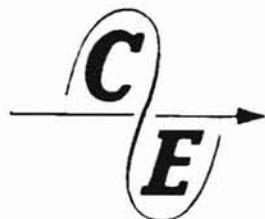


MODEL 301
OSCILLOSCOPE

P Y 4 A J
FABIO TEIXEIRA MAGALHÃES



CUSHMAN
ELECTRONICS, INC.

manufacturers of precision electronic instruments

ARQUIVO TÉCNICO

166 San Lazaro Avenue
Sunnyvale, Calif. 94086

WARRANTY

CUSHMAN ELECTRONICS, INC. WARRANTS EACH OF THE INSTRUMENTS OF THEIR MANUFACTURE TO BE FREE FROM DEFECTS IN MATERIAL AND WORKMANSHIP FOR A PERIOD OF ONE YEAR FROM THE DATE OF ORIGINAL PURCHASE. THE FOREGOING IS IN LIEU OF ANY OTHER WARRANTY, EXPRESS, IMPLIED OR STATUTORY. THE COMPANY, UNDER THEIR LIBERAL WARRANTY, WILL REPAIR OR REPLACE ANY INSTRUMENT FOUND DEFECTIVE.

THIS WARRANTY MAY NOT APPLY TO INSTRUMENTS WHICH, IN THE OPINION OF THE COMPANY, HAVE BEEN ALTERED OR MISUSED.

This manual is intended to give the user a comprehensive knowledge of the instrument and its operation or repair. In the event of trouble, study the manual carefully. Most instrument malfunctions can be corrected by the user with a minimum of lost usage time.

For assistance or information of any kind, contact the factory. Give full details of the nature of your problem and include the model and serial number of the instrument.

Should it appear that the instrument needs to be returned to the factory for service or recalibration, let us know. Shipping instructions will be promptly given to you. There will be no charge for repair on instruments within the one year warranty other than transportation costs after 90 days of ownership. Estimates of charges for non-warranty or any other service will be supplied by the factory upon request before work is begun and such work will be done on an actual cost basis only.

CLAIM FOR DAMAGE IN SHIPMENT

Your instrument should be inspected and tested as soon as it is received. The instrument is insured for safe delivery. If the instrument is damaged in any way or fails to operate properly, file a claim with the carrier or, if insured separately, with the insurance company.

* * * *

*We sincerely pledge our immediate and fullest cooperation
to all users of our precision electronic instruments.*

PLEASE ADVISE US IF WE CAN ASSIST YOU IN ANY MANNER

CUSHMAN ELECTRONICS, INC.
166 San Lazaro Avenue
Dial Area Code 408 - REgent 9-6760
Sunnyvale, California

P Y 4 A J
FABIO TEIXEIRA MAGALHÃES

CONTENTS

	<i>Page</i>
SECTION 1. GENERAL DESCRIPTION	1-1
SECTION 2. OPERATING INSTRUCTIONS	
2.1 General	2-1
2.2 Deviation Measurements	2-1
2.3 Low-Frequency Phase Comparison	2-1
SECTION 3. CIRCUIT DESCRIPTION	
3.1 General	3-1
3.2 Vertical Amplifier Board	3-1
3.3 Horizontal Amplifier Board	3-1
3.4 High-Voltage Power Supply Board	3-3
3.5 Low-Voltage Power Supply Board	3-3
SECTION 4. MAINTENANCE	
4.1 Calibration Adjustments	4-1
4.1.1 Vertical Adjustments	4-1
4.1.2 Vertical Gain	4-1
4.1.3 CRT Balance	4-2
4.1.4 Horizontal Gain	4-2
4.2 Typical Voltages and Waveforms	4-2
SECTION 5. SCHEMATICS AND PARTS' LOCATIONS	
5.1 Reference Designators	5-1
5.2 Abbreviations	5-1
5.3 Schematics, Parts Lists, and Pictorials	5-2
ILLUSTRATIONS	
2-1 Oscilloscope Front Panel	2-1
3-1 Overall Block Diagram	3-2
3-2 Horizontal Amplifier Board, Block Diagram	3-3
4-1 Model 301 Oscilloscope, Left-Side View of Interior	4-1
4-2 Model 301 Oscilloscope, Right-Side View of Interior	4-2
5-1 Model 301 (FP)	5-2
5-2 High Voltage Power Supply	5-5
5-3 Low Voltage Power Supply	5-7
5-4 Vertical Amplifier (A3)	5-9
5-5 Horizontal Amplifier and Sweep (A4)	5-12
TABLES	
1-1 Model 301 Specifications	1-1
4-1 High-Voltage Supply Board Voltage Chart	4-3
4-2 Low-Voltage Power Supply Board Voltage Chart	4-3
4-3 CRT Voltage Chart	4-3
4-4 Vertical Amplifier Board, Voltage Chart	4-3
4-5 Horizontal Amplifier Board, Voltage Chart	4-4
4-6 Horizontal Amplifier Board Waveforms	4-4

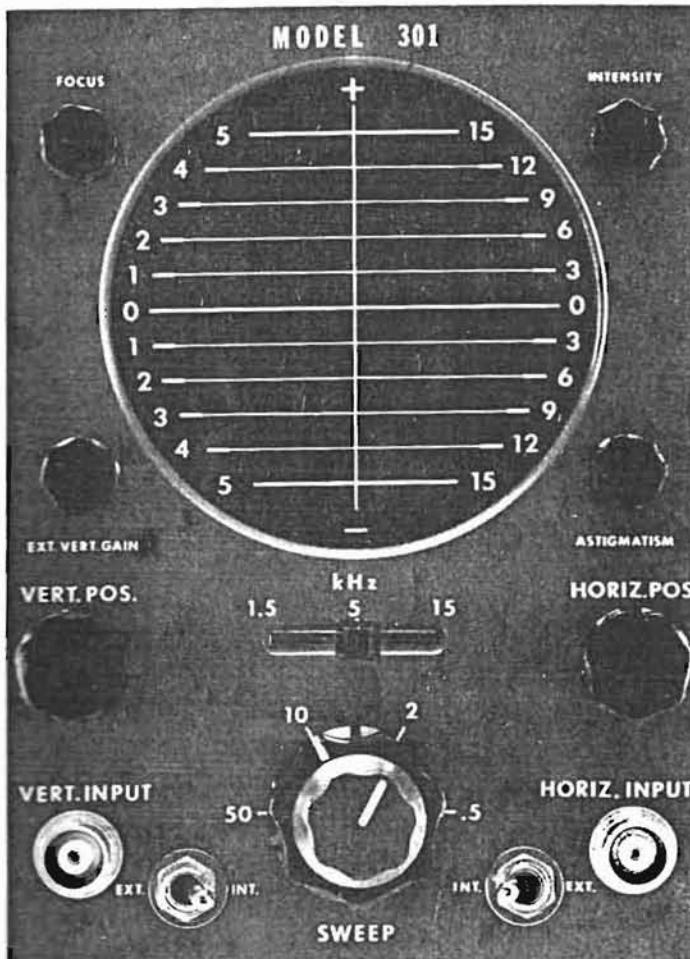
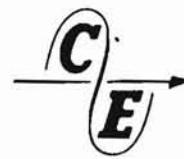


Figure 2-1 Oscilloscope Front Panel

SECTION 1 GENERAL DESCRIPTION

The Model 301 Oscilloscope is designed for use with the Cushman Models CE-3 and CE-7 FM Communications Monitors. The Oscilloscope enables the user to visually examine the exact modulation delivered by the transmitter. Any abnormal conditions such as unbalanced modulation, pulses from the power supply, unusual clipping, and other problems are immediately apparent so that appropriate repairs can be made. The Model 301 has an automatic sync circuit to lock on and precisely display waveforms. Vertical display is calibrated to show deviation on any one of three ranges: ± 1.5 , ± 5 , and ± 15 kHz.

Front panel switches permit an external vertical signal while using the internal sweep. External signals may also be connected to the horizontal input. Thus, the oscilloscope can be used independently of the monitoring function for routine testing, trouble shooting, displaying Lissajous patterns, etc.

Deviation Accuracy	$\pm 5\%$ full scale in three ranges: ± 1.5 kHz, 5.0 kHz, 15.0 kHz
External Inputs	Vertical and Horizontal
External Vertical Sensitivity	300 millivolts for full-scale
External Frequency Response (3 db)	30 kHz

Table 1-1. Model 301 Specifications

SECTION 2

OPERATING INSTRUCTIONS

2.1 General

The Model 301 Oscilloscope Plug-In has the standard oscilloscope front-panel controls, most of which are self-explanatory. (See Figure 2-1.) A focus control, intensity control, and astigmatism control are provided to ensure a clean trace on the face of the CRT. Vertical and horizontal position controls are provided to position the display and set the display with respect to the zero reference line on the scope graticule when the CE-3 function switch is in the CAL position.

The lever switch directly below the CRT is a range switch for vertical display of deviation and is directly calibrated in ± 1.5 , ± 5 , and ± 15 kHz of deviation. The scope circuit is dc-coupled and the displayed deviation will shift up or down if the transmitter under test is off frequency.

Separate external inputs to the vertical or horizontal amplifiers which drive the CRT are provided. Separate toggle switches permit selection of either internal or external signals in conjunction with the external input connectors.

The Model 301 sweep speeds can be adjusted with a sweep-speed switch and vernier control; it is normally unnecessary to adjust the sweep speed to lock the audio waveform being displayed. The Model 301 incorporates an automatic sync control which provides instant lock on any signal exceeding one division in peak-to-peak amplitude. An external-vertical-gain control permits attenuation of signals applied to the external vertical input.

The Model 301 Oscilloscope is removed by turning the large slotted knob on the rear panel of the monitor counterclockwise. When the locking shaft has been disengaged from the plug-in, pushing the knob inward will cause the plug-in connector to separate. The Model 301 can then be easily removed.

2.2 Deviation Measurements

Operation of the Model 301 Oscilloscope with the Model CE-3 or Model CE-7 to measure frequency deviation is described in the Monitor Manual.

2.3 Low-Frequency Phase Comparison

The Model 301 can be used to compare the phase of tone-operated squelch devices (such as the Vibra-sender) by using the Oscilloscope in conjunction with the Model 308 Decade Frequency Divider, as follows:

- a. generate an IF/Audio signal in the CE-3 or CE-7 that has a frequency of 10 times, 100 times, or 1000 times that of the signal with which it is to be compared.
- b. Set up the correct dividing ratio on the Model 308.
- c. Feed the Monitor output through the Model 308 to the HORIZ. INPUT connector of the Model 301.
- d. Feed the signal to be compared to the VERT. INPUT connector of the Model 301
- e. Adjust the Monitor output frequency for the slowest moving Lissajous pattern on the Model 301.

□

CAUTION

■ Prolonged display of a bright spot or trace may burn the face of the CRT. Avoid this condition if possible. Use the CRT at half brightness to prolong the life of the CRT.

■ High voltages are used in the Model 301 Oscilloscope and the plug-in must not be removed or replaced with the power switch on. It is recommended that the power switch be turned off for 30 seconds before the Model 301 Oscilloscope Plug-In is removed to allow discharge of all high-voltage capacitors.

SECTION 3

CIRCUIT DESCRIPTION

3.1 General

The circuits of the Model 301 Oscilloscope are on four boards—the Vertical Amplifier board, the Horizontal Amplifier board, the Low-Voltage Power Supply board, and the High-Voltage Power Supply board. The 115-vac supply is fused (2/10 amp S. B.) at the plug-in receptacle J3, in the monitor.

Figure 3-1 is the overall block diagram. The power supply circuits and the usual cathode-ray tube controls (ASTIGMATISM, FOCUS, and INTENSITY) are not shown in the interest of simplicity. The CRT sweep can be either external or internal. The external sweep is applied through the front-panel HORIZ. INPUT BNC connector when the associated switch is in the EXT. position. When the switch is in the INT. position, the sweep comes from the internal sweep generator. The internal-sweep period is selectable by a four-position switch and a vernier control from less than 0.5 milliseconds to more than 50 milliseconds, and is synchronized with the vertical input. A blanking signal is fed through an amplifier on the High-Voltage Power Supply board to the cathode of the CRT to blank the return trace. The horizontal position of the CRT trace is adjustable with the HORIZ. POS. control, which determines the dc voltage on the horizontal deflection plates.

The input to the CRT vertical deflection plates can be an external signal (when the VERT. INPUT switch is in the EXT. position) or it can come from the 2nd I.F. and Discriminator board of the CE-3 or CE-7 (when the VERT. INPUT switch is in the INT. position). In the INT. position, the oscilloscope displays the modulation deviation of the signal being monitored by the CE-3 or CE-7. In addition, after the oscilloscope trace has been centered vertically by the VERT. POS. control while the Monitor is in the CAL. mode, the Monitor can be switched to the FREQ.-DEV. MEASURE mode and the monitored transmitter's center-frequency error is represented by the displacement of the trace from the CRT center line. Thus, the Model 301 Oscilloscope can be used to indicate frequency error as well as FM deviation.

