts No.	ADVANTEST Stock No.	Parts No.	ADVANTEST Stock No.
D1 D2 -11 D12 D13 -17 D18 -75 J1 R1 -17 S1 -29	NLD-000111 NLD-000010		
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Parts No.	ADVANTEST Stock No.	Parts No.	ADVANTEST Stock No.
C1	CCP-BA100CP50V CCP-BA1P50V CCP-BA7P50V CCP-BA7P50V CCP-BA10OP50V CCP-BA10OP50V CCK-CD22U16V CCP-BAR01U50V CCK-CD10U25V CCP-BA15P50V CCP-BA5P50V CCP-BA7P50V CCP-BA7P50V CCP-BA7P50V CCP-BA7P50V CCP-BA7P50V CCP-BA7P50V CCP-BA10OP50V CCP-BA10OP50V CCP-BA10OP50V CCP-BA33P50V CCP-BA3	L150 -2 3 -2 3 -2 3 -3 -9 23 L150 M1 24561234567 60112346789012345689012345689012345678112346788888888888888888888888888353512	LCL-C00010 LCL-A00066 LCL-A00066 DEE-000736 STN-2SC2759 STN-2SC2757 STN-2SC2757 STN-2SC2757 STP-2SA1226 RCP-AJ56 RCP-AJ56 RCP-AJ33 RCP-AJ220 RCP-AJ33 RCP-AJ470 RCP-AJ56 RCP-AJ56 RCP-AJ56 RCP-AJ56 RCP-AJ586K RCP-AJ586K RCP-AJ586K RCP-AJ586K RCP-AJ586 RCP-AJ10K RCP-AJ10K RCP-AJ10C R

Parts No.	ADVANTEST Stock No.	Parts No.	ADVANTEST Stock No.
C1 K1 -3 R1 R2 R3 R4	CCP-ADR47U5OV KRL-000350 RCP-AM91 RCP-AM68 RCP-AM91 RCP-AM62		
R5 R6 R7 −8 R9 R10	RCP-AL120 RCP-AL130 RCP-AM62 RCP-AL120 RCP-AL130 RCP-AM62		

BTB-015120

Parts No.	ADVANTEST Stock No.	Parts No.	ADVANTEST Stock No.
D1 R1 -3	SDS-DMJ4317-1 RCP-AJ100		
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Parts No.	ADVANTES	3T Stock	No.	Parts No.	ADVANTEST Stock No.
R1 -2	RCP-AJ100				
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Parts No.	ADVANTEST Stock No.	Parts No.	ADVANTEST Stock No.
C1 -2 C3 -4 C5 C5 -8 C9	CCP-AC100P50V CCP-ACR01U50V CCP-ADR47U50V CCP-ACR01U50V CCP-AC15P50V		
C10 C11 -12 C13 D1 L1 L3 L4 L5 L6	CCP-AC1000P50V CCP-ACR01U50V CCP-AC2P50V SDS-ND587T LCL-E00932 LCL-A00671 LCL-E00934 LCL-E00939 LCL-E00388		
Q1 Q2 Q3 R1 -2 R3 R4 R5 R6 R7 R8	SFN-2SK571 STN-2SC2585 STN-2SC3356 RCP-AJ100 RCP-AJ82 RCP-AJ1K RCP-AJ100K RCP-AJ2R7K RCB-AG10K RCP-AJ100 RCP-AJ62		
R10 R11 R12 -13 R14 R15 R16 R17 R18 R19	RCP-AJ100 RCP-AJ680 RCP-AJ2R2K RCB-AQ330 RCP-AJ10K RCP-AJ3R3K RCP-AJ8R2 RCP-AJ220 RCP-AJ180		
V1 Y1 -2 Y3	SHB-001697 DXD-000792 DXD-001050		
			·

Parts No.	ADVANTEST Stock No.	Parts No.	ADVANTEST Stock No.
R1 R2	DMF-001496 DCB-FF1223X03-1 DCB-FF1223X12-1 DCB-FF2023X26-1 DCB-FF2680X15-1 DCB-Q2805X01-1 DCB-Q2805X01-1 DCB-QF2802X01-1 DCB-QF2803X01-1 DCB-QF2804X01-1 DCB-QF2801X01-1 DCB-QF2801X01-1 DCB-QF2801X01-1 JCI-AF003JX05-3 JCF-AB001JX03 JCS-AV004JX01 JCD-AV003PX01 DEE-001427 JTE-AG001EX01 RVR-BA10K RVR-BL200K AAA-ME5813A		