The vertical signal input is dc-amplified and applied to one of the vertical deflection plates through reversing relay K1. Input to the other vertical deflection plate is an amplifier dc voltage from the VERT. POS. potentiometer. For most settings of the CE-3 frequency selector switches, the LO frequency is 10 MHz lower than the dialed-in frequency. When this is the case, relay K1 is deenergized and its contacts feed the vertical input signal to the vertical deflection plates in such a way as to cause an upward deflection of the CRT trace for an increase in the frequency of the monitored signal. However, when the 1st-digit frequency selector switch is in position 0 or 1 while the 2nd-digit switch is in position 2 or 3, the CE-3 LO frequency is 10 MHz higher than the dialed-in frequency. This would cause the beam to deflect downward for an increase in signal frequency if it were not for the reversing action of relay

K1, which is energized in these particular combinations of switch positions. The gain of the vertical amplifiers is controlled by the deviation range switch, which connects different calibrating resistances for the 1.5-, 5-, and 15-kHz ranges into the circuit.

3.2 Vertical Amplifier Board

Figure 5-4 is the schematic diagram of the Vertical Amplifier board. Transistors Q1 through Q6 comprise a three-stage differential amplifier that drives the vertical deflection plates of the CRT. The base of Q1 is biased to about +3.2 volts, by the Monitor discriminator circuit when the VERT. INPUT switch is in the INT. position, or by voltage divider R3 and R4 when the switch is in the EXT. position. The bias on the base of Q2 is varied around +3.2 volts by the VERT. POS. potentiometer (R5) to move the CRT trace up or down. The vertical amplifier is designed to have the proper gain to produce full-scale vertical deflection of the CRT when the signal from the Monitor 2nd I.F. and Discriminator board corresponds to a deviation of ± 15 kHz. For the 1.5-kHz and 5-kHz deviation ranges, the negative feedback resistors in the emitter circuits of Q5 and Q6 are shunted by appropriate resistances (R30 + R32 for the 1.5-kHz range and R29 + R31 for the 5-kHz range) to increase the amplifier gain by the right amount to provide the correct vertical deflection for those ranges. Variable resistors R32 and R31 provide a means for calibrating the two ranges.

Transistor Q7 is the sync amplifier. Its input comes from one of the CRT vertical deflection plates, and its output is fed to the trigger circuit on the Horizontal Amplifier board.

3.3 Horizontal Amplifier Board

On this board are the internal sweep generator for the CRT and the amplifiers to drive the horizontal deflection plates of the CRT from either the internal sweep or an external horizontal input. Refer to the block diagram (Figure 3-2) and the schematic diagram (Figure 5-5).

Assume that the HORIZ. INPUT switch is in the INT. position, the SWEEP switch is in the 10-msec position, and that there is no sync input. With no sync input, the output of Schmitt Trigger Q1/Q2 is essentially ground potential; therefore, resistor R12 is shunted by approximately 10 kilohms (R13 plus R39). This lowered resistance demands more current than the constant-current source (Q6) can supply to keep regenerative switch Q3/Q4 on hence it turns off. When the regenerative switch turns off, timing capacitor C6 starts charging linearly from the constant-current source. When C6 charges to the threshold level of the regenerative switch, the switch closes and discharges C6. After C6 discharges, the shunt across R12 again causes the regenerative switch to turn off, and C6 recharges. Thus, a sawtooth waveform is generated at the top plate (in the block diagram) of C6. This sweep

waveform is fed through emitter follower Q7 and one half of a differential amplifier (Q9/Q11) to one of the horizontal deflection plates. The other horizontal deflection plate is fed by the other half of the differential amplifier (Q10/Q12), the input to which is a dc voltage adjustable by HORIZ. POS. potentiometer R7.

The bottom plate of timing capacitor C6 is not returned to ground as is conventional, but instead, is fed a portion of the potential from the top plate of the capacitor through emitter followers Q7 and Q8. This changes the effective capacitance of C6 and therefore the timing period. The amount by which the effective capacitance is changed is controlled by the SWEEP vernier potentiometer, R6. Thus, the sweep period is determined not only by the timing capacitor selected by the SWEEP switch, but also by the setting of the SWEEP vernier control (the red knob concentric with the black SWEEP switch).

Now assume that a sync signal is being received from the vertical-deflection input of the CRT. This sync

signal is converted to a rectangular waveshape by Schmitt Trigger Q1/Q2, and is further converted to positive trigger pulses by the circuit consisting of C2, R9, and CR1. The output of the Schmitt Trigger also charges C9 to a positive voltage; this voltage back-biases CR5, thus removing shunt resistors R13 and R39 from the regenerative switch circuit. The regenerative switch can now be turned off only by a positive pulse from the trigger circuit. The regenerative switch stays off until timing capacitor C6 charges to the threshold level of the switch. When this happens, C6 discharges, and cannot start charging again to produce a new sweep until a trigger pulse is received from the sync circuit. Thus, the sweep is synchronized with the vertical input.

When the HORIZ. INPUT switch is in the EXT. position, the junction of R12 and CR5 in the regenerative switch circuit is grounded to stop generation of the internal sawtooth. The gain of the differential amplifier is adjustable by potentiometer, R38, which varies the amount of negative feedback in the output stage.

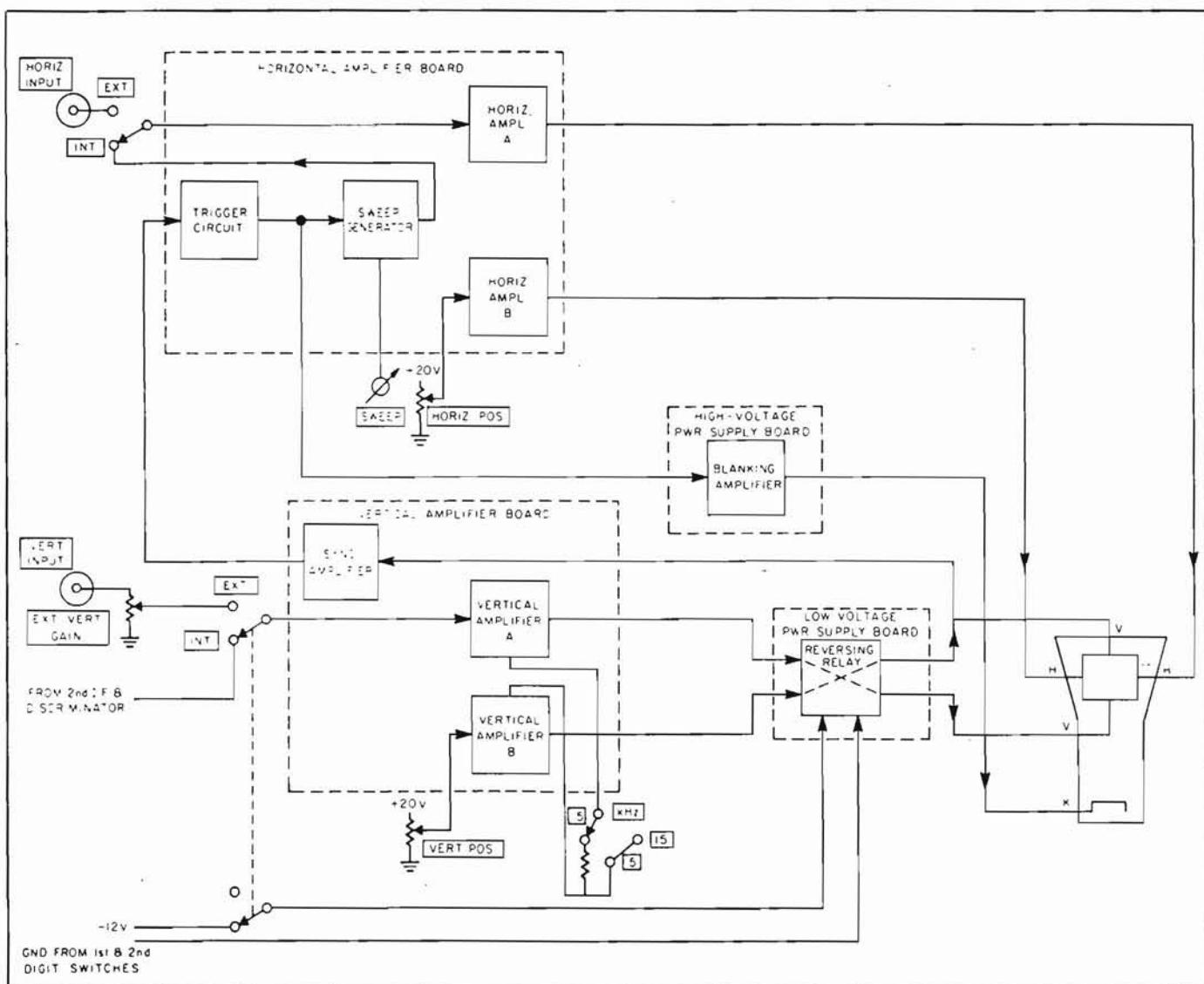


Figure 3-1 Overall Block Diagram

3.4 High -Voltage Power Supply Board

This board furnishes -1200 volts for the CRT. As shown in the schematic diagram, Figure 5-2, the 920 volts across the high-voltage secondary of transformer T1 is applied to the voltage-doubler rectifier consisting of C1, CR1, CR2, and C2. The approximately -2100 volts across C2 is reduced by the regulator circuit consisting of R1, R2, R3, and V1 to a regulated -1200 volts. The -1200 volt output is applied through R9, 100k ohm, to the grid of the CRT, and also across the voltage-divider chain consisting of INTENSITY control CH-R2, R11, FOCUS control CH-R1, R12, and R13. The potential on the cathode of the CRT is set by the INTENSITY control, which determines the negative bias on the grid (pin 2) with respect to the cathode; in addition, during retrace a negative pulse from the horizontal amplifier is inverted by Q1 and applied through capacitor C4 to the cathode of the CRT, cutting off the beam.

3.5 Low -Voltage Power Supply Board

This board (Figure 5-3) furnishes 240 volts to the horizontal amplifier on the Horizontal Amplifier board, and to the ASTIGMATISM potentiometer and blanking amplifier on the High-Voltage Power Supply board. An ac voltage of 245 volts rms from one of the secondaries of T1 on the High-Voltage Power Supply board is applied to bridge rectifier CR1 through CR4. The positive 320-volt output of the bridge rectifier is fed through the emitter-follower regulator, consisting of Q1, R2, CR5, and CR6, to the board connector.

Resistors R3 and R5, together with R4, are the collector load resistors of vertical amplifiers Q5 and Q6 on the Vertical Amplifier board. Potentiometer R4 permits changing the values of the load resistances slightly to compensate for any difference in gain of the two transistors and to adjust the electrical displacement of the beam to mid-scale. The function of relay K1 is explained in Section 3.1.

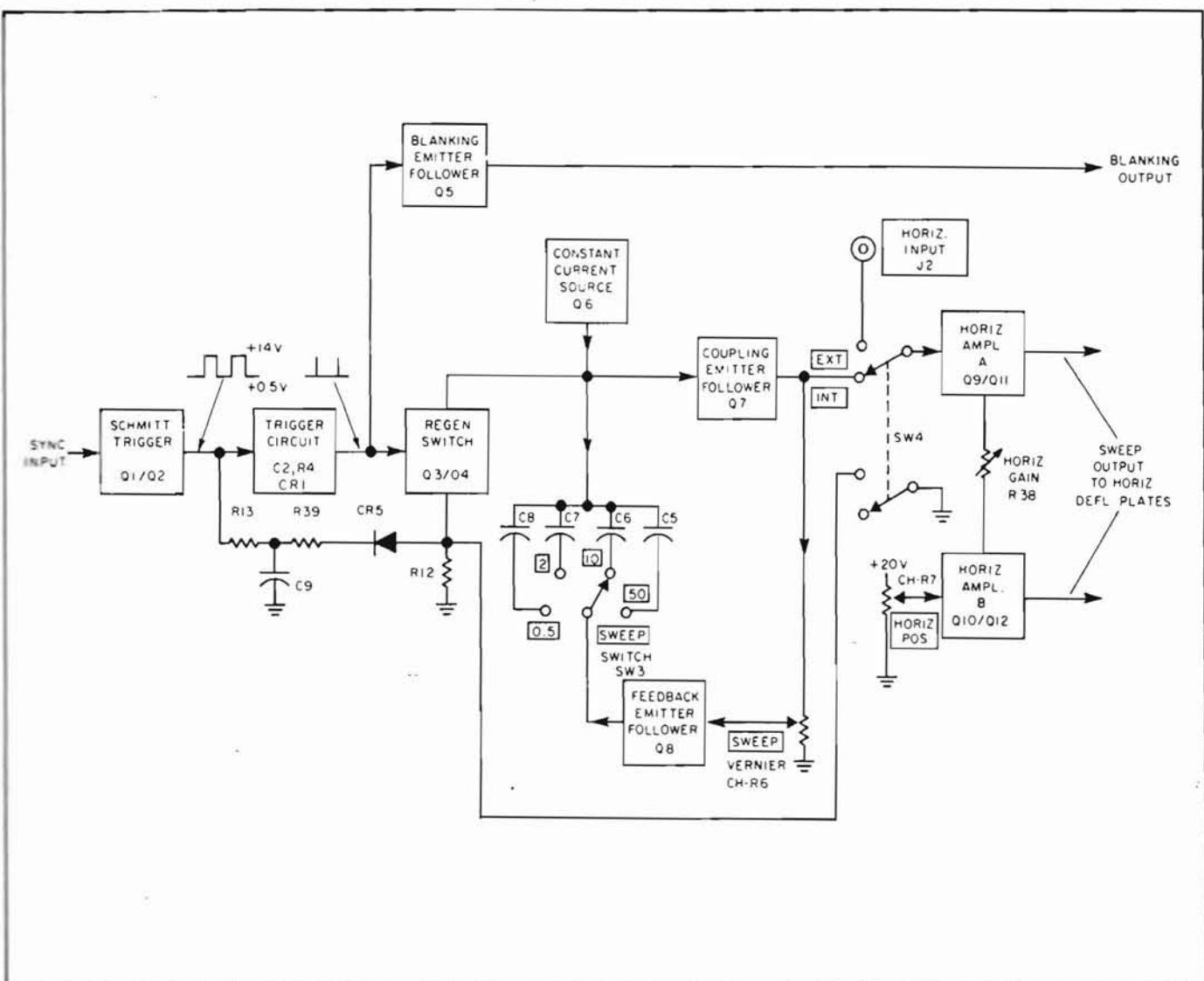


Figure 3-2 Horizontal Amplifier Board, Block Diagram

SECTION 4

MAINTENANCE

4.1 Calibration Adjustment

There are six internal adjustments in the Model 301 Oscilloscope. Controls for these adjustments are located as shown in Figures 4-1 and 4-2. The adjustments (except the horizontal gain adjustment) are made with the Model 301 plugged into the CE-3 or CE-7 Monitor and operating. The controls shown in Figure 4-1 can be reached by removing the left-hand CE-3 or CE-7 cover, as described in the maintenance section of the Monitor Manual. The only test equipment required is an unmodulated transmitter operating on any frequency covered by the Monitor. Unless otherwise stated, the controls mentioned in the following procedures are on the Model 301 Oscilloscope.

4.1.1 Vertical Amplifier Balance. Before making any adjustments of the vertical deviation ranges, the vertical amplifier balance should be checked as follows:

- a. Place the Monitor function switch in the CAL. position.
- b. Set the front-panel VERT. POS. control to mid-position.
- c. Set the front-panel range switch to the ± 15 -kHz position and note the position of the trace on the CRT.
- d. Switch to the ± 1.5 -kHz range and adjust the position of the trace to the same point noted above, using the vertical amplifier balance control (see Figure 4-1).

4.1.2 Vertical Gain.

- a. "Dial-in" the frequency of the transmitter being used for the test.
- b. In the CAL. position of the Monitor function switch, adjust the CAL. control for a zero or mid-scale reading on the FREQUENCY meter.
- c. Set the front-panel deviation range switch to ± 1.5 kHz and adjust the VERT. POS. control so the trace is on the "O" reference line.
- d. Set the Monitor function switch to the FREQ.—DEV. MEAS. position. Key the transmitter and adjust it "right on frequency" on the ± 1.5 -kHz FREQUENCY error meter range.
- e. With the deviation range switch set to ± 15 kHz, offset the "dialed-in" frequency 15 kHz high and low from the FCC assigned frequency. The scope trace should shift up and down by 15 kHz. If adjustment is required, use the ± 15 -kHz potentiometer; then recheck the transmitter frequency.
- f. Repeat step e. on the ± 5 -kHz deviation range by offsetting the "dialed-in" frequency by 5 kHz. Adjust the ± 5 -kHz potentiometer if necessary.
- g. Repeat step e. on the ± 1.5 -kHz deviation range by offsetting the "dialed-in" frequency by 1.0 kHz. Adjust the ± 1.5 -kHz potentiometer if necessary.
- h. Recheck the vertical amplifier balance adjustment by repeating steps a. through d. of Section 4.1.1.

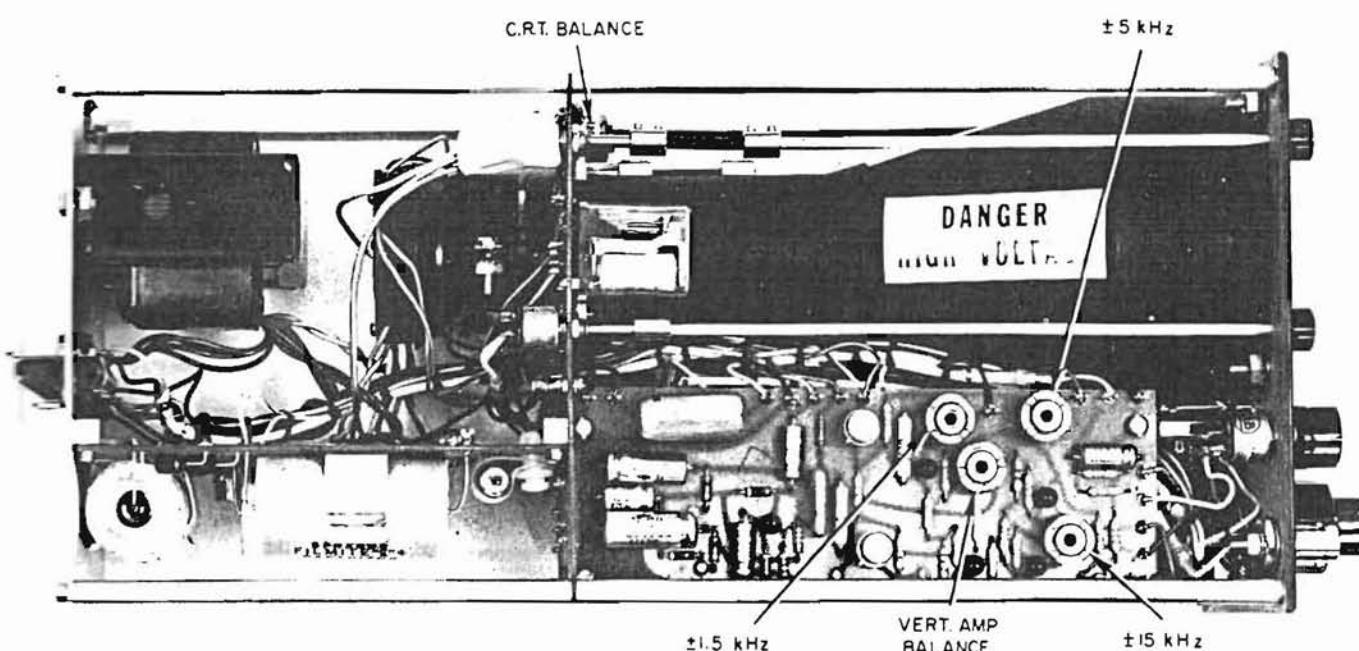


Figure 4-1. Model 301 Oscilloscope, Left-Side View of Interior

4.1.3 CRT Balance. If the trace is not on the zero reference line after adjusting the vertical amplifier balance control, adjust the CRT balance control until the trace is on the zero reference line. (See Figure 4-1 for location of the CRT balance control.) This control centers the range of the front-panel VERT. POS. control.

4.1.4 Horizontal Gain. An internal gain control on the horizontal amplifier permits adjustment of the sweep length. This control is located on the right side of the Oscilloscope plug-in and cannot be reached with the scope plugged into the Monitor. If the sweep length is short, remove the Oscilloscope from the Monitor and adjust the horizontal gain potentiometer (see Figure 4-2) in the counterclockwise direction.

CAUTION

Turn the ac power to the monitor OFF before removing the Oscilloscope plug-in and while replacing the plug-in.

The sweep length is not critical, but should be adjusted to provide a full sweep of the trace when it is on the zero reference line.



HORIZONTAL
GAIN CONTROL

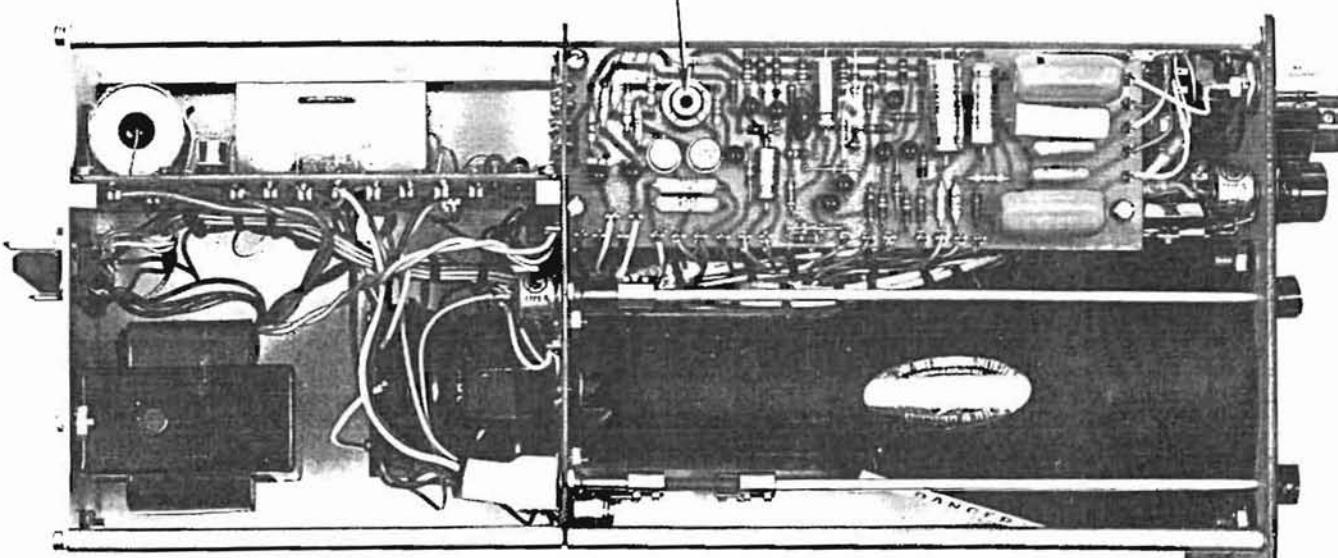


Figure 4-2. Model 301 Oscilloscope, Right-Side View of Interior

4.2 Typical Voltages and Waveforms

Tables 4-1 through 4-6 give typical data for the four Oscilloscope boards and the CRT. Test conditions are as follows unless otherwise stated:

Line Voltage	115 vac
Control Settings:	
FOCUS	Mid-range or optimum
INTENSITY	Mid-range
EXT. VERT. GAIN	Minimum
ASTIGMATISM	Mid-Range or optimum
VERT. POS.	Mid-range
HORIZ. POS.	Mid-range
Deviation Range	±15 kHz
SWEET Range	50 msec
SWEET Vernier	Fully ccw.