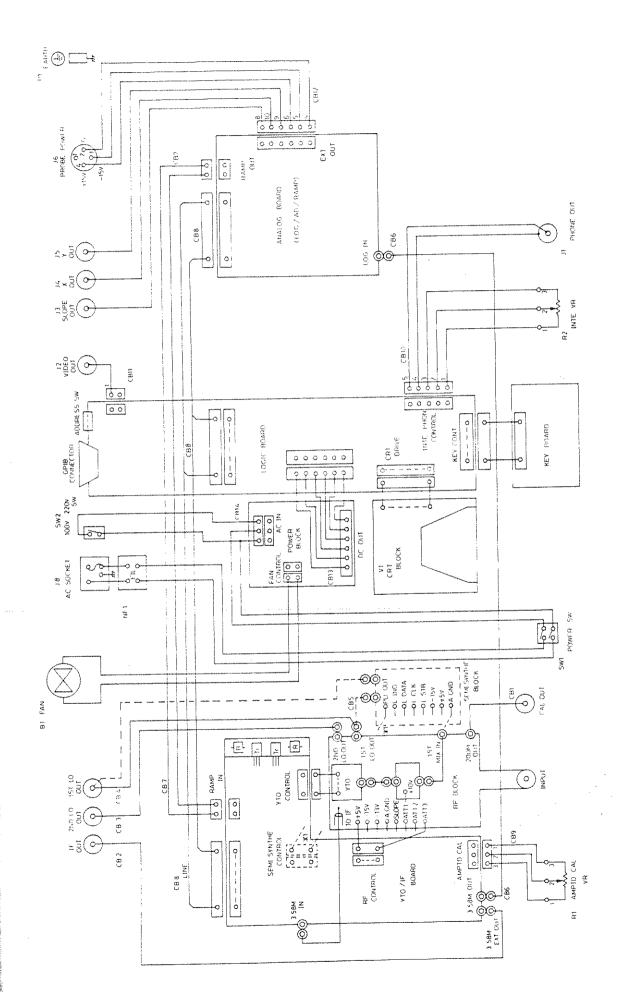
Parts No.	ADVANTEST Stock No.	Parts No.	ADVANTEST Stock No.
FL1 -7 J1 -2 J3	DNF-001052 JCF-AA001JX01 YEE-000868-1		
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Parts No.	ADVANTEST Stock No.	Parts No	. ADVANTEST Stock No.
FL1 -10 J1 J2 J3 J4 J5 -6 J7 -9 J11 J12	DNF-001052 JCF-AF001JX09-1 JCF-AA001JX39-1 JCF-AA001JX01 JCF-AA001JX06-1 JCF-AA001JX01 JCF-AC001JX02 JCR-AE010JX02 JCS-BZ010JX01		
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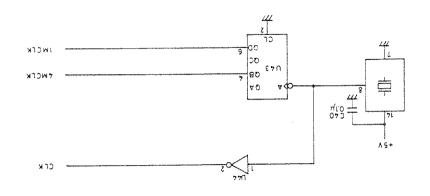
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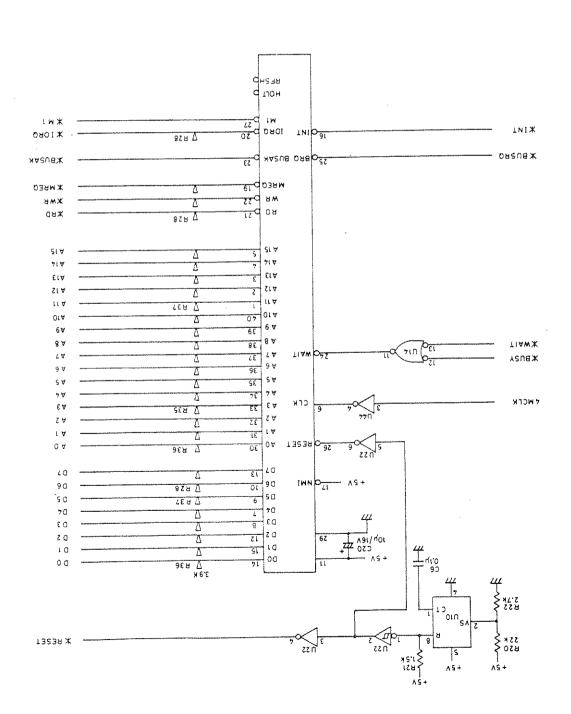
CB1
J2 J3 J6-5 J6-5 J8 NF1 DEE-001427 P1 R1 RVR-BA10K R2 RVR-BL200K V1 AAA-ME5813A

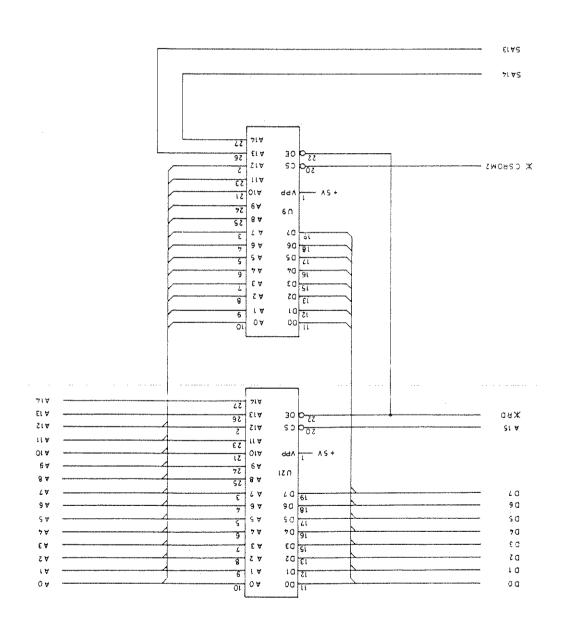
Parts No.	ADVANTEST Stock No.	Parts No.	ADVANTEST Stock No.	
CCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCCC	CCP-AC100P50V CCP-ACR01U50V LCL-A00670 SFN-2SK878 RCP-AJ100 RCP-AJ39 RCP-AJ18 RCP-AJ18 RCP-AJ18 RCP-AJ18 RCP-AJ18 RCP-AJ18 RCP-AJ100K RCP-AJ39 RCP-AJ39 RCP-AJ100K RCP-AJ39 RCP-AJ100K RCP-AJ39 RCP-AJ100K RCP-AJ100K RCP-AJ100K RCP-AJ39 RCP-AJ100K RCP-AJ10K RCP-AJ100K RCP-AJ10K	J2 -5 J6 J8 NF1 R1 R2 V1	JCF-AB001JX03 JCS-AV004JX01 JCD-AV003PX01 DEE-001427 JTE-AG001EX01 RVR-BA10K RVR-BL200K AAA-ME5813A	

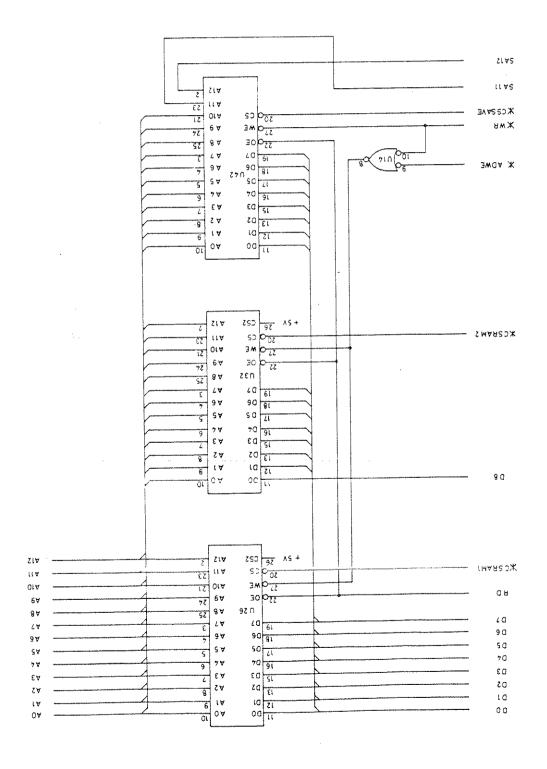


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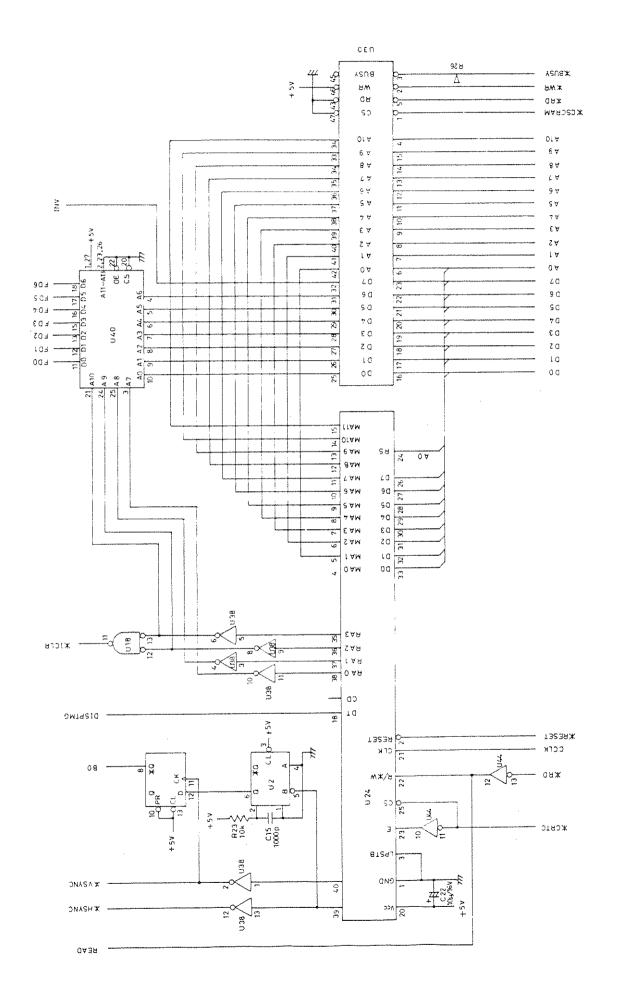


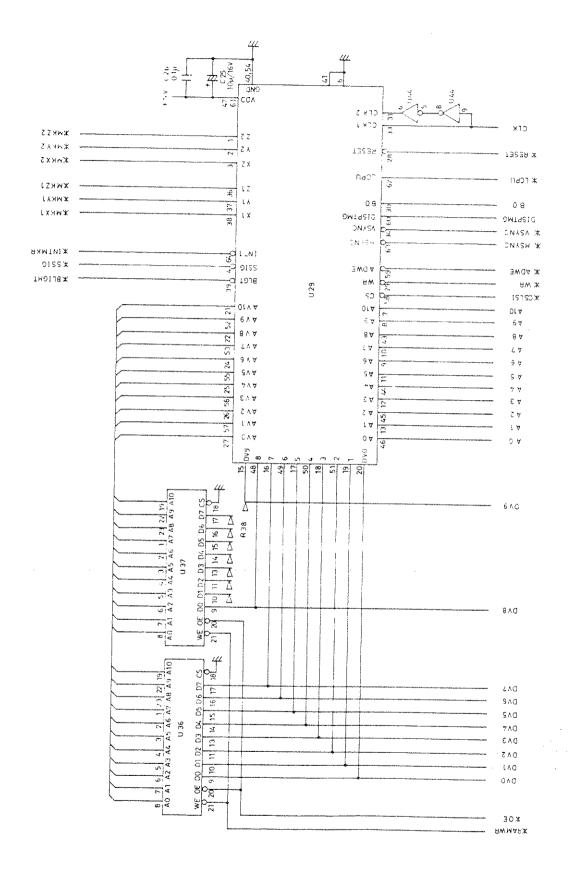


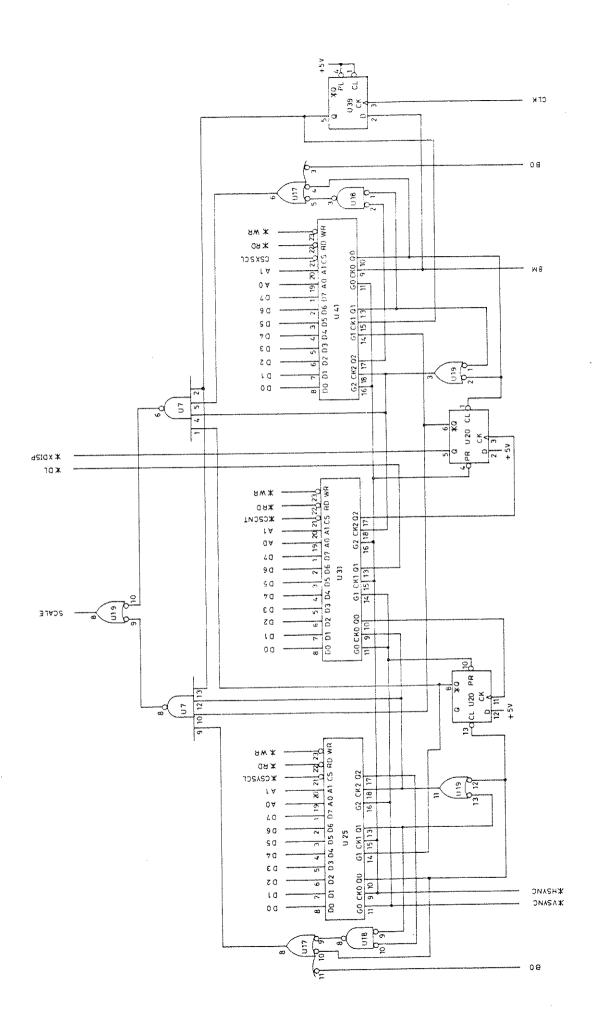
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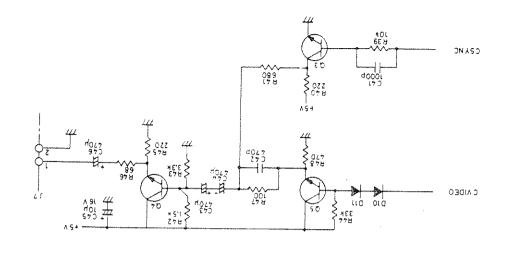
R4131 SENIES LOGIC BLR--015114 6/14