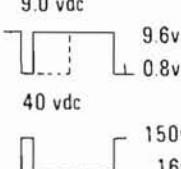
(No vertical input; INTENSITY control max clockwise)		
	HORIZ. INPUT Switch in INT. Position	HORIZ. INPUT Switch in EXT. Position
Transformer	920 vac	--
HV Secondary	-1125 vdc	--
CR1 Anode	-2180 vdc	--
C2	-1200 vdc	-1200 vdc
V1 Cathode	0.24 vdc	0.26 vdc
V1 Anode	9.0 vdc	10.0 vdc
Q1 Base (dc) (ac)	 9.6v 0.8v 40 vdc 150v 16v	--
Q1 Collector (dc) (ac)		9.5 vdc --

Table 4-1. High -Voltage Power Supply Board
Voltage Chart

Input to Bridge Rectifier	245 vac
C1	300 \pm 5 vdc
	0.7 vac
Q1 Base	240 \pm 10 vdc
Test Point TP1	240 \pm 10 vdc

Table 4-2. Low -Voltage Power Supply Board
Voltage Chart

Filament:	
Pin 12	-1200 vdc
Pin 1	-1200 vdc
Grid; Pin 2	-1200 vdc
Cathode; Pin 3	-1160 vdc
Focus; Pin 4	-840 vdc
Vertical Plates:	
Pin 6	+125 vdc
Pin 7	+125 vdc
Astigmatism; Pin 8	+125 vdc
Horizontal Plates:	
Pin 9	+125 vdc
Pin 10	+125 vdc

Table 4-3. CRT Voltage Chart

(With 0.32-vac, 1-kHz, 3.2-vdc input; 15-kHz range)

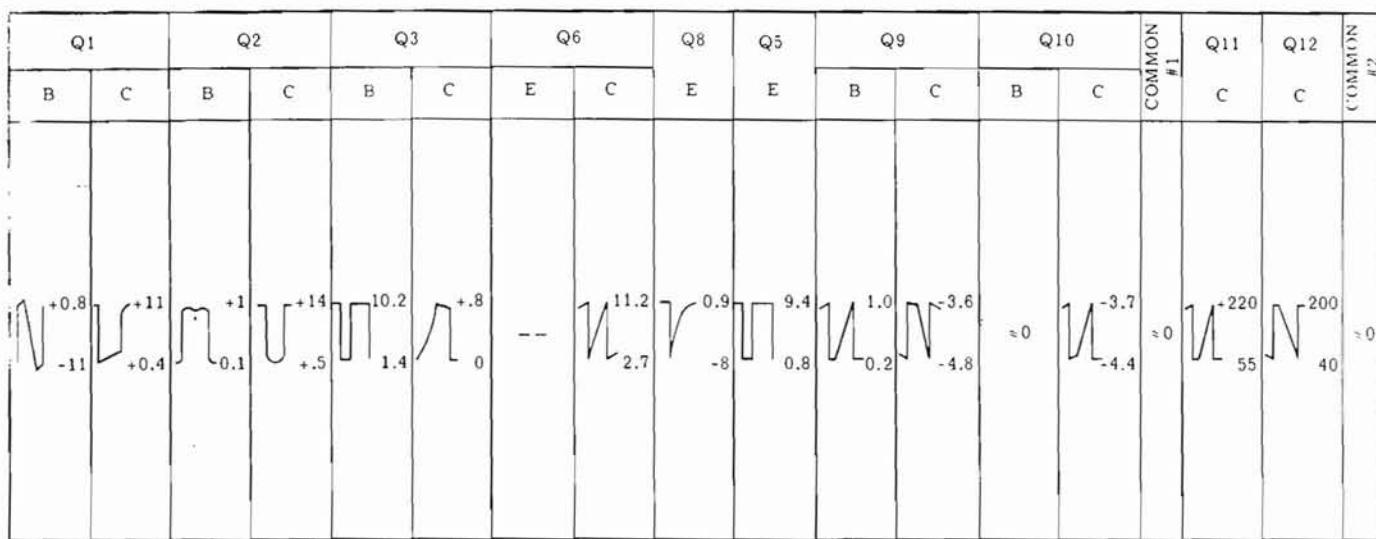
Q1 Base	3.2 vdc
Q1 Collector	0.90v p-p ac
Q2 Base	7.4 vdc
Q2 Collector	1.5v p-p ac
Common #1	3.2 vdc
Q3 Emitter	0v p-p ac
Q4 Emitter	0.87v p-p ac
Common #2	-3.6 vdc
Q5 Base	0v p-p ac (approx.)
Q5 Collector	8.4 vdc
Q6 Base	10.7 vdc
Q6 Collector	0v p-p ac (approx.)
Common #3	-1.2 vdc
Q7 Base	5.2v p-p ac
Q7 Collector	125 vdc
Q8 Base	80v p-p ac
Q8 Collector	-1.2 vdc
Common #4	4.2v p-p ac
Q9 Base	125 vdc
Q9 Collector	77v p-p ac
Common #5	-8.8 vdc
Q10 Base	0v p-p ac (approx.)
Q10 Collector	1.4 vdc
Q11 Base	1.0v p-p (clipped) ac
Q11 Collector	8.0 vdc
Q12 Base	12.0v p-p (clipped) ac

(Above readings depend on settings of all potentiometers)

Table 4-4. Vertical Amplifier Board, Voltage Chart

		No Vertical Signal; HORIZ. INPUT Switch in EXT. Position	Vertical Signal; HORIZ. INPUT Switch in INT. Position; SWEEP Switch at 10 msec; SWEEP Vernier Max. CCW
Q1 Base	0 vdc	-4.9 vdc	
Q1 Collector	11.7 vdc	6.8 vdc	
Q2 Base	1.0 vdc	0.60 vdc	
Q2 Collector	0.50 vdc	6.5 vdc	
Q3 Base	10.6 vdc	9.6 vdc	
Q3 Collector	0 vdc	0.16 vdc	
C9	0.40 vdc	6.0 vdc	
Q6 Emitter	14.2 vdc	14.2 vdc	
Q6 Collector	11.8 vdc	7.0 vdc	
Q8 Emitter	0.95 vdc	0.70 vdc	
Q5 Emitter	10.0 vdc	9.2 vdc	
Q9 Base	0.5 vdc	0.5 vdc	
Q9 Collector	-4.0 vdc	-4.0 vdc	
Q10 Base	0.7 vdc	0.7 vdc	
Q10 Collector	-4.0 vdc	-4.0 vdc	
Common #1	+3.7 vdc	+3.7 vdc	
Q11 Collector	128 vdc	125 vdc	
Q12 Collector	134 vdc	135 vdc	
Common #2	-7.8 vdc	-7.8 vdc	

Table 4 -5. Horizontal Amplifier Board, Voltage Chart



Numbers on waveforms indicate voltages with respect to ground.

B = Base; C = Collector; E = Emitter.

S-1-59

Table 4 -6. Horizontal Amplifier Board Waveforms

SECTION 5

SCHEMATICS AND PARTS' LOCATIONS

5.1 Reference Designators

A	assembly	H	henries
B	board	IF	intermediate frequency
C	capacitor	Incd	incandescent
CR	diode	INT	internal
DS	device signaling (lamp)	k	kilo - 1000
E	misc. electronic part	Lin	linear
F	fuse	Log	logarithmic
FIL	filter	LPF	low pass filter
J	jack	m	milli - 10 ⁻³
K	relay	M	meg - 10 ⁻⁶
L	inductor, RF, choke	MFLM	metal film
M	meter	MFR	manufacturer
CH	chassis	Minat	miniature
FP	front panel	Mom	momentary
RP	rear panel	My	"mylar"
P	plug	N	nano (10 ⁻⁹)
Q	transistor	NC	not connected
R	resistor	N/C	normally closed
Hz	hertz	N/O	normally open
RT	thermistor	NPO	negative positive zero
SW	switch	NRFR	not recommended for field replacement
T	transformer	P	peak
TB	terminal board	PC	printed circuit
TP	test point	pF	picofarads - 10 ⁻¹² farads
V	vacuum & display tubes	PIV	peak inverse voltage
X	crystal	P/O	part of
SCR	silicon controlled rectifier	Poly	polystyrene or polyester
LS	speaker	Porc	porcelain
		Pos	position(s)
		Pot	potentiometer
		P/P	peak-to-peak
		Rect	rectifier
		RF	radio frequency
		S-B	slow-blow
		Semi	semiconductor
		Si	silicon
		Ta	tantalum
		Tog	toggle
		Tol	tolerance
		Trim	trimmer
		TYP	typical
		μ	micro - 10 ⁻⁶
		V	volts
		Var	variable
		VDCW	dc working volts
		W/	with
		W	watts
		ww	wirewound
		W/O	without

5.2 Abbreviations

Amp	amperes	Rect	rectifier
AFC	automatic frequency control	RF	radio frequency
Ampl	amplifier	S-B	slow-blow
BP	bandpass	Semi	semiconductor
CCW	counter-clockwise	Si	silicon
Cer	ceramic	Ta	tantalum
Coef	coefficient	Tog	toggle
Com	common	Tol	tolerance
Comp	composition	Trim	trimmer
Conn	connector	TYP	typical
CRT	cathode-ray tube	μ	micro - 10 ⁻⁶
CW	clockwise	V	volts
DepC	deposited carbon	Var	variable
Electro	electrolytic	VDCW	dc working volts
Encap	encapsulated	W/	with
EXT	external	W	watts
F	farads	ww	wirewound
Ge	germanium	W/O	without
GRD	ground(ed)		

5.3 Schematics, Parts Lists and Pictorials

On the following pages schematics, parts lists and pictorials are grouped according to subassembly. Thus, schematic, parts list, and pictorial of a given PC board are found on adjacent pages.

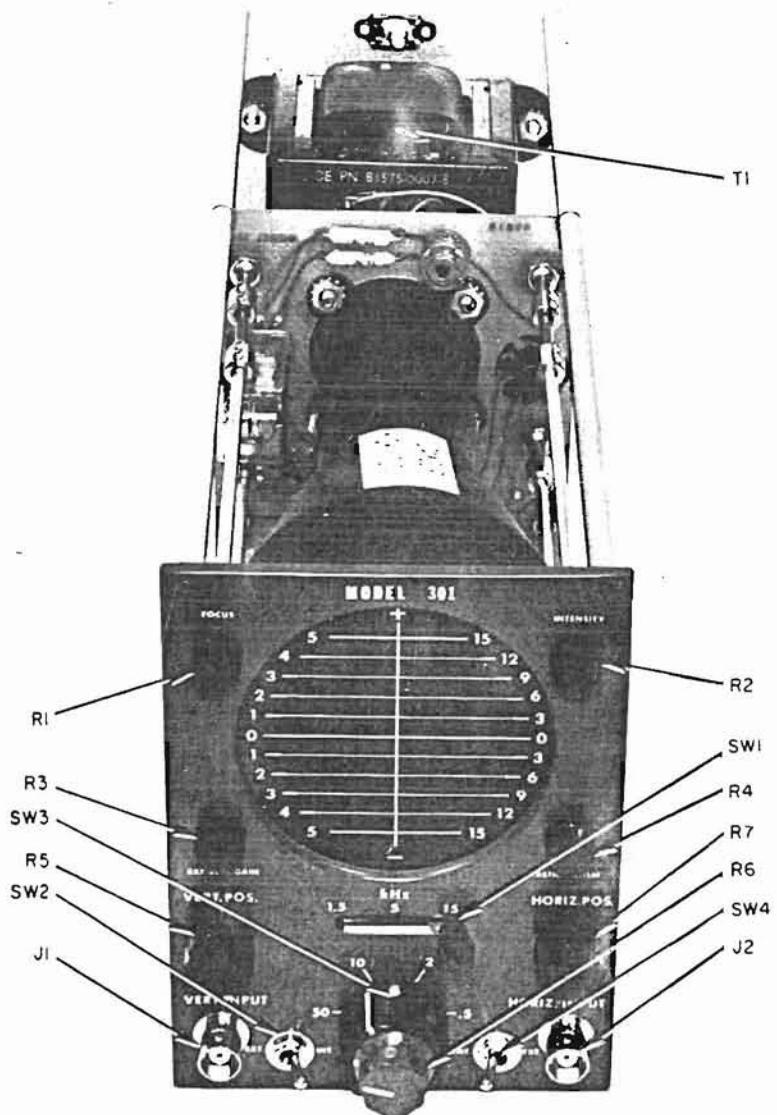
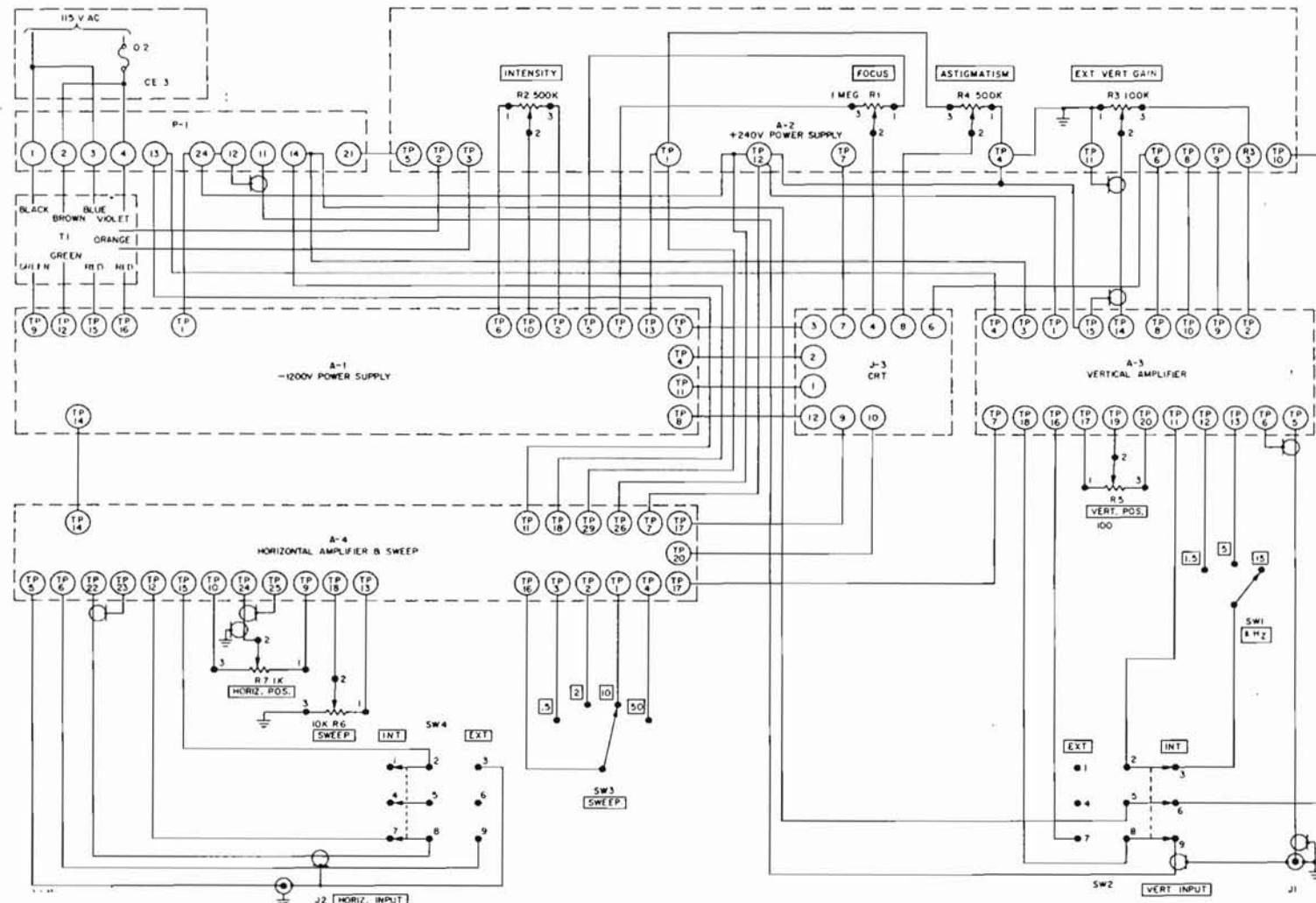