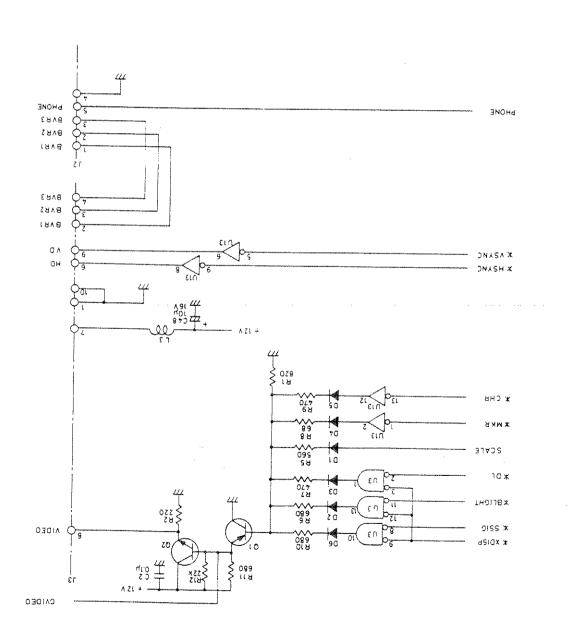


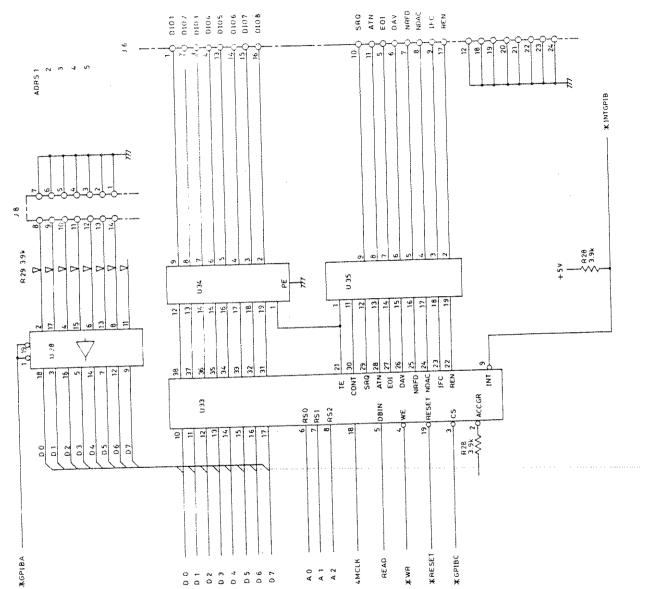


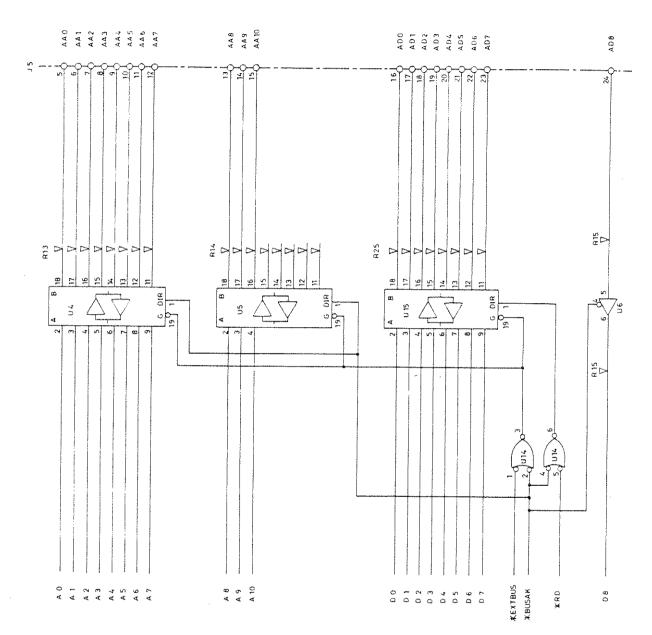


R4131 SERIES LOGIC BLR-015114 9/14

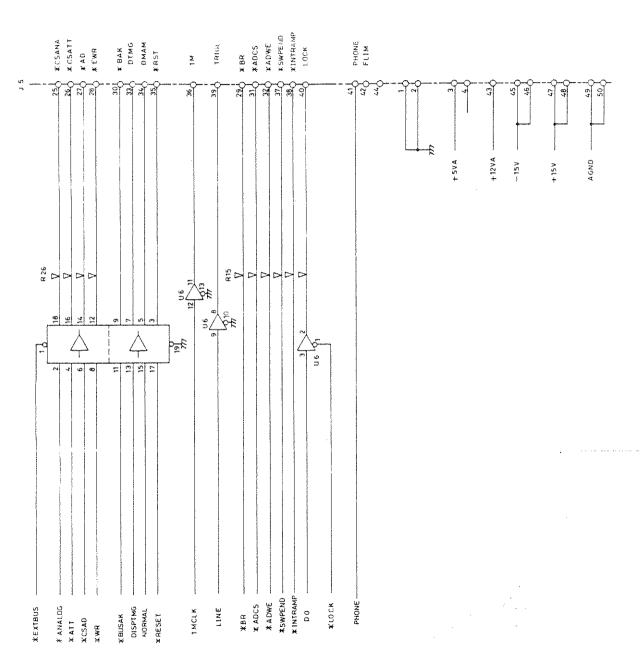


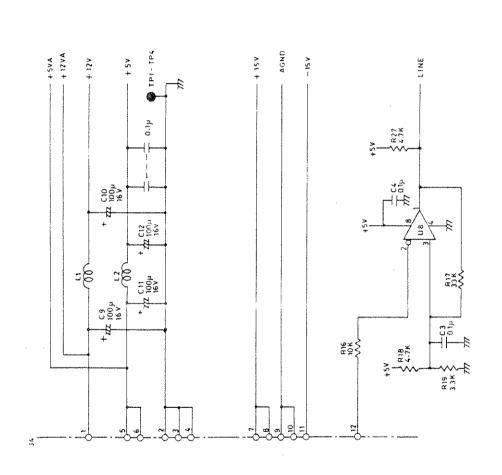




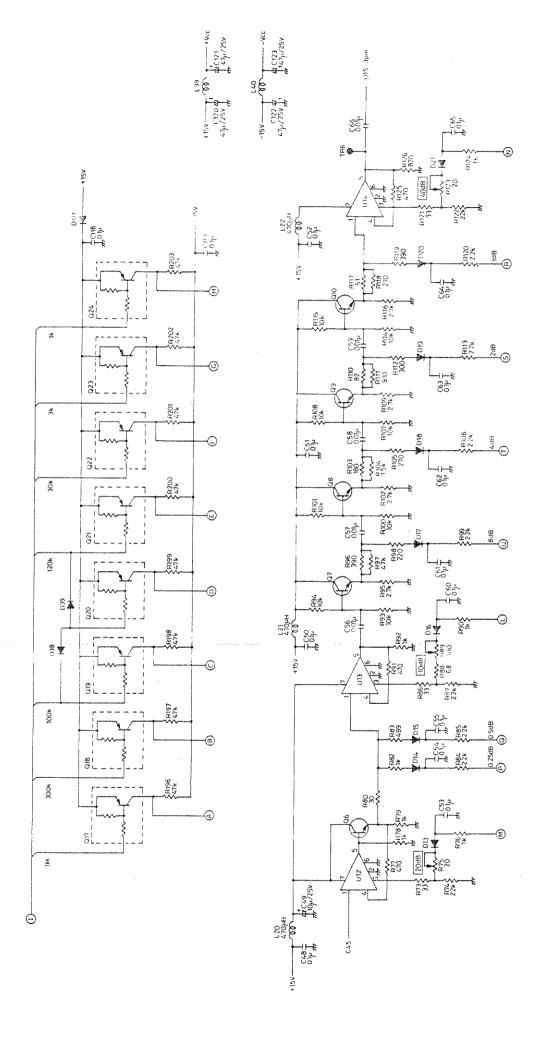


R4131 SERIES LOGIC BLR-015114 13/14 A - 51





R4131 SERIES YTO CNT/IF BLR-015116 1/6

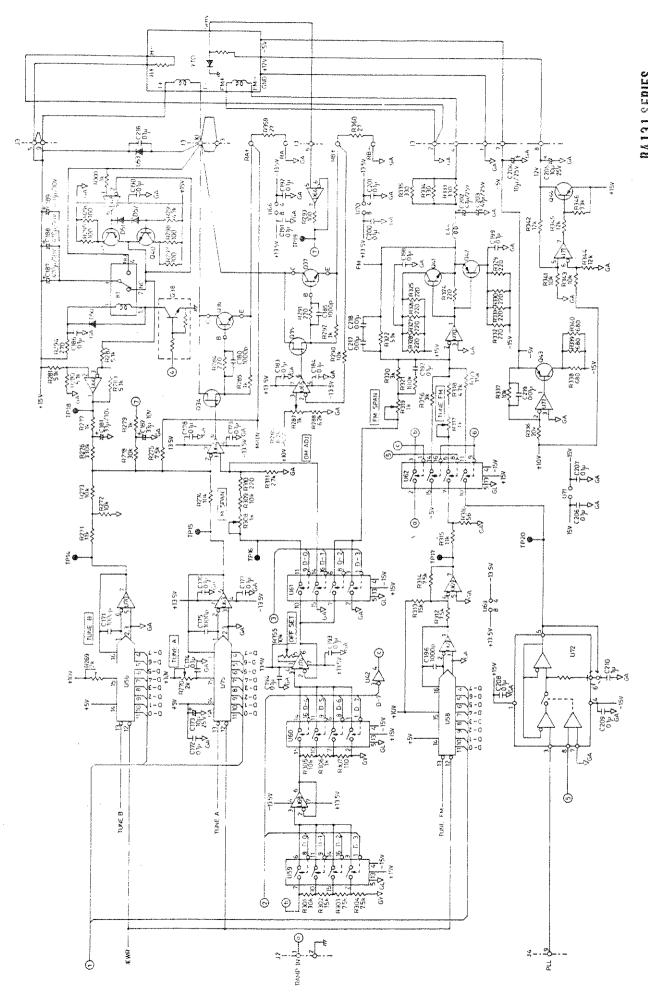


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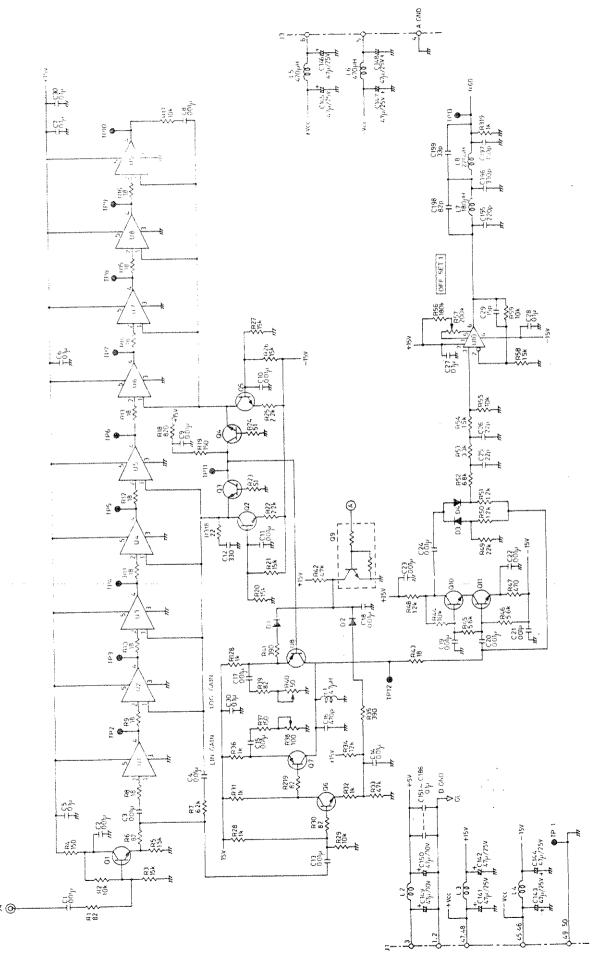
R4131 SERIES YTO CNT/IF BLR-015116 3/6