Figure 5-1. Model 301 (FP)

Front Panel & Chassis Components

CKT. REF.	DESCRIPTION	C/E STOCK NO.	MFR.
	RESISTORS		
R1	Pot., comp. 1 megohm, 1/2W	1203-0027	Allen-Bradley
R2	Pot., comp. 500k ohm, 1/2W	1203-0026	Allen-Bradley
R3	Pot., comp. 100k ohm, 1/2W	1203-0025	Allen-Bradley
R4	Pot., comp. 500k ohm, 1/2W	1203-0026	Allen-Bradley
R5	Pot., comp. 100 ohm, 1/2 W	1203-0023	Allen-Bradley
R6	Pot., comp. 10k ohm, 1/2 W	P/O1851-0024	CTS
R7	Pot., comp. 1k ohm, 1/2 W	1203-0024	Allen-Bradley
	SWITCHES		
SW1	Lever, 1P3 pos	1851-0016	Centralab
SW2	Toggle, 3P2 pos	1850-0007	Alco
SW3	Rotary, 1P4 pos	P/O1851-0024	CTS
SW4	Toggle, 3P2 pos	1850-0007	Alco
	JACKS		
J1	Connector	2536-00	Kings
J2	Same as J1		
J3	Tube socket, 12 pin	2605-0011	Cinch
	PLUGS		
P1	Connector, 24 pin, male	2535-0019	
	TRANSFORMERS		
T1	Power Transformer	1575-0007	Basler
	KNOBS		
	Knob, FOCUS	7/16 in dia Black	King-Kasch
	Knob, INTENSITY	7/16 in dia Black	King-Kasch
	Knob, EXT. VERT. GAIN	7/16 in dia Black	King-Kasch
	Knob, ASTIGMATISM	7/16 in dia Black	King-Kasch
	Knob, VERT. POS.	5/8 in dia Black	King-Kasch
	Knob, HOR. POS.	5/8 in dia Black	King-Kasch
	Knob, SWEEP, inner	3/4 in dia Red	King-Kasch
	Knob, SWEEP, outer	1 in dia Black	King-Kasch
	TUBES		
	CRT, 3P1A, 3-inch flat faceplate	1270-0019	RCA
	MISCELLANEOUS		
	Graticule, CRT.	4941-004	Cushman



1. RESISTORS: 1% UNLESS OTHERWISE NOTED
 2. CAPACITORS: 10% UNLESS OTHERWISE NOTED
 3. INDUCTORS: 10% UNLESS OTHERWISE NOTED
 4. - FACTORY SELECTED VALUE
 5. ALL VOLTAGE READINGS ARE DC UNLESS
 OTHERWISE NOTED

WIRING DIAGRAM
MODEL 501 OSCILLOSCOPE

COPYRIGHT 1968 BY CUSHMAN ELECTRONICS INC.
THIS DRAWING IS INTENDED FOR THE OPERATION AND
MAINTENANCE OF THIS EQUIPMENT ONLY.
IT IS NOT TO BE USED OTHERWISE OR REPRODUCED
WITHOUT WRITTEN CONSENT FROM CUSHMAN ELECTRONICS INC.

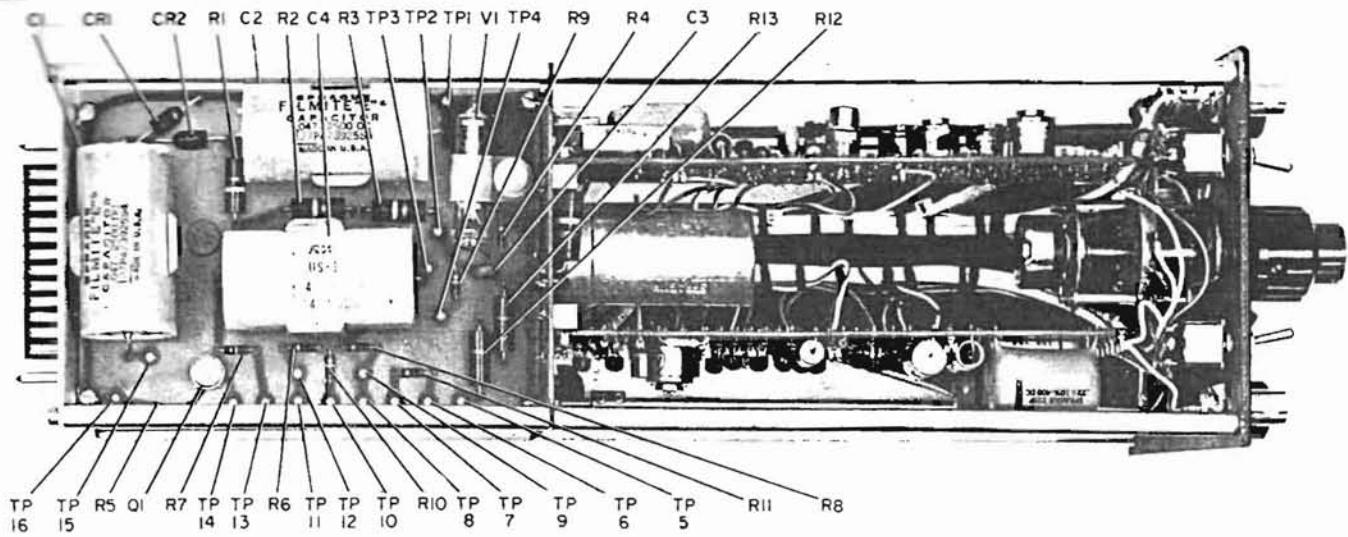
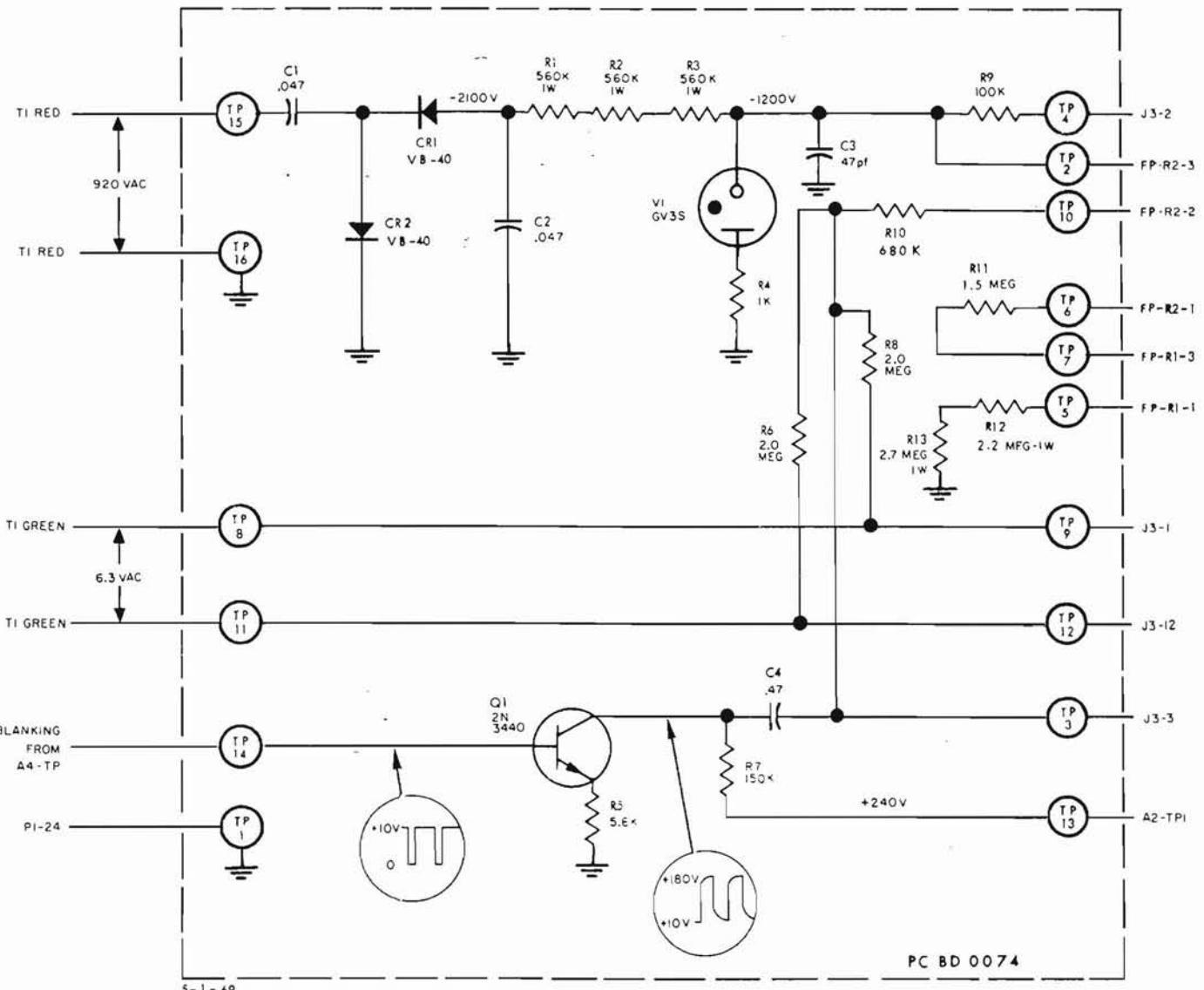


Figure 5-2. High Voltage Power Supply (A1)