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R4131 SERIES YTO CNT/IF BLR-015116 5/6



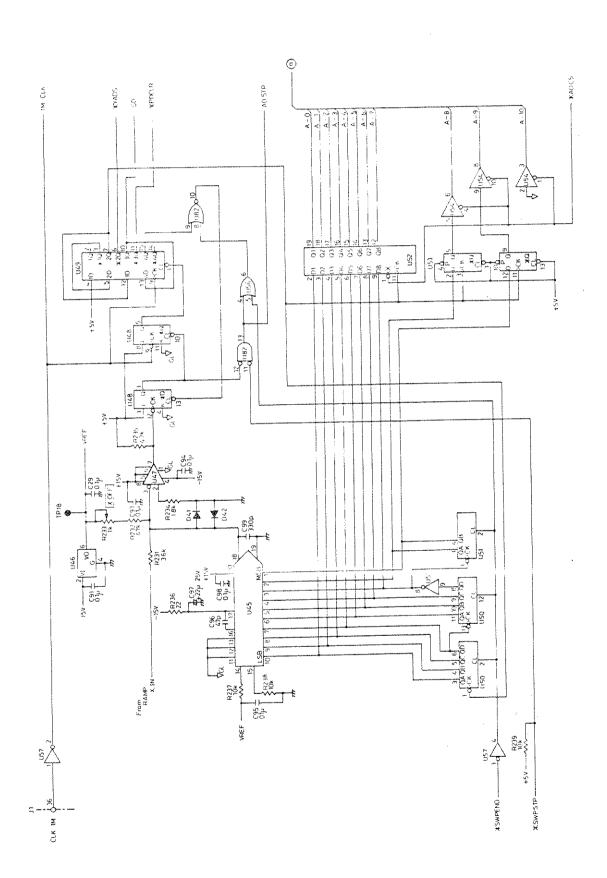
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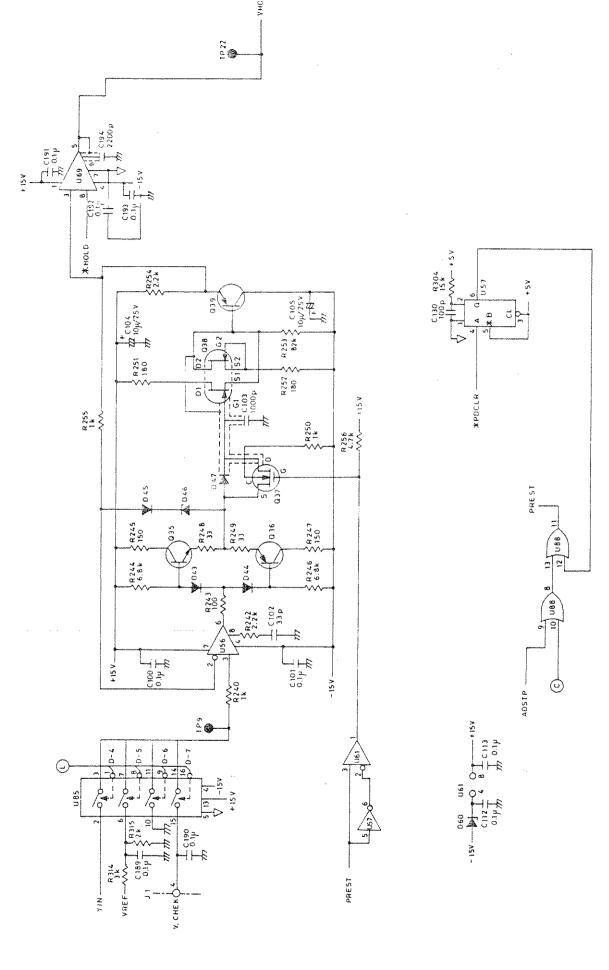


R4131 SENIES ANALOG(LO9) BLR-015117 2/8 A - 60

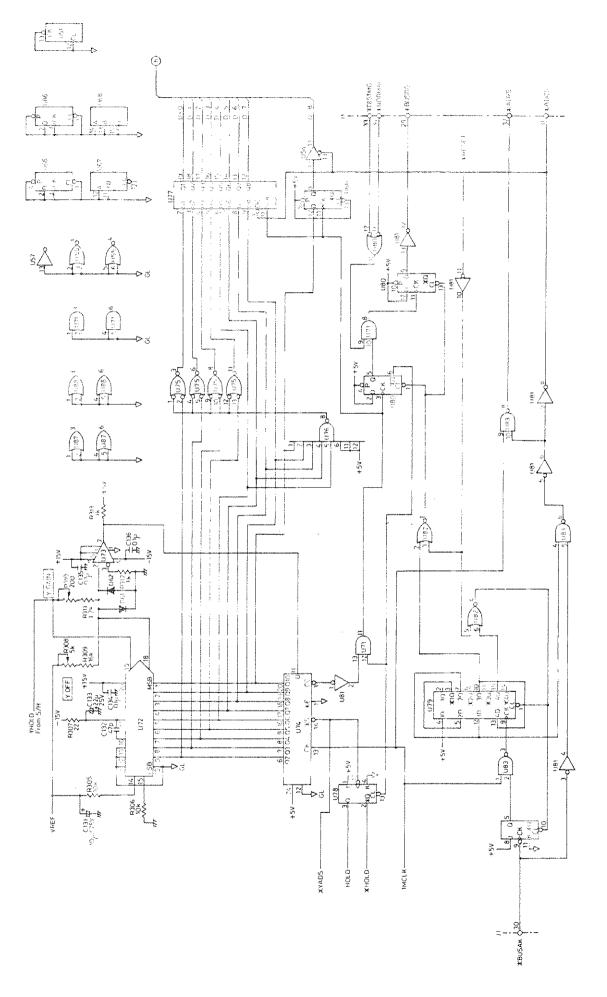
R4131 SERIES ANALOG(Ramp) BLR-015117 3/8

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BLR-015117 4/8



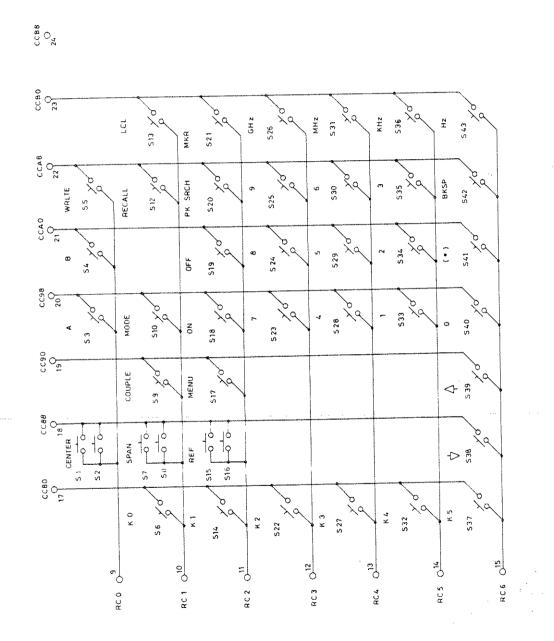


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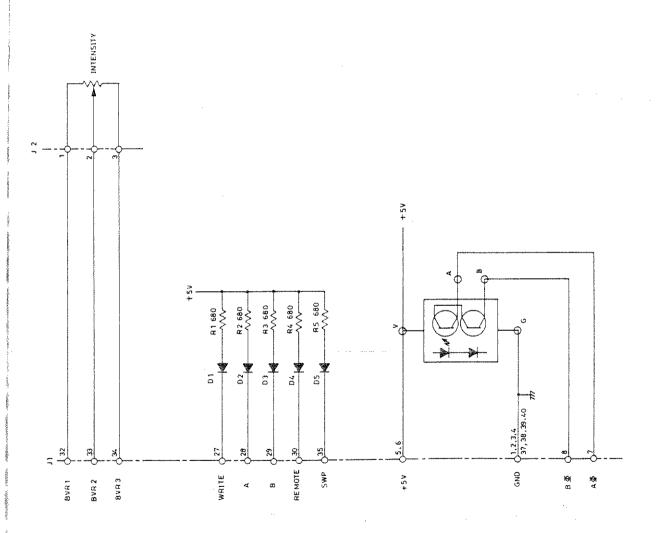
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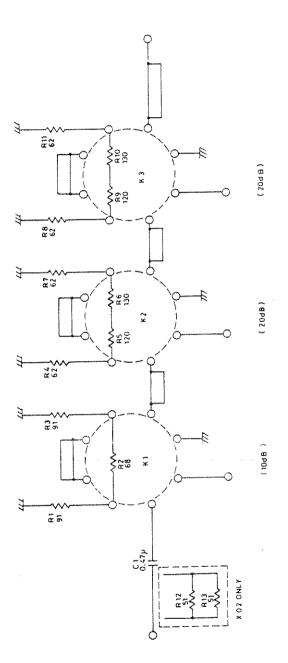
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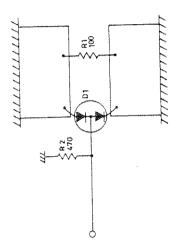


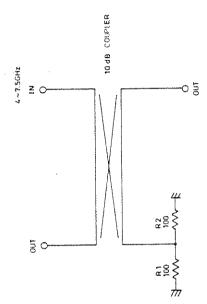
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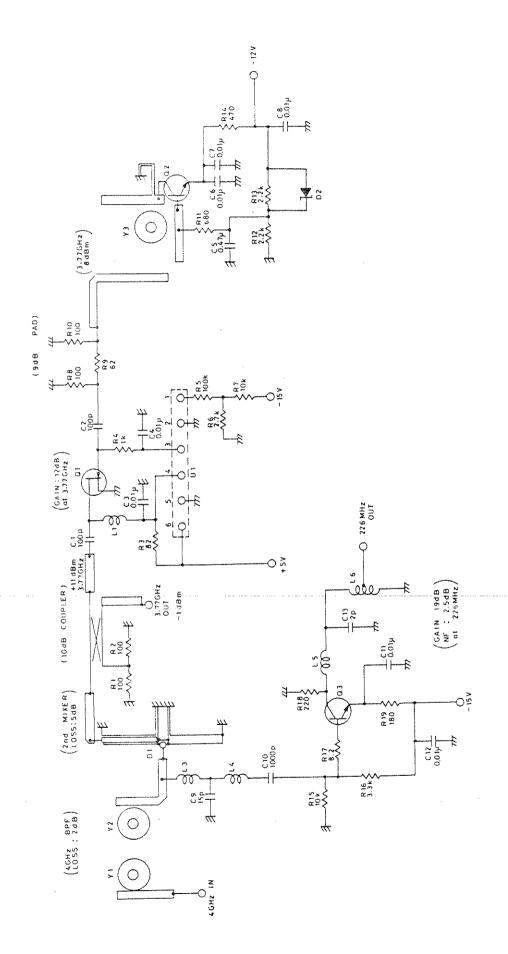
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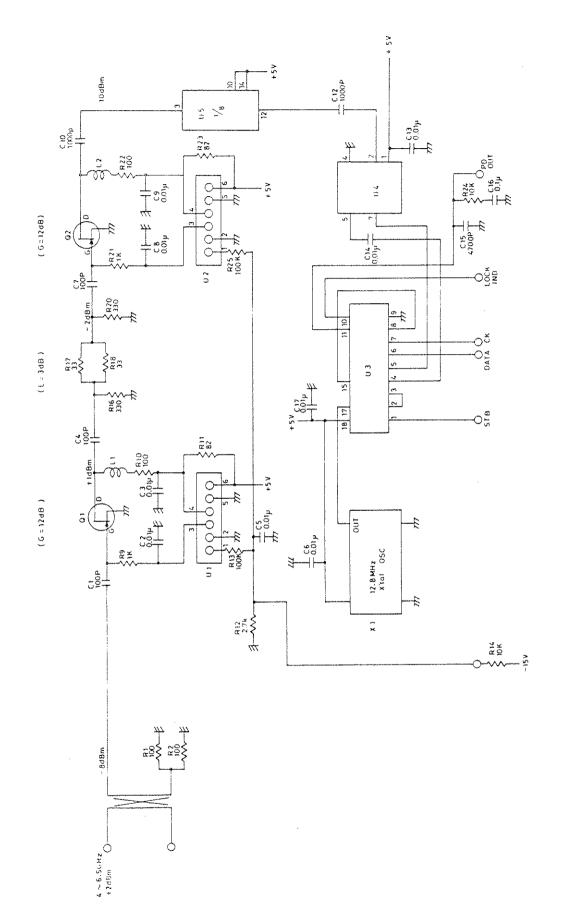