Chassis A-1, High Voltage Power Supply, PC Board 0074

CKT. REF.	DESCRIPTION	C/E STOCK NO.	MFR.
CAPACITORS			
C1	Metalized Film, 0.047 μ F \pm 10%, 2500V	1008-0030	Sprague
C2	Metalized Film, 0.047 μ F \pm 10%, 2500V	1008-0030	Sprague
C3	Cer, 47 pF \pm 20%, 3000V	1005-0045	Centralab
C4	Mylar, 0.47 μ F \pm 20%, 1400V	1008-0044	TRW
DIODES			
CR1	Si, VB-40	1281-0029	Varo
CR2	Si, VB-40	1281-0029	Varo
TRANSISTORS			
Q1	Si, 2N3440	1271-0004	RCA
RESISTORS			
R1	Comp, 560k ohm \pm 5%, 1W	1068-5645	Allen-Bradley
R2	Comp, 560k ohm \pm 5%, 1W	1068-5645	Allen-Bradley
R3	Comp, 560k ohm \pm 5%, 1W	1068-5645	Allen-Bradley
R4	Comp, 1k ohm \pm 5%, 1/4W	1066-1025	Allen-Bradley
R5	Comp, 5.6k ohm \pm 5%, 1/4W	1066-5625	Allen-Bradley
R6	Comp, 2M ohm \pm 5%, 1/4W	1066-2055	Allen-Bradley
R7	Comp, 150k ohm \pm 5%, 1/4W	1066-1545	Allen-Bradley
R8	Comp, 2M ohm \pm 5%, 1/4W	1066-2055	Allen-Bradley
R9	Comp, 100k ohm \pm 5%, 1/4W	1066-1045	Allen-Bradley
R10	Comp, 680k ohm \pm 5%, 1/4W	1066-6845	Allen-Bradley
R11	Comp, 1.5M ohm \pm 5%, 1/4W	1066-1555	Allen-Bradley
R12	Comp, 2.2M ohm \pm 5%, 1W	1068-2255	Allen-Bradley
R13	Comp, 2.7M ohm \pm 5%, 1W	1068-2755	Allen-Bradley
TUBES			
V1	Regulator, 1200V \pm 1.5%, GV3S	1281-0017	Victoreen



1. RESISTORS: 1/4W 5% VALUES IN OHMS UNLESS OTHERWISE NOTED
2. FACTOR 10 VALUES IN μ UNLESS OTHERWISE NOTED
3. INDUCTORS VALUES IN μ H UNLESS OTHERWISE NOTED
4. - FACTORY SELECTED VALUE
5. ALL VOLTAGE READINGS ARE DC UNLESS OTHERWISE NOTED

-1200V POWER SUPPLY
MODEL 301 OSCILLOSCOPE

COPYR. GT 1968 BY CUSHMAN ELECTRONICS INC.
THIS DRAWING IS INTENDED FOR THE OPERATION AND
MAINTENANCE OF CUSHMAN ELECTRONICS EQUIPMENT
AND IS NOT TO BE USED OTHERWISE OR REPRODUCED
WITHOUT WRITTEN CONSENT FROM CUSHMAN ELECTRONICS INC.

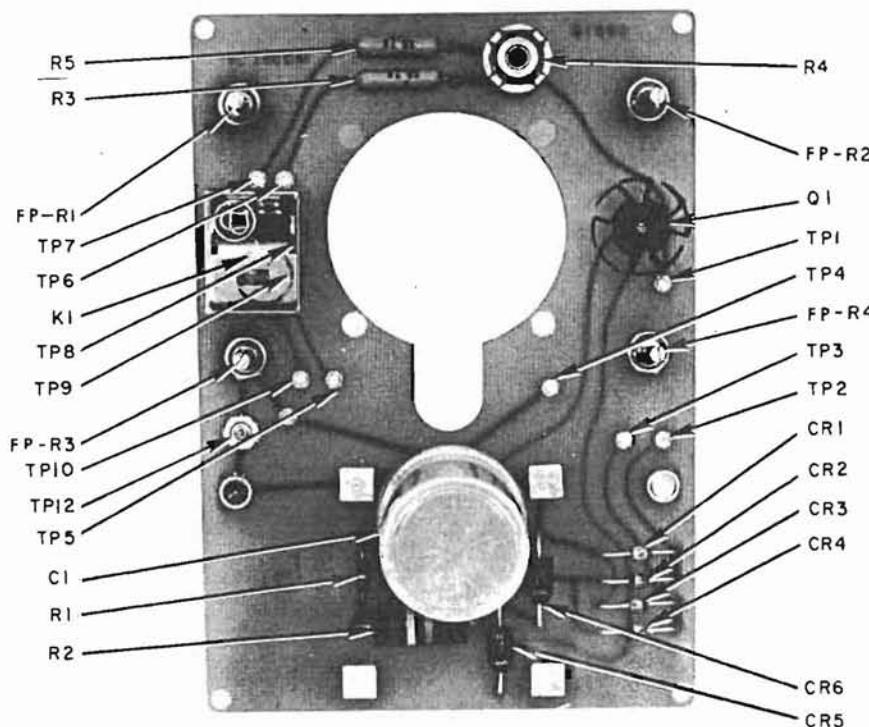
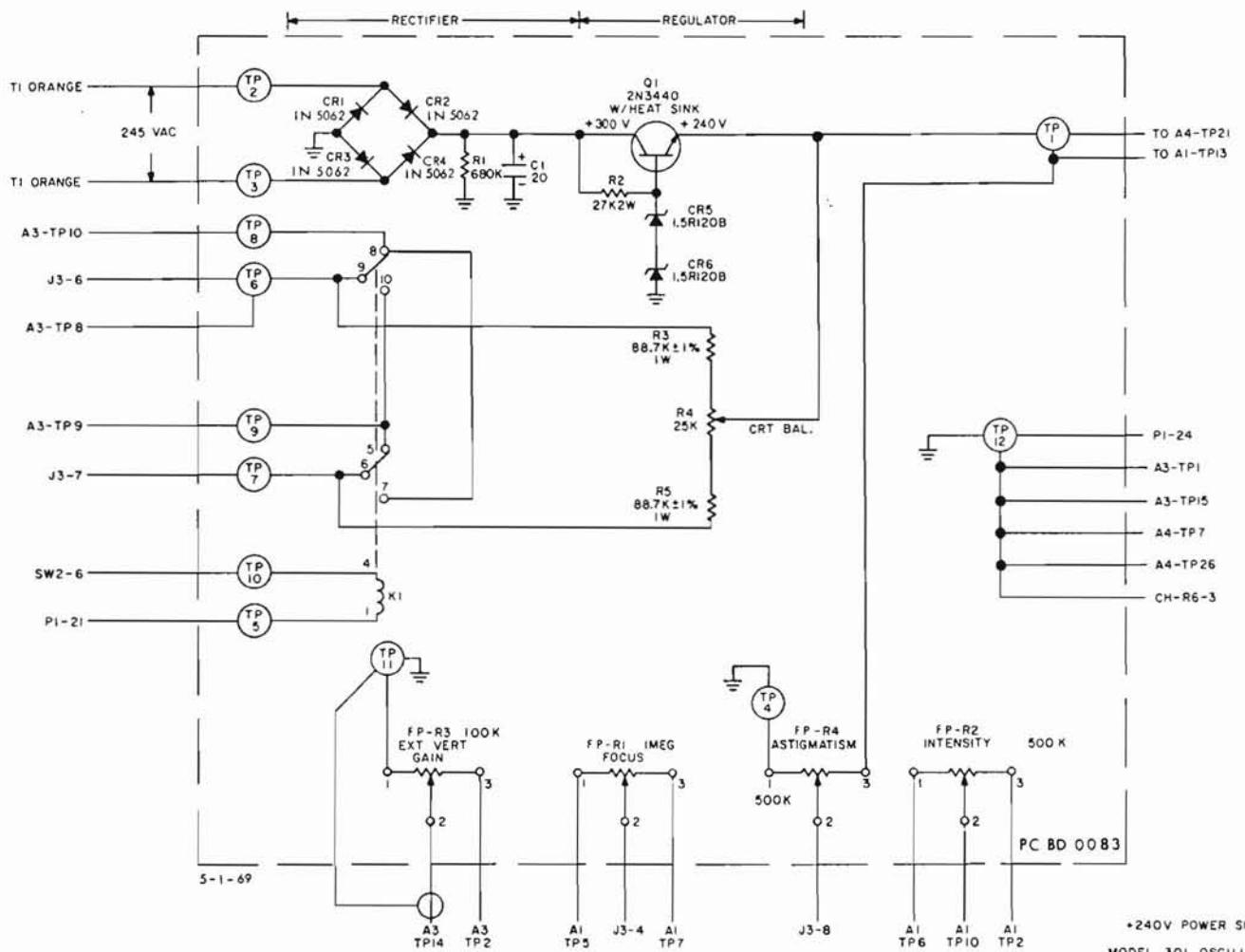


Figure 5-3. Low Voltage Power Supply (A2)

Chassis A-2, Low Voltage Power Supply, PC Bd 0083

CKT. REF.	DESCRIPTION	C/E STOCK NO.	MFR.
C1	CAPACITORS Electro, 20 μ F, 450V	1010-0004	Sprague
CR1	DIODES Si, 1N-5062	1281-0030	General Electric
CR2	Si, 1N-5062	1281-0030	General Electric
CR3	Si, 1N-5062	1281-0030	General Electric
CR4	Si, 1N-5062	1281-0030	General Electric
CR5	Si, Zener, 120V, 1.5R120B	1281-0022	Solitron
CR6	Si, Zener, 120V, 1.5R120B	1281-0022	Solitron
K1	RELAYS 12 VDC, DPDT	1313-0002	Parelco
Q1	TRANSISTORS Si, NPN, 2N3440	1271-0004	RCA
R1	RESISTORS Comp. 680k ohm \pm 5%, 1/4W	1066-6845	Allen-Bradley
R2	Comp. 27k ohm \pm 5%, 2W	1069-2735	Allen-Bradley
R3	M Film, 88.7k ohm \pm 1%, 1W	1076-0008	Electra
R4	Pot., comp, 25k ohm, 1/2W	1200-0014	Allen-Bradley
R5	M Film, 88.7k ohm \pm 1%, 1W	1076-0008	Electra



1. RESISTORS - 1/2W 5% VALUES IN OHMS UNLESS OTHERWISE NOTED
2. CAPACITORS - VALUES IN μ F UNLESS OTHERWISE NOTED
3. INDUCTORS - VALUES IN μ H UNLESS OTHERWISE NOTED
4. - FACTORY SELECTED VALUE
5. ALL VOLTAGE READINGS ARE DC UNLESS OTHERWISE NOTED

COPYRIGHT 1968 BY CUSHMAN ELECTRONICS, INC.
THIS DRAWING IS INTENDED FOR THE OPERATION AND
MAINTENANCE OF CUSHMAN ELECTRONICS EQUIPMENT
AND IS NOT TO BE USED OTHERWISE OR REPRODUCED
WITHOUT WRITTEN CONSENT FROM CUSHMAN ELECTRONICS, INC.

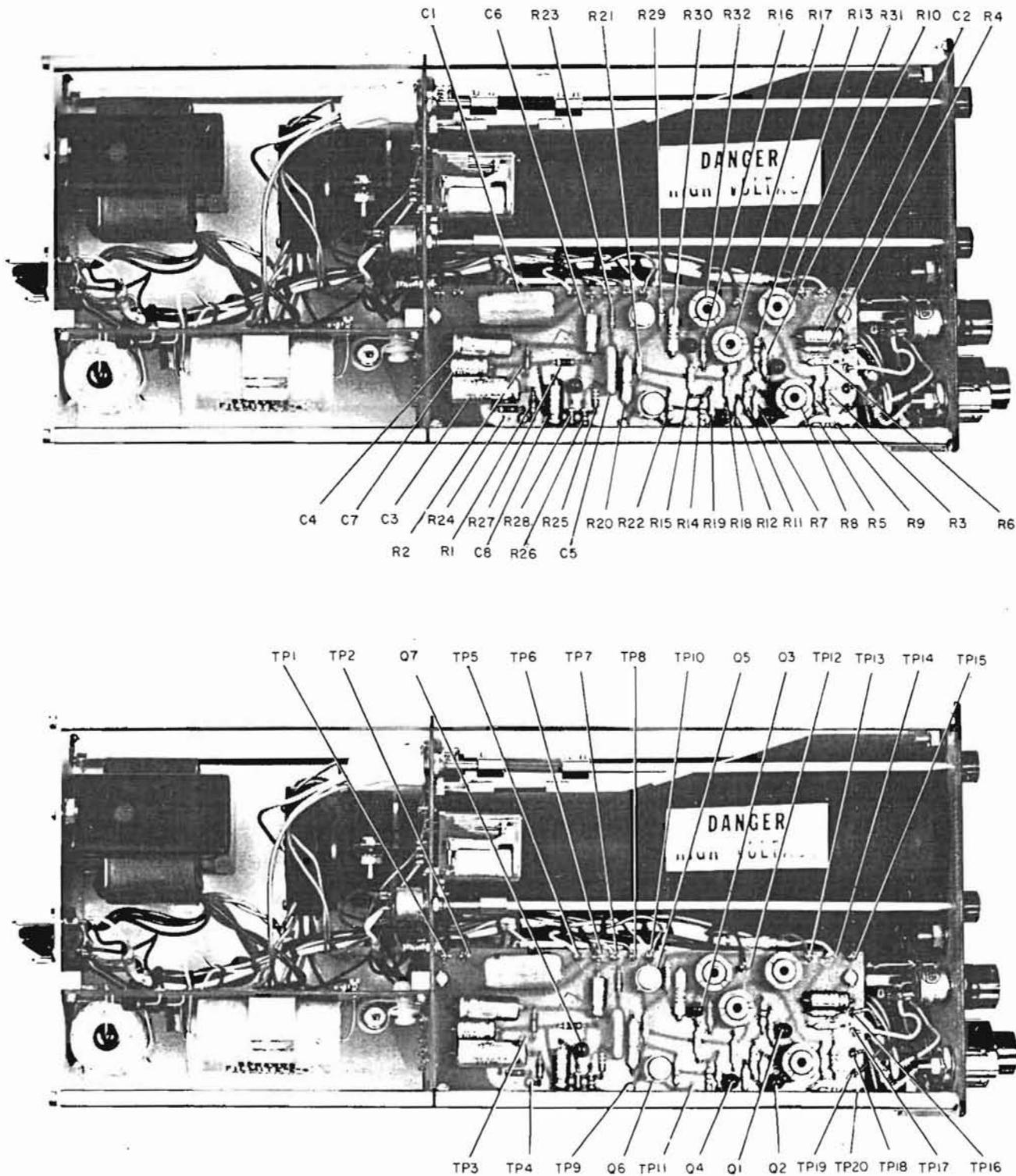
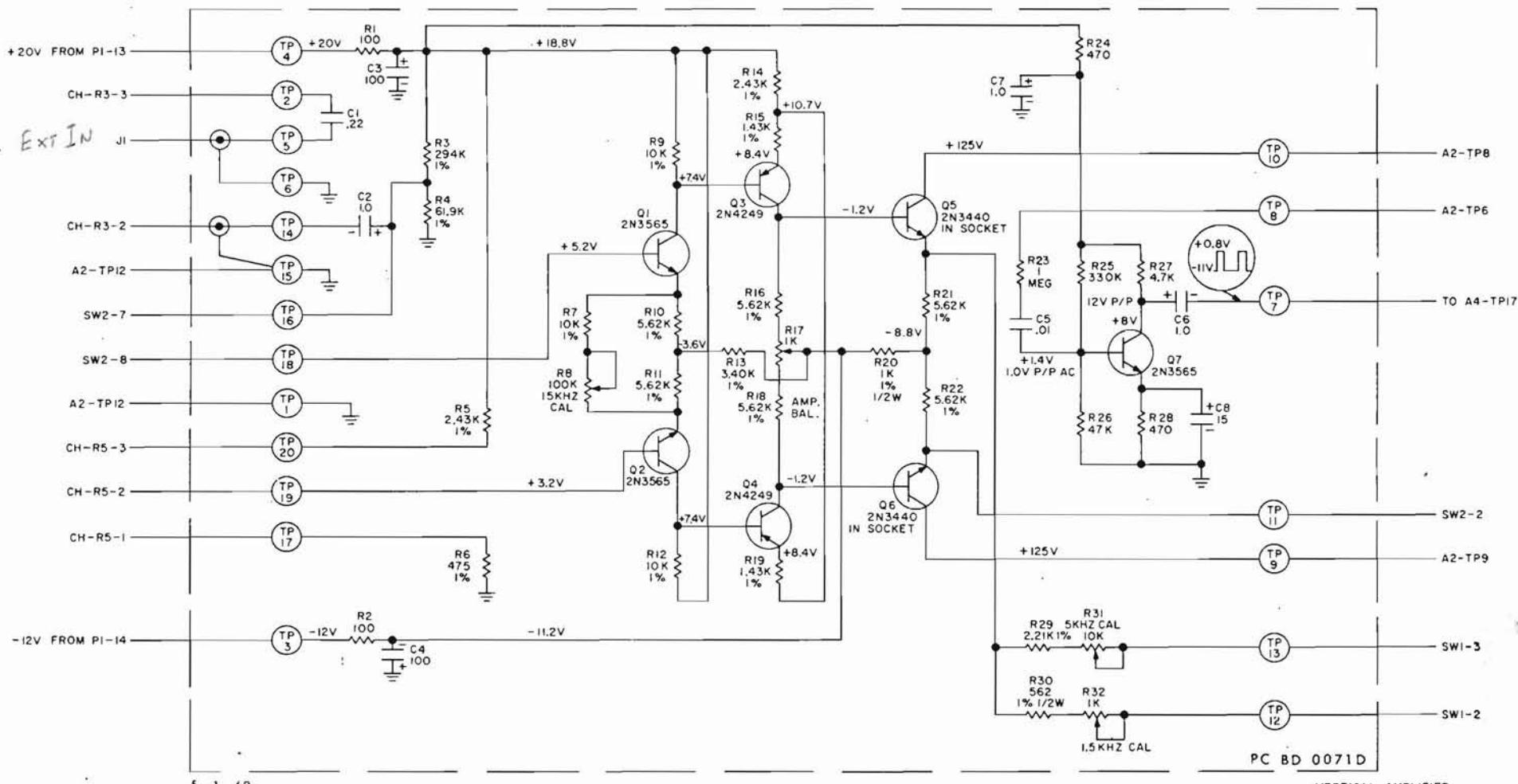


Figure 5-4. Vertical Amplifier (A3)

Chassis A-3, Vertical Amplifier, PC Board 0071D

CKT. REF.	DESCRIPTION	C/E STOCK NO.	MFR.
	CAPACITORS		
C1	Poly, 0.22 μ F \pm 10%, 400V	1008-0042	Sprague
C2	Electro, 1 μ F, 25V	1013-0004	Sprague
C3	Electro, 100 μ F, 25V	1013-0003	Sprague
C4	Electro, 100 μ F, 12V	1013-0011	Sprague
C5	Poly, 0.01 μ F \pm 10%, 200V	1008-0043	Sprague
C6	Electro, 1 μ F, 25V	1013-0004	Sprague
C7	Electro, 1 μ F, 25V	1013-0004	Sprague
C8	Electro, 15 μ F, 6V	1013-0009	Sprague
	TRANSISTORS		
Q1	Si, NPN, 2N3565	1272-0017	Fairchild
Q2	Si, NPN, 2N3565	1272-0017	Fairchild
Q3	Si, PNP, 2N4249	1272-0024	Fairchild
Q4	Si, PNP, 2N4249	1272-0024	Fairchild
Q5	Si, NPN, 2N3440	1271-0004	RCA
Q6	Si, NPN, 2N3440	1271-0004	RCA
Q7	Si, NPN, 2N3565	1272-0017	Fairchild
	RESISTORS		
R1	Comp, 100 ohm \pm 5%, 1/4W	1066-1015	Allen-Bradley
R2	Comp, 100 ohm \pm 5%, 1/4W	1066-1015	Allen-Bradley
R3	M Film, 294k ohm \pm 1%, 1/4W	1075-0028	Electra
R4	M Film, 61.9k ohm \pm 1%, 1/4W	1075-0018	Electra
R5	M Film, 2.43k ohm \pm 1%, 1/4W	1075-0019	Electra
R6	M Film, 475 ohm \pm 1%, 1/4W	1075-0023	Electra
R7	M Film, 10k ohm \pm 1%, 1/4W	1075-0009	Electra
R8	Pot., comp, 100k ohm, 1/2W	1200-0015	Allen-Bradley
R9	M Film, 10k ohm \pm 1%, 1/4W	1075-0009	Electra
R10	M Film, 5.62k ohm \pm 1%, 1/4W	1075-0013	Electra
R11	M Film, 5.62k ohm \pm 1%, 1/4W	1075-0013	Electra
R12	M Film, 10k ohm \pm 1%, 1/4W	1075-0009	Electra
R13	M Film, 3.4k ohm \pm 1%, 1/4W	1075-0020	Electra
R14	M Film, 2.43k ohm \pm 1%, 1/4W	1075-0019	Electra
R15	M Film, 1.43k ohm \pm 1%, 1/4W	1075-0021	Electra
R16	M Film, 5.62k ohm \pm 1%, 1/4W	1075-0013	Electra
R17	Pot., comp, 1k ohm, 1/2W	1200-0012	Allen-Bradley
R18	M Film, 5.62k ohm \pm 1%, 1/4W	1075-0013	Electra
R19	M Film, 1.43k ohm \pm 1%, 1/4W	1075-0021	Electra
R20	M Film, 1k ohm \pm 1%, 1/2W	1076-0007	Electra
R21	M Film, 5.62k ohm \pm 1%, 1/4W	1075-0013	Electra
R22	M Film, 5.62k ohm \pm 1%, 1/4W	1075-0013	Electra
R23	Comp, 1M ohm \pm 5%, 1/4W	1066-1055	Allen-Bradley
R24	Comp, 470 ohm \pm 5%, 1/4W	1066-4715	Allen-Bradley
R25	Comp, 330k ohm \pm 5%, 1/4W	1066-3345	Allen-Bradley
R26	Comp, 47k ohm \pm 5%, 1/4W	1066-4735	Allen-Bradley
R27	Comp, 4.7k ohm \pm 5%, 1/4W	1066-4725	Allen-Bradley
R28	Comp, 470 ohm \pm 5%, 1/4W	1066-4715	Allen-Bradley
R29	M Film, 2.21k ohm \pm 1%, 1/4W	1075-0010	Electra
R30	M Film, 562 ohm \pm 1%, 1/2W	1076-0006	Electra
R31	Pot., comp, 10k ohm, 1/2W	1200-0013	Allen-Bradley
R32	Pot., comp, 1k ohm, 1/2W	1200-0012	Allen-Bradley



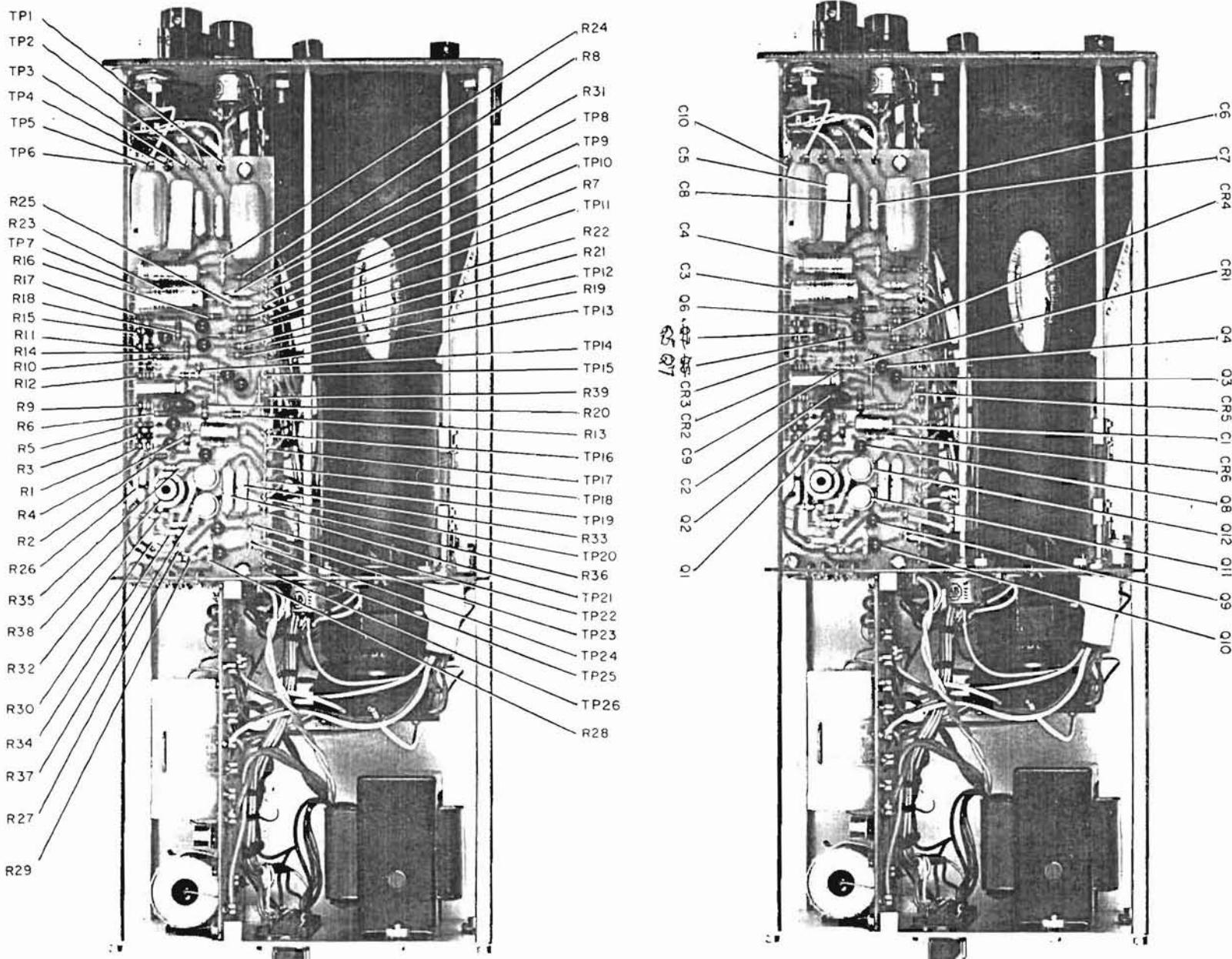
1. RESISTORS: 1/4W 5% VALUES IN OHMS UNLESS OTHERWISE NOTED.
 2. CAPACITORS: VALUES IN μ F UNLESS OTHERWISE NOTED.
 3. JUMPS: 100% UNLESS OTHERWISE NOTED.
 4. -12V TYPICALLY SELECTED VALUE.
 5. ALL VOLTAGE READINGS ARE DC UNLESS OTHERWISE NOTED.