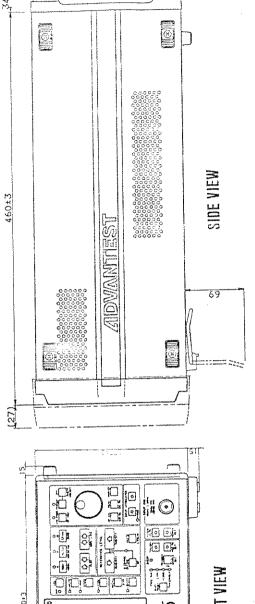








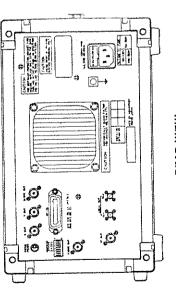
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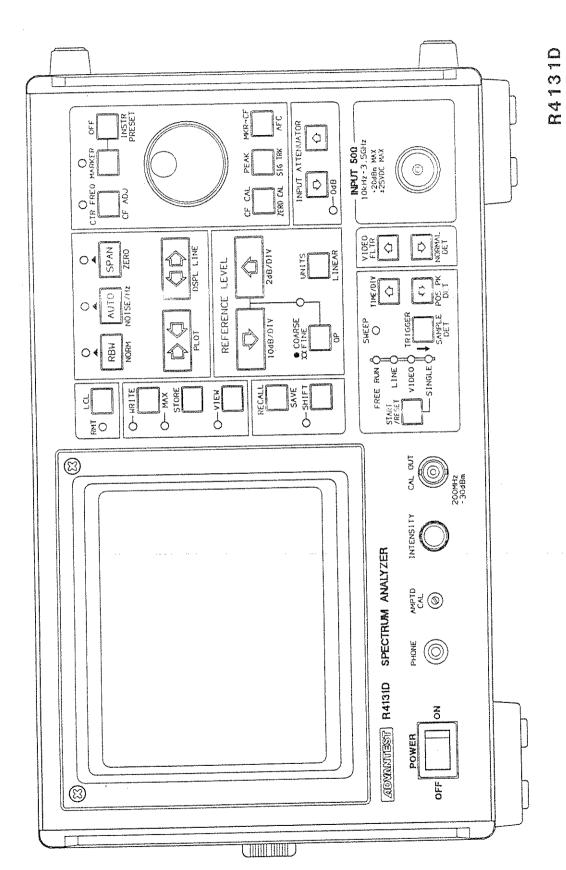
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REAR VIEW

MEN VIEW

R4131EXT2-9402-B



R4131EXT4-9402-B

R4131EXT6-9402-C

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ADVANTEST product is warranted against defects in material and workmanship for a period of one year from the date of delivery to original buyer.

LIMITATION OF WARRANTY

The foregoing warranty shall not apply to defects resulting from improper or inadequate maintenance by buyer, unauthorized modification or misuse, accident or abnormal conditions of operations.

No other warranty is expressed or implied. ADVANTEST specifically disclaims the implied warranties of merchantability and fitness for a particular purpose.

ADVANTEST shall not be liable for any special incidental or consequential damages, whether in contract, tort or otherwise.

Any and all warranties are revoked if the product is removed from the country in which it was originally purchased.

SERVICE

During the warranty period, ADVANTEST will, at its option, either repair or replace products which prove to be defective.

When trouble occurs, buyer should contact his local supplier or ADVANTEST giving full details of the problem and the model name and serial number.

For the products returned to ADVANTEST for warranty service, buyer shall prepay shipping and transportation charges to ADVANTEST and ADVANTEST shall pay shipping and transportation charges to return the product to buyer. However, buyer shall pay all charges, duties, and taxes incurred in his country for products returned from ADVANTEST.

CLAIM FOR DAMAGE IN SHIPMENT TO ORIGINAL BUYER

The product should be throughly inspected immediately upon original derivery to buyer. All material in the container should be checked against the enclosed packing list or the instruction manual alternatively. ADVANTEST will not be responsible for shortage unless notified immediately.

If the product is damaged in any way, a claim should be filed by the buyer with carrier immediately. (To obtain a quotation to repair shipment damage, contact ADVANTEST or the local supplier.) Final claim and negotiations with the carrier must be completed by buyer.

SALES & SUPPORT OFFICES

Advantest America, Inc. 300 Knightsbridge Parkway Suite 200, Lincolnshire, IL 60069 U.S.A.

> Phone (708) 634-2552 Facsimile (708) 634-2872

Advantest UK Limited

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CI. Tower, St. Georges Square, High Street,

New Malden Surrey, KT3 4HH The United Kingdom

Phone (081) 336-1606 Facsimile (081) 336-1657 Advantest (Singapore) Pte. Ltd.
150 Beach Road, #01-00
Gateway West Singapore 0718
Phone (65) 299-4268
Facsimile (65) 299-4226

Advantest Giga S.A.

5, Av du Quebec Z.A. de Courtaboeuf B.P.203-91941 Les Ulis Cedex, France Phone (331)6918-2500 Facsimile (331)6982-9224



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