COPYRIGHT 1968 BY CUSHMAN ELECTRONICS INC.
 THIS DRAWING IS INTENDED FOR THE OPERATION AND
 MAINTENANCE OF CUSHMAN ELECTRONIC EQUIPMENT
 AND IS NOT TO BE USED OTHERWISE OR REPRODUCED
 WITHOUT WRITTEN CONSENT FROM CUSHMAN ELECTRONICS, INC.

MODEL 301 OSCILLOSCOPE

VERTICAL AMPLIFIER

Figure 5-5. Horizontal Amplifier and Sweep (A4)



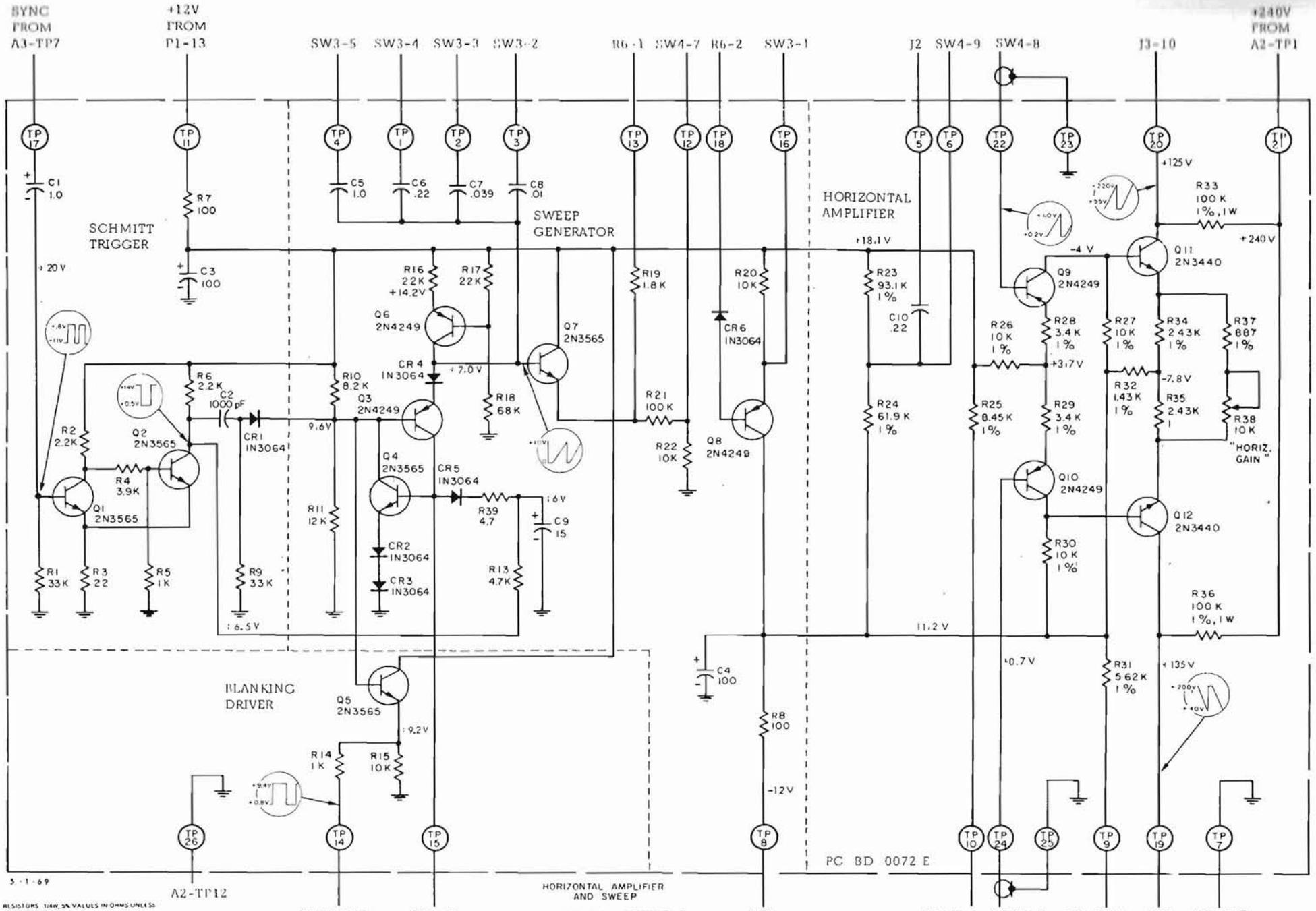
Chassis A-4, Horizontal Amplifier and Sweep, PC Board 0072E

CKT. REF.	DESCRIPTION	C/E STOCK NO.	MFR.
	CAPACITORS		
C1	Electro, 1 μ F, 25V	1013-0004	Sprague
C2	Mica, 1000 pF \pm 5%, 100V	1002-0015	Elmenco
C3	Electro, 100 μ F, 25V	1013-0003	Sprague
C4	Electro, 100 μ F, 12V	1013-0011	Sprague
C5	Metalized Mylar, 1.0 μ F \pm 10%, 100V	1008-0033	TRW
C6	Poly, 0.22 μ F \pm 10%, 400V	1008-0042	Sprague
C7	Poly, 0.039 μ F \pm 10%, 100V	1008-0045	Sprague
C8	Poly, 0.01 μ F \pm 10%, 200V	1008-0043	Sprague
C9	Electro, 15 μ F, 25V	1013-0005	Sprague
C10	Poly, 0.22 μ F \pm 10%, 400V	1008-0042	Sprague
	DIODES		
CR1	Si, 1N3064	1281-0013	Sylvania
CR2	Si, 1N3064	1281-0013	Sylvania
CR3	Si, 1N3064	1281-0013	Sylvania
CR4	Si, 1N3064	1281-0013	Sylvania
CR5	Si, 1N3064	1281-0013	Sylvania
CR6	Si, 1N3064	1281-0013	Sylvania
	TRANSISTORS		
Q1	Si, NPN, 2N3565	1272-0017	Fairchild
Q2	Si, NPN, 2N3565	1272-0017	Fairchild
Q3	Si, PNP, 2N4249	1272-0024	Fairchild
Q4	Si, NPN, 2N3565	1272-0017	Fairchild
Q5	Si, NPN, 2N3565	1272-0017	Fairchild
Q6	Si, PNP, 2N4249	1272-0024	Fairchild
Q7	Si, NPN, 2N3565	1272-0017	Fairchild
Q8	Si, PNP, 2N4249	1272-0024	Fairchild
Q9	Si, PNP, 2N4249	1272-0024	Fairchild
Q10	Si, PNP, 2N4249	1272-0024	Fairchild
Q11	Si, NPN, 2N3440	1271-0004	RCA
Q12	Si, NPN, 2N3440	1271-0004	RCA
	RESISTORS		
R1	Comp, 33k ohm \pm 5%, 1/4W	1066-3335	Allen-Bradley
R2	Comp, 2.2k ohm \pm 5%, 1/4W	1066-2225	Allen-Bradley
R3	Comp, 22 ohm \pm 5%, 1/4W	1066-2205	Allen-Bradley
R4	Comp, 3.9k ohm \pm 5%, 1/4W	1066-3925	Allen-Bradley
R5	Comp, 1k ohm \pm 5%, 1/4W	1066-1025	Allen-Bradley
R6	Comp, 2.2k ohm \pm 5%, 1/4W	1066-2225	Allen-Bradley
R7	Comp, 100 ohm \pm 5%, 1/4W	1066-1015	Allen-Bradley
R8	Comp, 100 ohm \pm 5%, 1/4W	1066-1015	Allen-Bradley
R9	Comp, 33k ohm \pm 5%, 1/4W	1066-3335	Allen-Bradley
R10	Comp, 8.2k ohm \pm 5%, 1/4W	1066-8225	Allen-Bradley
R11	Comp, 12k ohm \pm 5%, 1/4W	1066-1235	Allen-Bradley
R12	Comp, 33k ohm \pm 5%, 1/4W	1066-3335	Allen-Bradley
R13	Comp, 4.7k ohm \pm 5%, 1/4W	1066-4725	Allen-Bradley
R14	Comp, 1k ohm \pm 5%, 1/4W	1066-1025	Allen-Bradley
R15	Comp, 10k ohm \pm 5%, 1/4W	1066-1035	Allen-Bradley

Chassis A-4, Horizontal Amplifier and Sweep, PC Board 0072E, continued

CKT.	DESCRIPTION	C/E STOCK NO.	MFR.
R16	Comp. 22k ohm $\pm 5\%$, 1/4W	1066-2235	Allen-Bradley
R17	Comp. 22k ohm $\pm 5\%$, 1/4W	1066-2235	Allen-Bradley
R18	Comp. 68k ohm $\pm 5\%$, 1/4W	1066-6835	Allen-Bradley
R19	Comp. 1.8k ohm $\pm 5\%$, 1/4W	1066-1825	Allen-Bradley
R20	Comp. 10k ohm $\pm 5\%$, 1/4W	1066-1035	Allen-Bradley
R21	Comp. 100k ohm $\pm 5\%$, 1/4W	1066-1045	Allen-Bradley
R22	Comp. 10k ohm $\pm 5\%$, 1/4W	1066-1035	Allen-Bradley
R23	M Film, 93.1k ohm $\pm 1\%$, 1/4W	1075-0029	Electra
R24	M Film, 61.9k ohm $\pm 1\%$, 1/4W	1075-0018	Electra
R25	M Film, 8.45k ohm $\pm 1\%$, 1/4W	1075-0031	Electra
R26	M Film, 10k ohm $\pm 1\%$, 1/4W	1075-0009	Electra
R27	M Film, 10k ohm $\pm 1\%$, 1/4W	1075-0009	Electra
R28	M Film, 3.4k ohm $\pm 1\%$, 1/4W	1075-0020	Electra
R29	M Film, 3.4k ohm $\pm 1\%$, 1/4W	1075-0020	Electra
R30	M Film, 10k ohm $\pm 1\%$, 1/4W	1075-0009	Electra
R31	M Film, 5.62k ohm $\pm 1\%$, 1/4W	1075-0013	Electra
R32	M Film, 1.43k ohm $\pm 1\%$, 1/4W	1075-0021	Electra
R33	M Film, 100k ohm $\pm 1\%$, 1W	1076-0009	Electra
R34	M Film, 2.43k ohm $\pm 1\%$, 1/4W	1075-0019	Electra
R35	M Film, 2.43k ohm $\pm 1\%$, 1/4W	1075-0019	Electra
R36	M Film, 100k ohm $\pm 1\%$, 1W	1076-0009	Electra
R37	M Film, 887 ohm $\pm 1\%$, 1/4W	1075-0022	Electra
R38	Pot., comp, 10k ohm, 1/2W	1200-0013	Allen-Bradley
R39	Comp. 4.7k ohm $\pm 5\%$, 1/4W	1066-4725	Allen-Bradley

□



COPYRIGHT 1968 BY CUSHMAN ELECTRONICS, INC.
 THIS DRAWING IS INTENDED FOR THE OPERATION AND
 MAINTENANCE OF CUSHMAN ELECTRONICS EQUIPMENT
 AND IS NOT TO BE USED OTHERWISE OR REPRODUCED
 WITHOUT WRITTEN CONSENT FROM CUSHMAN ELECTRONICS, INC.

-12V
 FROM P1-14

Variaciones en 10 mils.

